

1. Determine if each of the following are valid autocorrelation matrices of WSS processes. (Correlation Matrix)

$$\mathbf{R}_a = \begin{bmatrix} 4 & 1 & 1 \\ -1 & 4 & 1 \\ -1 & -1 & 4 \end{bmatrix}, \mathbf{R}_b = \begin{bmatrix} 2 & 1 & 1 \\ 1 & 2 & 0 \\ 1 & 0 & 2 \end{bmatrix}, \mathbf{R}_c = \begin{bmatrix} 2j & 0 & j \\ 0 & 2j & 0 \\ -j & 0 & 2j \end{bmatrix}, \mathbf{R}_d = \begin{bmatrix} 1 & 0 & 2 \\ 0 & 1 & 0 \\ 2 & 0 & 1 \end{bmatrix}.$$

*Solution:*

Recall that the properties of an autocorrelation matrix for a WSS process is that (1)  $\mathbf{R}$  is Toeplitz; (2)  $\mathbf{R}^H = \mathbf{R}$ ; (3)  $\mathbf{R}$  is non-negative definite.

$\mathbf{R}_a$  is NOT Hermitian;  $\mathbf{R}_b$  is NOT Toeplitz;  $\mathbf{R}_c$  is NOT Hermitian;  $\mathbf{R}_d$  is NOT non-negative definite ( $\lambda = 1, -1, 3$ ).

2. Consider the random process  $y(n) = x(n) + v(n)$ , where  $x(n) = Ae^{j(\omega n + \phi)}$  and  $v(n)$  is zero mean white Gaussian noise with a variance  $\sigma_v^2$ . We also assume the noise and the complex sinusoid are independent. Under the following conditions, determine if  $y(n)$  is WSS. Justify your answers. (WSS Process)

(a)  $\omega$  and  $A$  are constants, and  $\phi$  is a uniformly distributed over the interval  $[0, 2\pi]$ .

(b)  $\omega$  and  $\phi$  are constants, and  $A$  is a Gaussian random variable  $\sim \mathcal{N}(0, \sigma_A^2)$ .

(c)  $\phi$  and  $A$  are constants, and  $\omega$  is a uniformly distributed over the interval  $[\omega_0 - \Delta, \omega_0 + \Delta]$  for some fixed  $\Delta$ .

*Solution:*

(a)

$$\begin{aligned} E[y(n)] &= Ae^{j\omega n} E_\phi[e^{j\phi}] + E_v[v(n)] = 0 \\ E[y(n)y^*(n-k)] &= E_\phi[(Ae^{j(\omega n + \phi)} + v(n))(A^*e^{-j(\omega(n-k) + \phi)} + v^*(n-k))] \\ &= |A|^2 E_\phi[e^{j\omega k}] + \sigma_v^2 \delta(k) \\ &= |A|^2 e^{j\omega k} + \sigma_v^2 \delta(k) \end{aligned}$$

1st and 2nd moments are independent of  $n$ . Thus, the process is WSS.

(b)

$$\begin{aligned} E[y(n)] &= E_A[A]e^{j(\omega n + \phi)} + E_v[v(n)] = 0 \\ E[y(n)y^*(n-k)] &= E_A[(Ae^{j(\omega n + \phi)} + v(n))(A^*e^{-j(\omega(n-k) + \phi)} + v^*(n-k))] \\ &= E_A[AA^*]e^{j\omega k} + \sigma_v^2 \delta(k) \\ &= \sigma_A^2 e^{j\omega k} + \sigma_v^2 \delta(k) \end{aligned}$$

1st and 2nd moments are independent of  $n$ . Thus, the process is WSS.

(c)

$$\begin{aligned}
 E[y(n)] &= E_\omega[x(n)] + E_v[v(n)] = A \cdot E_\omega[e^{j\omega n}] \cdot e^{j\phi} = \frac{Ae^{j\phi}}{2jn\Delta} e^{j\omega n} \Big|_{\omega_0-\Delta}^{\omega_0+\Delta} \\
 \Rightarrow |E[y(n)]| &\leq \left| \frac{Ae^{j\phi}}{2jn\Delta} \right| \cdot 2 \rightarrow 0 \text{ as } n \rightarrow \infty \\
 E[y(n)y^*(n-k)] &= E_\omega[(Ae^{j(\omega n+\phi)} + v(n))(A^*e^{-j(\omega(n-k)+\phi)} + v^*(n-k))] \\
 &= |A|^2 E_\omega[e^{j\omega k}] + \sigma_v^2 \delta(k) \\
 &= |A|^2 e^{j\omega_0 k} \frac{\sin(k\Delta)}{k\Delta} + \sigma_v^2 \delta(k)
 \end{aligned}$$

The sequence defined here is actually NOT a WSS process, but its 1st and 2nd moment statistics are approximately independent of  $n$  as  $n \rightarrow \infty$ .

**3.** [Rec.II P2(a) revisited] Determine the PSD of the WSS process  $y(n) = Ae^{j(\omega_0 n + \phi)} + v(n)$ , where  $v(n)$  is zero mean white Gaussian noise with a variance  $\sigma_v^2$ , and  $\phi$  is uniformly distributed over the interval  $[0, 2\pi]$ . (Power Spectral Density)

*Solution:*

In the autocorrelation function in P2(a) is

$$r_y(k) = A^2 e^{j\omega k} + \sigma_v^2 \delta(k)$$

By taking discrete time Fourier transform on  $r_y(k)$ , we get

$$P_y(\omega) = 2\pi A^2 \delta(\omega - \omega_0) + \sigma_v^2$$

**4.** Assume  $v(n)$  is a white Gaussian random process with zero mean and variance 1. The two filters in Fig. RII.4 are  $G(z) = \frac{1}{1-0.4z^{-1}}$  and  $H(z) = \frac{2}{1-0.5z^{-1}}$ . (Auto-Regressive Process)

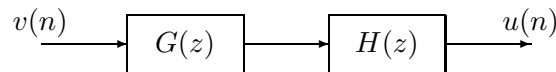


Figure RII.4:

(a) Is  $u(n)$  an AR process? If so, find the parameters.

(b) Find the autocorrelation coefficients  $r_u(0)$ ,  $r_u(1)$ , and  $r_u(2)$  of the process  $u(n)$ .

*Solution:*

(a)  $U(z) = \frac{2}{1-0.9z^{-1}+0.2z^{-2}}V(z)$ ,  $u(n) = 0.9u(n-1) - 0.2u(n-2) + 2v(n)$ ,  $a_1 = -0.9$ ,  $a_2 = 0.2$ .

(b) Apply the Yule-Walker equation,

$$\begin{pmatrix} r_u(0) & r_u(1) \\ r_u(1) & r_u(0) \end{pmatrix} \begin{pmatrix} -0.9 \\ 0.2 \end{pmatrix} = - \begin{pmatrix} r_u(1) \\ r_u(2) \end{pmatrix},$$

from which we get

$$\begin{cases} r_u(1) = -\frac{a_1}{1+a_2}r_u(0) = \frac{3}{4}r_u(0) \\ r_u(2) = \left(\frac{a_1^2}{1+a_2} - a_2\right)r_u(0) = \frac{19}{40}r_u(0) \end{cases}$$

Moreover, since  $r_u(0) + a_1r_u(1) + a_2r_u(2) = 4\sigma_v^2$  (Here, ‘4’ because in this model it is ‘ $2v(n)$ ’ rather than ‘ $v(n)$ ’), we have  $r_u(0) = \frac{1+a_2}{1-a_2} \frac{4\sigma_v^2}{(1+a_2)^2 - a_1^2} = \frac{200}{21}$ . Then,  $r_u(1) = \frac{50}{7}$ , and  $r_u(2) = \frac{95}{21}$ .

*Note:*

1. In general, for a  $p$ -order AR model, given  $\{\sigma_v^2, a_1, a_2, \dots, a_p\}$ , we can find  $\{r(0), r(1), r(2), \dots\}$ ; and vice versa. They are related by Yule-Walker Equations.

2.  $r(-k) = r^*(k)$  in general (and hence matrix  $\mathbf{R}$  is Hermitian), and  $r(-k) = r(k)$  for real-valued signals.  $r(0)$  is the power of sequence  $u(n)$ , and hence  $r(0) > 0$  from physical point of view.

3. For an AR model,  $u(n) = \sum_{k=1}^p -a_k u(n-k) + v(n)$  has NO correlation with future  $v(m)$ ,  $m = n+1, n+2, \dots$  (convince yourself). Simply multiply both sides by  $u^*(n)$  and take expectation, we get  $r(0) = \sum_{k=1}^p -a_k r(-k) + E(v(n)u^*(n))$ . Note that  $E(v(n)u^*(n)) = E(v(n)(\sum_{k=1}^p -a_k^* u^*(n-k) + v^*(n)))$  but  $E(v(n)u^*(n-k)) = 0$  for  $k \geq 1$ . Then,  $r(0) = \sum_{k=1}^p -a_k r(-k) + \sigma_v^2$ , which we have used to find the relation of  $r(0)$  (signal power) and  $\sigma_v^2$  (model parameter) in part (b). We could multiply  $u^*(n-k)$  instead of  $u^*(n)$  and take the expectation, and this is how the Yule-Walker equations are derived.

5. Let a real-valued AR(2) process be described by

$$u(n) = x(n) + a_1 x(n-1) + a_2 x(n-2)$$

where  $u(n)$  is a white noise of zero-mean and variance  $\sigma^2$ , and  $u(n)$  and past values  $x(n-1)$ ,  $x(n-2)$  are uncorrelated. (Yule-Walker Equation)

(a) Determine and solve the Yule-Walker Equations for the AR process.

(b) Find the variance of the process  $x(n)$ .

*Solution:* (a) Solve the Yule-Walker equation, we have

$$\begin{aligned} r_x(0) &= -a_1 r_x(-1) - a_2 r_x(-2) + \sigma^2 \\ r_x(1) &= -a_1 r_x(0) - a_2 r_x(-1) \\ r_x(2) &= -a_1 r_x(1) - a_2 r_x(0) \end{aligned}$$

Use the relation that  $r_x(k) = r_x(-k)$  and solve this we get

$$\begin{aligned}
r_x(0) &= \frac{\sigma^2}{1 - \frac{a_1^2}{1+a_2} + a_2(\frac{a_1^2}{1+a_2} - a_2)} \\
r_x(1) &= -\frac{a_1}{1+a_2} r_x(0) \\
r_x(2) &= (\frac{a_1^2}{1+a_2} - a_2) r_x(0)
\end{aligned}$$

(b) The process is zero mean, so the variance is  $r_x(0)$ .

6. [Problem II.4 continued] Assume  $v(n)$  and  $w(n)$  are white Gaussian random processes with zero mean and variance 1. The two filters in Fig. RII.6 are  $G(z) = \frac{1}{1-0.4z^{-1}}$  and  $H(z) = \frac{2}{1-0.5z^{-1}}$ . (Wiener Filter)

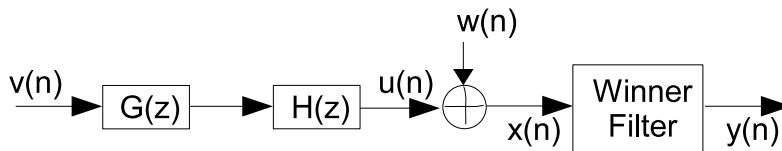


Figure RII.6:

- (a) Design a 1-order Wiener filter such that the desired output is  $u(n)$ . What is the MSE?
- (b) Design a 2-order Wiener filter. What is the MSE?

7. The autocorrelation sequence of a given zero-mean real-valued random process  $u(n)$  is  $r(0) = 1.25$ ,  $r(1) = r(-1) = 0.5$ , and  $r(k) = 0$  for any  $|k| \geq 2$ . (Wiener Filter)

(a) What model fits this process best: AR or MA? Find the corresponding parameters.

(b) Design the Wiener filter when using  $u(n)$  to predict  $u(n+1)$ . Can we do better (in terms of MSE) if we use both  $u(n)$  and  $u(n-1)$  as the input to the Wiener filter? What if using  $u(n)$  and  $u(n-2)$ ?

8. Consider the MIMO (multi-input multi-output) wireless communications system shown in Fig. RII.8. There are two antennas at the transmitter and three antennas at the receiver. Assume the channel gain from the  $i$ -th transmit antenna to the  $j$ -th receive antenna is  $h_{ji}$ . Take a snapshot at time slot  $n$ , the received signal is  $y_j(n) = h_{j1}x_1(n) + h_{j2}x_2(n) + v_j(n)$  where  $v_j(n)$  are white Gaussian noise (zero mean, variance  $N_0$ ) independent of signals. We further assume  $x_1(n)$  and  $x_2(n)$  are uncorrelated, and their power are  $P_1$  and  $P_2$ , respectively. Use  $y_1(n), y_2(n)$  and  $y_3(n)$  as input, find the optimal Wiener filter to estimate  $x_1(n)$  and  $x_2(n)$ . (Wiener Filter)

9. Given an real-valued AR(3) model with parameters  $\Gamma_1 = -4/5$ ,  $\Gamma_2 = 1/9$ ,  $\Gamma_3 = 1/8$ , and  $r(0) = 1$ . Find  $r(1), r(2)$ , and  $r(3)$ . (Levinson-Durbin Recursion)

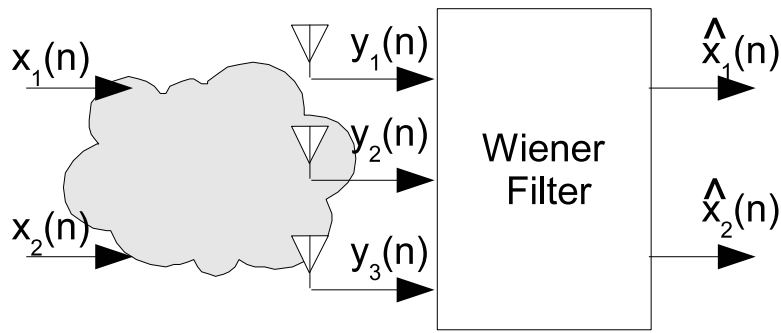


Figure RII.8:

**10.** Consider the MA(1) process  $x(n) = v(n) + bv(n - 1)$  with  $v(n)$  being a zero-mean white sequence with variance 1. If we use  $\Gamma_k$  to represent this system, prove that (Levinson-Durbin Recursion)

$$\Gamma_{m+1} = \frac{\Gamma_m^2}{\Gamma_{m-1}(1 - |\Gamma_m|^2)}.$$

**11.** Given a  $p$ -order AR random process  $\{x(n)\}$ , it can be equivalently represented by any of the three following sets of values: (Levinson-Durbin Recursion)

- $\{r(0), r(1), \dots, r(p)\}$
- $\{a_1, a_2, \dots, a_p\}$  and  $r(0)$
- $\{\Gamma_1, \Gamma_2, \dots, \Gamma_p\}$  and  $r(0)$

(a) If a new random process is defined as  $x'(n) = cx(n)$  where  $c$  is a real-valued constant, what will be the new autocorrelation sequence  $r'(k)$  in terms of  $r(k)$  (for  $k = 1, 2, \dots, p$ )? How about  $a'_k$  and  $\Gamma'_k$ ?

(b) Let a new random process be defined as  $x'(n) = (-1)^n x(n)$ . Prove that  $r'(k) = (-1)^k r(k)$ ,  $a'_k = (-1)^k a_k$  and  $\Gamma'_k = (-1)^k \Gamma_k$ . (Hint: use induction when proving  $\Gamma_k$ , since  $\Gamma_k$  is calculated recursively.)

**12.** Given a lattice predictor that simultaneously generate both forward and backward prediction errors  $f_m(n)$  and  $b_m(n)$  ( $m = 1, 2, \dots, M$ ). (Lattice Structure)

- (a) Find  $E(f_m(n)b_i^*(n))$  for both conditions when  $i \leq m$  and  $i > m$ .
- (b) Find  $E(f_m(n+m)f_i^*(n+i))$  for both conditions when  $i = m$  and  $i < m$ .
- (c) Design a joint process estimation scheme using the forward prediction errors.

(d) If for some reason we can only obtain part of forward prediction error (from order 0 to order  $k$ ) and part of backward prediction error (from order  $k + 1$  to order  $M$ ), i.e., we have

$\{f_0(n), f_1(n), \dots, f_k(n), b_{k+1}(n), b_{k+2}(n), \dots, b_M(n)\}$ . Describe how to use such mixed forward and backward prediction errors to perform joint process estimation.

(Hint: the results from (a) and (b) will be useful for questions (c) and (d). )

**13.** Consider the backward prediction error sequence  $b_0(n), b_1(n), \dots, b_M(n)$  for the observed sequence  $\{u(n)\}$ . (Properties of FLP and BLP Errors)

(a) Define  $\mathbf{b}(n) = [b_0(n), b_1(n), \dots, b_M(n)]^T$ , and  $\mathbf{u}(n) = [u(n), u(n-1), \dots, u(n-M)]^T$ , find  $\mathbf{L}$  in terms of the coefficients of the backward prediction-error filter where  $\mathbf{b}(n) = \mathbf{L}\mathbf{u}(n)$ .

(b) Let the correlation matrix for  $\mathbf{b}(n)$  be  $\mathbf{D}$ , and that for  $\mathbf{u}(n)$  be  $\mathbf{R}$ . Is  $\mathbf{D}$  diagonal? What is relation between  $\mathbf{R}$  and  $\mathbf{D}$ ? Show that a lower triangular matrix  $\mathbf{A}$  exists such that  $\mathbf{R}^{-1} = \mathbf{A}^H \mathbf{A}$ .

(c) Now we are to perform joint estimation of a desired sequence  $\{d(n)\}$  by using either  $\{b_k(n)\}$  or  $\{u(n)\}$ , and their corresponding optimal weight vectors are  $\mathbf{k}$  and  $\mathbf{w}$ , respectively. What is relation between  $\mathbf{k}$  and  $\mathbf{w}$ ?