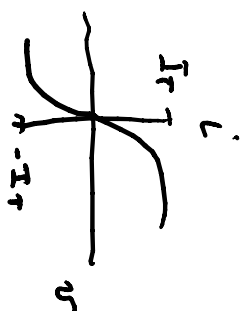


Mistake on  $V$  open book, notes  
but not computer

BST diff pair  $\Rightarrow$  OTA

$$i = \infty I_T \tanh\left(\frac{v}{2V_T}\right)$$

$$\left. \frac{di}{dv} \right|_{v=0} = \infty I_T \left(\frac{1}{2V_T}\right) \left. \frac{d \tanh x}{dx} \right|_{x=0}$$



$$\tanh x = \frac{\sinh x}{\cosh x} = \frac{e^x - e^{-x}}{e^x + e^{-x}}$$

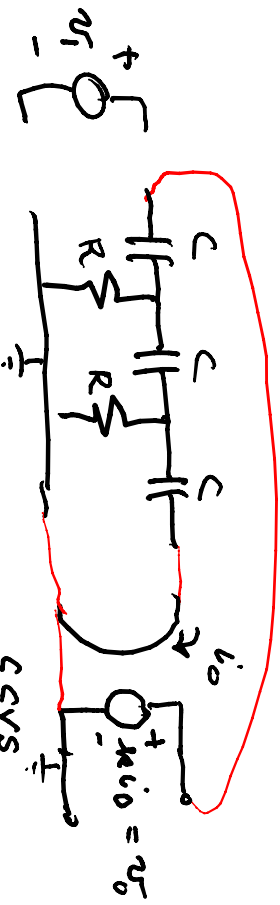
$$\begin{aligned} \frac{d \tanh x}{dx} &= \frac{(e^x + e^{-x}) \cdot 1 - (e^x - e^{-x}) \cdot (e^x - e^{-x})}{(e^x + e^{-x})^2} \\ &= 1 - \tanh^2 x \end{aligned}$$

0 around origin

$$\therefore i = 0 + \infty \frac{I_T}{2V_T} \left(1 - \tanh^2\left(\frac{v}{2V_T}\right)\right) (v - v) + \text{higher order terms}$$

$G_m$  at the bias point  $30$

RC phase shift oscillator p. 1345



CCVS  
(H in Amperes)

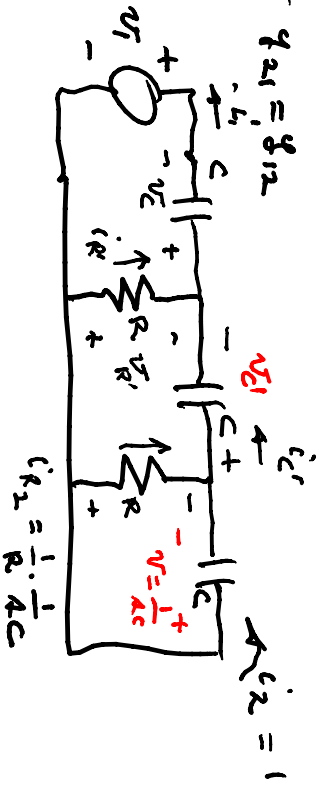
$$Y = \begin{bmatrix} y_{11} & y_{12} \\ y_{21} & y_{22} \end{bmatrix}$$

choose  $i_o = -i_2$  when  $v_2 = 0$

$$\begin{bmatrix} i_1 \\ i_2 \end{bmatrix} = \begin{bmatrix} y_{11} & y_{12} \\ y_{21} & y_{22} \end{bmatrix} \begin{bmatrix} v_1 \\ v_2 \end{bmatrix} \Rightarrow \begin{matrix} v_2 = 0 \\ i_1 \\ i_2 \end{matrix} = \begin{bmatrix} y_{11} \\ y_{21} \end{bmatrix} \begin{bmatrix} v_1 \end{bmatrix}$$

$\therefore$  want  $y_{21} = g_{12}$

analysis



$$i_{R_2} = \frac{1}{R} \cdot \frac{1}{sC}$$

$$i_c' = i_2 + i_{R_2} = 1 + \frac{1}{RCa} ; \quad v_c' = \frac{1}{aC} \cdot \left(1 + \frac{1}{RCa}\right)$$

$$v_{R_1}' = v_c' + v = \frac{1}{aC} + \frac{1}{RC(aC)^2} + \frac{1}{aC} = \frac{2}{aC} + \frac{1}{RC^2a^2}$$

$$i_{R_1}' = \frac{1}{R} v_{R_1}' = \frac{2}{RCa} + \frac{1}{(RCa)^2}$$

$$i_c' = i_{R_1}' + i_c' = \frac{2}{RCa} + \frac{1}{(RCa)^2} + \left(1 + \frac{1}{RCa}\right) = 1 + \frac{3}{RCa} + \frac{1}{(RCa)^2}$$

$$-v_1' = v_c + v_{R_1}' = \frac{1}{aC} \left[1 + \frac{3}{RCa} + \frac{1}{(RCa)^2}\right] + \frac{2}{aC} + \frac{1}{RC^2a^2}$$

$$= \frac{1}{RC^2a^3} + \frac{4}{RC^2a^2} + \frac{3}{aC} = \frac{1}{RC^2a^3} \left[1 + 4RCa + RC^2a^2\right]$$

$$y_{21} = \frac{1}{v_1'} = - \left[ \frac{RC^2a^3}{RC^2a^3 + 4RCa + 1} \right]$$

$$v_{O_0} = R_0 i_{O_0} = R_0 (-i_2) = R_0 \left(\frac{-v_2}{v_1'}\right) \cdot v_1' = R_0 (-y_{21}) v_1'$$

*but when connect output to input  $v_{O_0} = v_1'$*

$$v_{O_0} \approx v_1' = R_0 \left[ \frac{RC^2a^3}{1 + 4RCa + RC^2a^2} \right] v_1'$$

look at  $(1 - R_0 \frac{RC^2a^3}{1 + 4RCa + RC^2a^2}) v_1' = 0$

$$\Rightarrow -k R^2 C^3 s^3 + R^2 C^3 s^2 + 4RCs + 1 = 0 \quad s = \sqrt{-1}$$

Let  $k = j\omega$  for sinusoidal steady state

$\Rightarrow$  oscillations

$$-k R^2 C^3 (-j\omega^3) - R^2 C^3 \omega^2 + 4jRC\omega + 1 = 0$$

$\Rightarrow$  set real & imaginary parts = 0

$$Re: \quad 1 - R^2 C^3 \omega^2 = 0$$

$$Im: \quad k R^2 C^3 \omega^3 + 4RC\omega = 0 \Rightarrow RC + k R^2 C^3 \omega^2 = 0$$

$$\text{oscillates at } \omega_0^2 = 1/R^2 C^2 \Rightarrow \omega_0 = 2\pi f_0 = \frac{1}{RC}$$

$$\text{if we set } k: \quad k = \frac{-RC}{R^2 C^3 \omega^2} = -\frac{R^2 C^2}{R^2 C^3 \omega^2} = -\frac{1}{C}$$

$\therefore$  can make an oscillator with this circuit

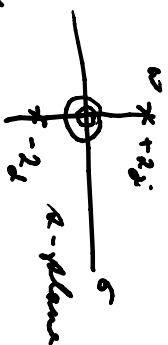
impulse response:  $S(s) = \text{unit impulse}$

$$= \frac{dI(t)}{dt}, \quad I(t) = \text{unit step}$$

$$f[S] = 1$$

$$\text{output (a)} = T(a) = \begin{cases} 1 & t > 0 \\ 0 & t < 0 \end{cases}$$

$$Z_k: T(s) = \frac{s^2}{s^2+4}$$



$\mathcal{L}^{-1}[T(s)] =$  inverse Laplace transform  
use partial fraction

$$T(s) = \frac{k}{s+2j} + \frac{k^*}{s-2j} + 1 \quad k^* = j \rightarrow -j$$

$$= \frac{k s^2}{s^2+4} = \frac{k s - j 2 k}{s^2+4} + \frac{k^* s + j 2 k^*}{s^2+4} + \frac{s^2+4}{s^2+4}$$

$k = k_p + j k_i$

$$= \frac{s^2 + (k+k^*)s + j(2k^* - 2k) + 4}{s^2+4}$$

$$(k+k^*)s = 0, \quad j(2k^* - 2k) + 4 = 0$$

$\Downarrow$

$$k+k^* = k_p = 0 \quad ; \quad j(2k^* - 2k) = -4$$

$$2j(2k_i) = -4 \Rightarrow k_i = \frac{4}{4j} = -j$$

$$T(s) = \frac{s^2}{s^2+4} = \frac{-j}{s+2j} + \frac{j}{s-2j} + 1 = \frac{-js^2 - 2 + js^2 - 2 + s^2 + 4}{s^2+4}$$

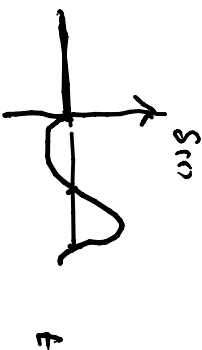
$$\mathcal{L}^{-1}[e^{at} \mathcal{L}(s)] = \frac{1}{s+a} \text{ for } \Re a > -a$$

$$\mathcal{L}^{-1}[\Gamma(s)] = -je^{-j2t} \mathcal{L}^{-1}\{1(s)\} + je^{-(2-j)t} \mathcal{L}^{-1}\{1(s)\} + \delta(t)$$

$$= j \left[ e^{j2t} - e^{-j2t} \right] \mathcal{L}^{-1}\{1(s)\} + \delta(t)$$

$$= -2 \frac{e^{j2t} - e^{-j2t}}{2j} \mathcal{L}^{-1}\{1(s)\} + \delta(t)$$

$$= -2 \sin(2t) \mathcal{L}^{-1}\{1(s)\} + \delta(t)$$



$i_{in} = v_{in}$ ,  $v_{out} = v_{out}$

$$T(s) = \frac{V_{out}(s)}{V_{in}(s)} = \frac{R^2}{s^2 + 4} \implies \text{let } s = \frac{d}{dt}$$

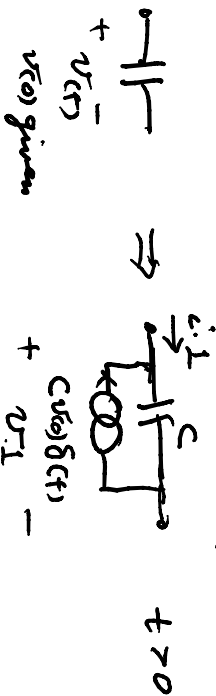
$$= \frac{v_{out}(t)}{v_{in}(t)}$$

$$(s^2 + 4)v_{out}(t) = s^2 v_{in}(t)$$

$$\text{if } v_{in}(t) = \delta(t)$$

This gives  $v_{out} = -2 \sin(2t) \mathcal{L}^{-1}\{1(s)\} + \delta(t)$

Initial conditions on a capacitor



$$c \frac{dV}{dt} = I(t) \quad \Rightarrow \quad \frac{dV(t)}{dt} = I(t) \quad t > 0$$

$$c \frac{d(V(t)I(t))}{dt} = c \frac{dV}{dt} I(t) + c V(t) \frac{dI(t)}{dt} = c \frac{dV}{dt} I(t) + c V(t) \delta(t)$$

$$I(t)I(t) = c \frac{dV(t)}{dt} I(t) - c V(t) \delta(t)$$