

#1. a) $V_{GS} - |V_{TOp}| = V_1 - \frac{1}{2} = \frac{V_2}{2} - \frac{1}{2} = \frac{1}{2}(V_2 - 1)$; $V_{SD} = V_2$

i.e. M_2 is in saturation if $V_2 > \frac{1}{2}(V_2 - 1) \Rightarrow V_2 > -1$, as $V_{DD} > 0$
 then $V_2 > 0 \Rightarrow V_2 > -1 \Rightarrow M_2$ is in saturation

b) $I_1 = \frac{K_{Pp}}{2} \frac{W}{L} (V_1 - |V_{TOp}|)^2 (1 + \lambda_p V_1)$

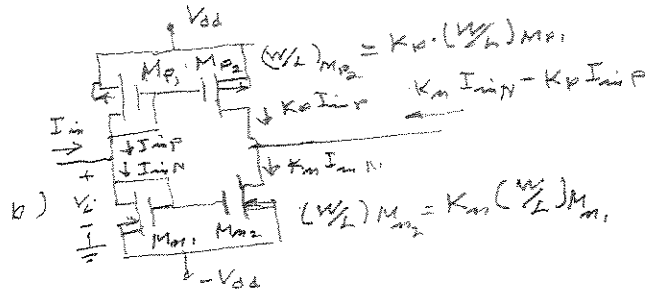
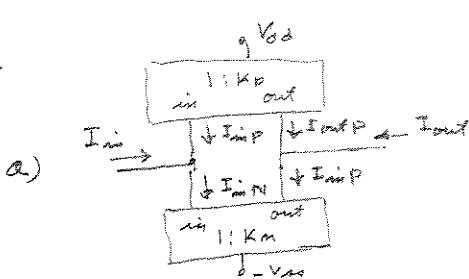
$I_2 = \frac{K_{Pp}}{2} \frac{W}{L} (V_1 - |V_{TOp}|)^2 (1 + \lambda_p V_2) = \frac{K_{Pp}}{2} \frac{W}{L} (V_1 - |V_{TOp}|)^2 (1 + 2\lambda_p V_1)$

$\Rightarrow \frac{I_2}{I_1} = \frac{(1 + 2\lambda_p V_1)}{(1 + \lambda_p V_1)} = \frac{1 + 0.04V_1}{1 + 0.02V_1}$

c) If $V_2 = 4 \Rightarrow V_1 = 2 \Rightarrow I_1 = \frac{10^{-5}}{2} \cdot 2 \left(2 - \frac{1}{2}\right)^2 (1 + 0.02 \times 2) = 10^{-5} \left(\frac{9}{4}\right) (1.04)$
 $\Rightarrow I_1 = 23.4 \mu A$

$I_2 = \frac{(1 + 0.02 \times 4)}{1 + 0.02 \times 2} I_1 = 10^{-5} \left(\frac{9}{4}\right) \cdot (1.08) = 10^{-5} (9 \times 0.270) = 24.3 \mu A$

#2.



c) When $I_{in} = 0 \Rightarrow I_{inP} = \frac{K_{Pp}}{2} \left(\frac{W}{L}\right) (V_{DD} - V_i - |V_{TOp}|)^2 (1 + \lambda_p [V_{DD} - V_i])$

$= I_{inN} = \frac{K_{Pn}}{2} \left(\frac{W}{L}\right) (V_i - (-V_{DD}) - V_{TOm})^2 (1 + \lambda_n [V_i - (-V_{DD})])$

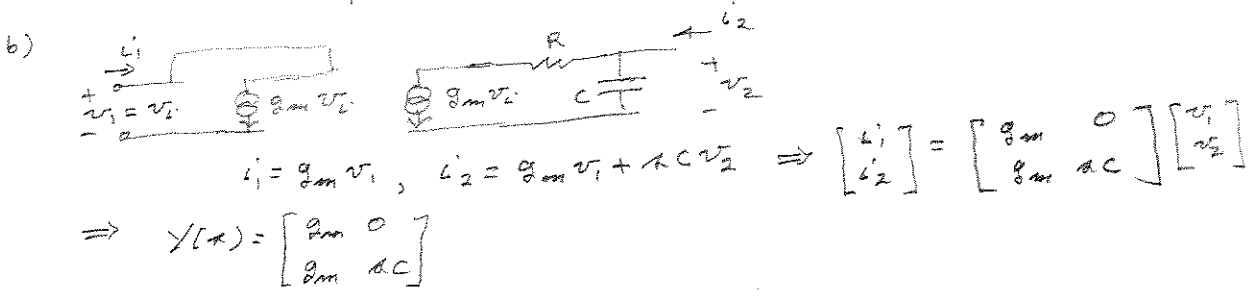
for $V_i = 0 \Rightarrow I_{inP} = I_{inN} = 10^{-5} \left(5 - \frac{1}{2}\right)^2 (1 + 0.02 \times 5) = 10^{-5} \left(\frac{81}{4}\right) (1.1) = 22.275 \times 10^{-6}$
 $= 222.75 \mu A$

d) When $K_m \neq K_p$, the output current has an offset

so that $I_{out} = K_m I_{inN} - K_p I_{inP} = K_m (I_{inN} - I_{inP}) + (K_m I_{inP} - K_p I_{inP})$
 $= K_m I_{in} + \underbrace{[K_m I_{inP} - K_p I_{inP}]}_{\text{output offset}}$

#3, as $\frac{d \tanh x}{dx} \Big|_{x=0} = 1$ and $g_m = \frac{d I_{out}}{d V_i} \Big|_{V_i=0}$

a) $\Rightarrow g_m = \alpha I_T \cdot \frac{1}{2V_T}$

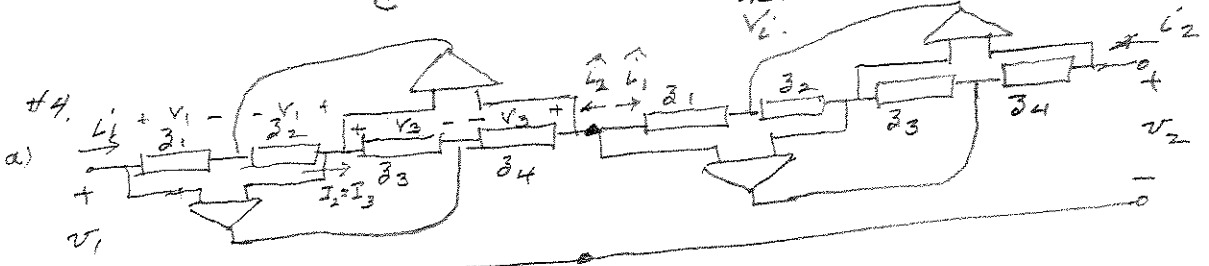


c) as $i_2 \equiv 0, -g_m v_1 = sC v_2 \Rightarrow v_2 = -\frac{g_m}{sC} v_1$

$\therefore v_2(t) = -\frac{g_m}{C} \int_{-\infty}^t v_1 \cdot 1(t-\tau) d\tau = -\frac{g_m}{C} v_1 \left(\int_0^t 1 \cdot d\tau \right) 1(t)$

$= -\frac{g_m}{C} v_1 \cdot t \cdot 1(t)$

Note: The output increases even though the OTA does not saturate for a mod v_i .



b) By virtual ties of op-amp inputs, $v_2 = v_1$

over currents $i_1 = v_1/\beta_1 \Rightarrow I_2 = -v_1/\beta_2 = -\frac{\beta_1}{\beta_2} i_1 = I_3 = \frac{v_1}{\beta_3} \Rightarrow v_1 = -\frac{\beta_1 \beta_3}{\beta_2} i_1 \Rightarrow i_2 = \frac{v_1}{\beta_4} = -\frac{\beta_1 \beta_3}{\beta_2 \beta_4} i_1$

$\Rightarrow i_2 = -\left(\frac{\beta_1 \beta_3}{\beta_2 \beta_4} \right) i_1$

c) $\beta_1 = \frac{1}{sC_1}, \beta_3 = \frac{1}{sC_3}, \beta_2 = R_2, \beta_4 = R_4 \Rightarrow i_2 = -\left(\frac{1}{s^2 R_1 C_1 R_2 C_2} \right) i_1$

$\Rightarrow \frac{i_2}{i_1} = \frac{1}{(R_1 C_1 R_2 C_2)^2} \cdot \frac{-1}{s^4}$

gives i_2 as 4th integral of i_1 with a negative coefficient

i May be useful where frequency dependent current mirrors are of interest.