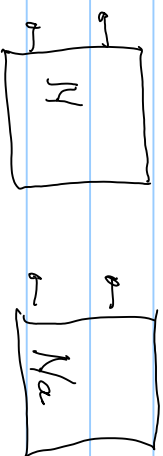


EE618
09/27/16b

adjoint



$$C_a^T K = 0_{b \times b} \Rightarrow X_a = Y^T$$

$$C_a^T K = 0_{b \times b}$$

$$S = (1_b \quad Y)^{-1} (1_b \quad Y)$$

$$S_a = (1_b \quad Y_a)^{-1} (1_b \quad Y_a)$$

$$= (1_b \quad Y^T)^{-1} (1_b \quad Y^T)$$

$$S = (1_b \quad Y^T) (1_b \quad Y)^{-T}$$

$$= (1_b \quad Y^T)^{-1} (1_b \quad Y^T)$$

$$\Leftrightarrow S_a = S^T$$

$$\mathcal{L}[f(t)] = \int_{-\infty}^{\infty} f(t) e^{-\sigma t} dt = \int_{-\infty}^{\infty} f(t) e^{-\sigma t - j\omega t} dt = \int_{-\infty}^{\infty} [f(t) e^{-\sigma t}] e^{-j\omega t} dt$$


$= \int_{-\infty}^{\infty} [f(t)e^{-\sigma t}]$ if $\sigma > 0 \Rightarrow$ if $f(t) = 0$ for $t < 0$ & square integrable (= finite energy)

then $f(t)e^{-\sigma t}$ is also finite energy (energy)

If a circuit is causal [$v'(t) = 0$ for $t < 0$ if $v(t)$ is 0 for $t < 0$]

\Rightarrow $S(t)$ is zero for $t < 0$; physically $S(t)$ is $v'(t)$ for $v'(t)$ being an impulse

$$v'(t) = \int_{-\infty}^{\infty} S(\tau) \delta(t-\tau) d\tau \quad \text{for } v'(t) = \delta(t) = \text{unit impulse}$$

$$\Rightarrow S(t) = \int_{-\infty}^{\infty} \delta(t-\tau) d\tau$$


If $f(t) = v'(t)$ is square integrable & 0 for $t < 0$ then for causality we have $v'(t)$ & $v'(t)e^{-\sigma t}$ also if $\sigma > 0$

Thus $\|v^i\|^2 - \|v^n\|^2 \geq 0$ if passive, here $\|S\|^2 = \int_{-\infty}^{\infty} f^*(t)f(t) dt$

$(s=j\omega \rightarrow -j)$ $1_m \sim S(j\omega)S^*(j\omega)$ is positive semidefinite

then also if causal $\frac{1}{1_m} \sim S^T(\sigma-j\omega)S(\sigma+j\omega)$ for $\sigma > 0$ with $\sigma > 0$

for a 1-root, $m=1$, $S(a)$ is the reflection coefficient

here $S^T(a) = S^*(a^*)$ for $\sigma > 0$ & real circuit

Ex: $\int_C S(a) = \frac{1-ac}{1+a}$ then $S^*(a) = \frac{1-a^*c^*}{1+a^*c^*} = S^*(a^*)$ for real C

If $S(a)$ is rational & comes from a passive circuit

then it is called BR (= bounded real & rational)

Bounded real conditions (BR if rational)

1. $S^*(a) = S(a^*)$ in $\sigma > 0$ (the RHP) [real circuit]
2. $S(a)$ is analytic in $\sigma > 0$ [stable; no poles in $\sigma > 0$ of BR]
3. $1_m \sim S^T(a)S(a)$ is positive semidefinite [passive]
 \Rightarrow if $m=1$, $|1-S(a)|^2 > 0$ in $\sigma > 0$

Known result: If BR then can synthesize by a passive circuit.

If lossless $\mathbf{1}_n - S^{T*}(j\omega)S(j\omega) = \mathbf{0}_n$

$$\Rightarrow \mathbf{1}_n - S^{T*}(-\alpha)S(\alpha) = \mathbf{0}_n \Rightarrow S(\alpha) = S^{T*}(-\alpha)$$

$$\text{E.g.: } \frac{1}{1-c} \Rightarrow S(\alpha) = \frac{1-\alpha c}{1+\alpha c}, \quad S(-\alpha) = \frac{1-(-\alpha)c}{1+(-\alpha)c} = \frac{1+\alpha c}{1-\alpha c} = \frac{1}{S(\alpha)}$$

lossless synthesis via $y(\alpha)$ if $S(\alpha)$ is bounded real

$$\Rightarrow y(\alpha) + y^{T*}(\alpha) \geq 0 \quad \text{in } \sigma > 0$$

$\Rightarrow y(\alpha)$ are positive-real $\stackrel{\text{R=jo}}$ ≥ 0 (average power in sinusoidal steady state)

if from a passive network (if rational \Rightarrow PR)

PR also means physically realizable.

of lossless $Y(a) + Y^T(a) = O_m$ on $a = j\omega$

$\Rightarrow Y(a) + Y^T(-a) = O_m$ in $a = \sigma + j\omega$, $\sigma > 0$

Ex: $\int_C Y(a) = CA$ $Y(-a) = -aC$ $\Rightarrow Y(a) + Y(-a) = 0$
PR not PR

Positive real conditions

1. $Y^*(a) = Y(a^*)$ in $\sigma > 0$ [real circuit]
2. $Y(a)$ is analytic in $\sigma > 0$ [stable circuit]
3. $Y(a) + Y^*(a)$ is positive semidefinite in $\sigma > 0$ [passivity]

of rational \Rightarrow PR, of lossless LPR

\Rightarrow synthesis of LPR scalars

properties a) all singularities are poles

b) all poles (zeros) are on $j\omega$ axis

c) & simple with positive residues

d) $Y(s) \geq 0$

all points where $g(s) \rightarrow \infty$ are poles if rational

$g(s) |_{p_1}$ is singular $g(s) = g(s)$ (zeros)
 $g(s) |_{-p_1}$ is singular

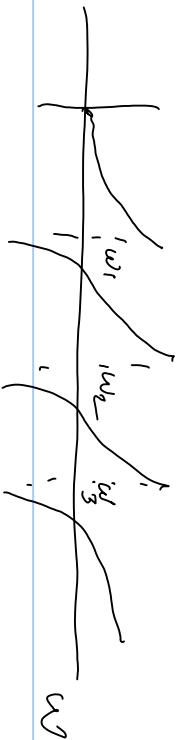
σ (pole here if $p_1 = \sigma_1 + j\omega$, then $\sigma_1 < 0$
 \uparrow can't be if lossless

\downarrow
 all poles are on $j\omega$ axis

$g(s) \rightarrow \frac{k_1 a}{a^2 + p_1^2}$ & not $\frac{k_1 a}{(a^2 + p_1^2)^2}$, $-k_1 a > 1$

$$g(s) = \sum_{l=1}^p \frac{2k_l a}{a^2 + \omega_l^2} + \frac{k_0}{a} + k_\infty a \quad ; \quad k_l > 0$$

den $g(s; \omega)$



$$\frac{d g(s; \omega)}{d \omega} = \frac{d}{d \omega} \left[\sum_{i=1}^D \frac{2 s^i k_i \omega}{(-\omega^2 + \omega_i^2)} + \frac{k_0}{j \omega} + k_\alpha j \omega \right]$$

Homework 3 due Tu 10/04/16