

general 1-root $g(s) = \frac{m(s)}{d(s)}$ design (1st order non-zero behavior at $\omega = 0$, = constant & poles, by

example

$$g(s) = \frac{m_1 s + m_0}{s^2 + a_1 s + a_0}$$

use companion matrix form long division)

$$i(s) = g(s)u(s) = (m_1 s + m_0)g_1(s)u(s), \quad g_1(s) = \frac{1}{s^2 + a_1 s + a_0} \Rightarrow a_1 i' + a_0 i = u$$

$$g(s) = (m_1 s + m_0) \cdot g_1(s)$$

set up state eqs: $x_1 = i, x_2 = i', x_3 = \dot{x}_2 = i'', j \neq a = g/s$

$$s \begin{bmatrix} x_1 \\ x_2 \end{bmatrix} = \begin{bmatrix} x_1 \\ -x_2 \end{bmatrix} = \begin{bmatrix} x_2 \\ -a_0 \end{bmatrix} = \begin{bmatrix} 0 & 1 \\ -a_0 & -a_1 \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \end{bmatrix} + \begin{bmatrix} 0 \\ 1 \end{bmatrix} u$$

but want i as output

$$i = m_1 x_1 + m_0 i' = m_1 x_2 + m_0 x_1$$

$$i = \begin{bmatrix} m_0 & m_1 \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \end{bmatrix} \Rightarrow x = Ax + Bv \quad i = Cx$$

C, X, \leq capacitors current in capacitor of capacitance C , if $x_1 =$ voltage

$\Rightarrow \dot{x} =$ capacitor currents $= A x + B v \Rightarrow$

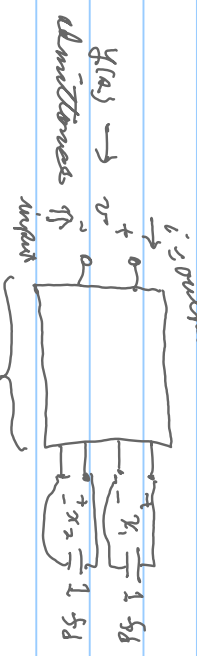
$$\begin{bmatrix} -B & -A \\ D & C \end{bmatrix} \begin{bmatrix} v \\ x \end{bmatrix} = \begin{bmatrix} \dot{x} \\ i \end{bmatrix}$$

output current = $[C \ x] + D v$

$$\begin{bmatrix} i \\ x \end{bmatrix} = \begin{bmatrix} 0 & C \\ -B & -A \end{bmatrix} \begin{bmatrix} v \\ x \end{bmatrix}$$

constant admittance

$$= Y_C$$



Numerical example

$$Y(s) = \frac{m_0 s + m_1}{s^2 + a_0 s + a_1}$$

$$\begin{bmatrix} i \\ x_1 \\ x_2 \end{bmatrix} = \begin{bmatrix} 0 & m_0 & m_1 \\ 0 & 0 & -1 \\ a_0 & a_1 & 0 \end{bmatrix} \begin{bmatrix} v \\ x_1 \\ x_2 \end{bmatrix}$$

voltages on unit capacitors

$$Y_C = \begin{bmatrix} 0 & m_0 & m_1 \\ 0 & 0 & -1 \\ -1 & a_0 & a_1 \end{bmatrix}$$

$$y(s) = \begin{bmatrix} m_0 & m_1 \end{bmatrix} \begin{bmatrix} s-0 & \dots & 1 \\ a_0 & a_1 & \dots & (-a_1) \end{bmatrix}^{-1} \begin{bmatrix} 0 \\ 1 \end{bmatrix}$$

or $y(s) = \text{admittance realer}$ no $y(s) = y(s)^T$

$$= C (sI_2 - A)^{-1} B = (C (sI_2 - A)^{-1} B)^T = B^T (sI_2 - A^T)^{-1} C^T$$

$$T \dot{x}' = T A T^{-1} x + T B u$$

$$y' = C^T T^{-1} x$$

$T = \text{non singular constant}$

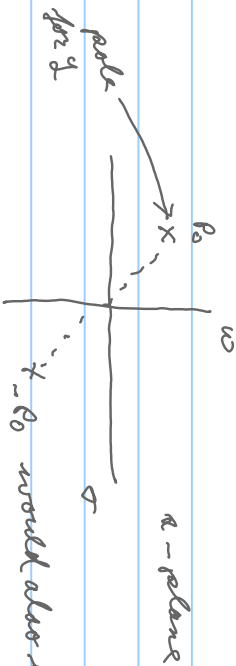
let $\hat{x} = T^{-1} x$ then $\dot{\hat{x}} = \hat{A} \hat{x} + \hat{B} u$ where $\hat{A} = T A T^{-1}$
 $y = \hat{C} \hat{x}$ $\hat{B} = T B$, $\hat{C} = C^T T^{-1}$

Thus: if $y(s)$ is PR then a T exists for \hat{y}_c to also be PR constant \Rightarrow passive

thus $\hat{y}_c = \begin{bmatrix} 1 & 0 \\ 0 & T \end{bmatrix} \begin{bmatrix} 1 & 0 \\ 0 & T^{-1} \end{bmatrix} = \begin{bmatrix} 1 & 0 \\ -\hat{B} & \hat{A} \end{bmatrix} = \begin{bmatrix} D & C^T T^{-1} \\ T B & -T A T^{-1} \end{bmatrix}$

downward

$$y(s) = -y(-s)$$



\hat{y}_c would also be in $y(s)$ but if PR cannot be

∴ all poles are on jω axis: $Y(s) = \frac{k_R}{s^2 + \omega_0^2}$ a pole @ $s = \pm j\omega_0$

but $\left(\frac{k_R}{s^2 + \omega_0^2}\right)^2$ is not PR $Y(s) = \frac{1}{s}$ is PR but $\frac{1}{s^2}$ is not PR

∴ Re is positive & real

if $Y(s)$ is PR then at is $\frac{1}{Y(s)} = Z(s)$

∴ if $Y(s)$ is lossless then at is $Z(s) = 1/Y(s)$

∴ make a partial fraction expansion of $Y(s) = \frac{k_0}{s} + k_0 a + \sum_{i=1}^m \left(\frac{k_i}{s - j\omega_i} + \frac{k_i^*}{s + j\omega_i} \right)$ if lossless

$Y(s) = \frac{k_0}{s} + k_0 a + \sum_{i=1}^m \frac{2k_i a}{s^2 + \omega_i^2}$; $k_i \geq 0$ $k_i = \text{real for PR } Y(s)$
 + means parallel



⇒ 2nd direct form of synthesis

