

Advantages in s -domain use bilateral Laplace transform

$$F(s) \approx \int_{-\infty}^{\infty} f(t) e^{-st} dt \quad \text{where } s = \sigma + j\omega, \quad j = \sqrt{-1}$$

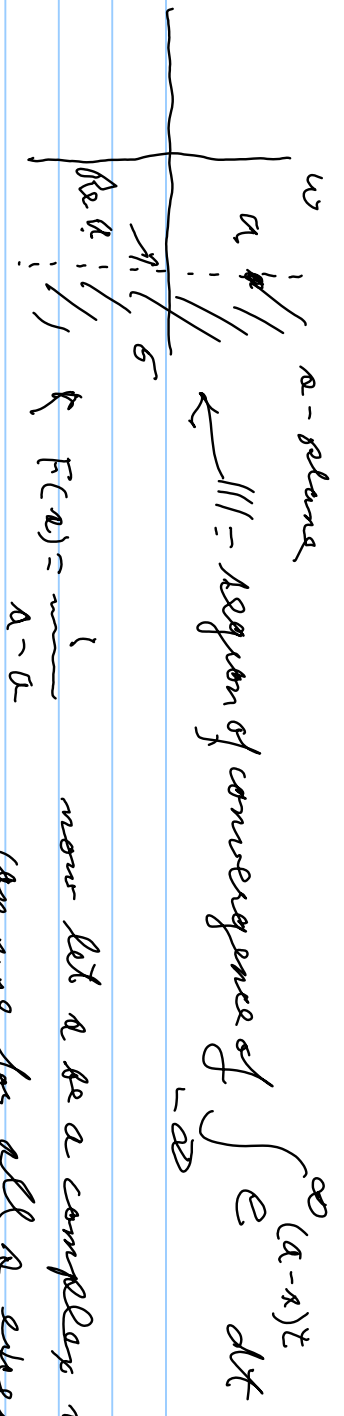
Ex:

$$e^{at} \mathcal{L}\{1(t)\}, \quad 1(t) = \begin{cases} 1 & t > 0 \\ 0 & t < 0 \end{cases}$$

$$\text{where } s = j\omega \quad F(s) = \int_{-\infty}^{\infty} f(t) e^{-st} dt$$

$$\int_{-\infty}^{\infty} e^{at} \mathcal{L}\{1(t)\} e^{-st} dt = \int_0^{\infty} e^{(a-s)t} dt = \frac{1}{(a-s)} \cdot e^{(a-s)t} \Big|_{t=0}^{t=\infty}$$

$$\approx \frac{1}{(a-s)} + \frac{1}{(a-s)} e^{(a-s)\cdot\infty} \quad \text{need } e^{(a-s)\cdot\infty} = 0 \Rightarrow \operatorname{Re}(a-s) < 0 \Rightarrow \operatorname{Re} a < \operatorname{Re} s$$



now let a be a complex variable
can we do for all a except $a = a$

look at $e^{at} \mathcal{L}\{1(t)\}$, $\int_{-\infty}^{\infty} -e^{at} \mathcal{L}\{1(t)\} e^{-st} dt = \int_{-\infty}^0 -e^{(a-s)t} dt = \int_0^{\infty} -e^{(a-s)t} dt$

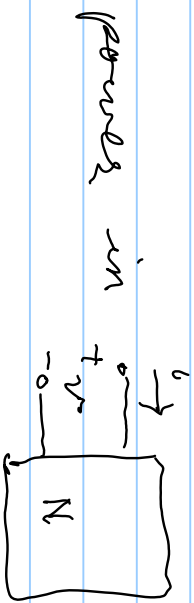
$= \frac{-1}{(a-s)} - \left(\frac{-1}{(a-s)} e^{(a-s)(-\infty)} \right)$ goes to zero if $\text{Re}(a-s) > 0 \Rightarrow \text{Re } a > \text{Re } s$

$F(s) = \frac{1}{s-a}$

\therefore find the region of convergence to identify



1. $\frac{1}{s-a}$ with a Laplace transform (with a in the positive half plane)



$$P_{N_1}(t) = \mathcal{L}^{-1}\{f\}$$

$$f^* = j \rightarrow -j$$

$$= \frac{(\mathcal{L}^{-1}\{f\})^* + (\mathcal{L}^{-1}\{f\})}{2} = \frac{\mathcal{L}^{-1}\{f\} + \mathcal{L}^{-1}\{f^*\}}{2}$$

↑
or $P_{N_1}(t)$ is real

Parseval's theorem

$$\int_{-\infty}^{\infty} |f(j\omega)|^2 G(j\omega) d\omega = \int_{-\infty}^{\infty} f^*(t) g(t) dt$$

if $f(t) = v(t)$, $g(t) = v(t)$ then

$$\int_{-\infty}^{\infty} |v(t)|^2 dt = \int_{-\infty}^{\infty} P(t) dt$$

if N is passive then $\int_{-\infty}^{\infty} P(t) dt \geq 0$

Then
$$\int_{-\infty}^{\infty} V^{Tx} Y(j\omega) I(j\omega) d\frac{\omega}{2\pi} = \int_{-\infty}^{\infty} V^{Tx} (Y(j\omega) Y(j\omega)) d\frac{\omega}{2\pi} \geq 0$$
 for any V if passive

$$= \int_{-\infty}^{\infty} V^{Tx} \left[\frac{Y(j\omega) + Y^{Tx}(j\omega)}{2} \right] V d\frac{\omega}{2\pi} \geq 0$$

for all V if passive ($V =$ fixed complex numbers)

$\therefore \frac{Y(j\omega) + Y^{Tx}(j\omega)}{2}$ is a positive semidefinite Hermitian form

Ex: If N is an inductor, $Y(s) = \frac{1}{sL}$, $Y^{*T}(j\omega) = \frac{1}{L(-j\omega)}$, $Y = \frac{1}{L(-j\omega)}$, $Y^{*T} = \frac{-1}{L(j\omega)}$

$$\frac{Y(j\omega) + Y^{*T}(j\omega)}{2} = 0 \Rightarrow \mathcal{E}(\omega) = 0 \leftarrow \text{Resistor}$$

\therefore for lossless: $Y(j\omega) + Y^*(j\omega) = 0_n \Rightarrow Y^*(j\omega) = -Y(j\omega)$

if rational $Y^*(j\omega) = Y(-j\omega) \Rightarrow Y(\alpha)$ poles for $Y^*(j\omega)$

\therefore for rational $Y(j\omega) + Y^*(j\omega) = Y(j\omega) + Y^T(-j\omega)$

2 extend to s -plane by $\omega = \alpha_j \Rightarrow Y(\alpha) + Y^T(-\alpha) = 0_n$ for all α except

the finite # of poles

for loss admittance: $Y(\alpha) = -Y^T(-\alpha)$

if passive: no poles in $\text{Re } \alpha > 0$ (as stable)

& if real coefficients are all real: $Y(\alpha)$ is real when α is real, in $\text{Re } \alpha > 0$

if passive: $Y(\alpha) + Y^*(\alpha)$ is positive semi-definite in $\sigma > 0$

such are called positive real, & PR if rational \Leftrightarrow can synthesize
with passive components