

EE610

10/22/15

3-part
 simulator
 $S = \begin{bmatrix} 0 & 0 & 1 \\ 1 & 0 & 0 \\ 0 & 1 & 0 \end{bmatrix}$

$$A_{NS} = B_i; \quad y_{NS} = i$$

$$(B+A)^{-1}(B-A) = S = \begin{pmatrix} 1+\gamma \\ 3 \end{pmatrix}^{-1} \begin{pmatrix} 1-\gamma \\ 3 \end{pmatrix}$$

$$(1_3 + \gamma)S = S + \gamma S = 1_3 - \gamma$$

$$S^{-1} 1_3 = -\gamma S^{-1} \gamma = -\gamma (S + 1_3)$$

$$(1_3 - S)(1_3 + S)^{-1} = \gamma$$

$$(1_3 + S)^{-1} = \begin{bmatrix} 1 & 0 & 1 \\ 1 & 1 & 0 \\ 0 & 1 & 1 \end{bmatrix}^{-1} = \frac{1}{2} \begin{bmatrix} 1 & 1 & -1 \\ -1 & 1 & 1 \\ 1 & -1 & 1 \end{bmatrix}$$

det = 1. det $\begin{bmatrix} 1 & 0 \\ 1 & 1 \end{bmatrix} = 1$ det $\begin{bmatrix} 0 & 1 \\ 1 & 1 \end{bmatrix} = 1$

$$(1_3 - S) = \begin{bmatrix} 1 & 0 & -1 \\ -1 & 1 & 0 \\ 0 & -1 & 1 \end{bmatrix} = 1, 1, -1 (-1) = 2$$

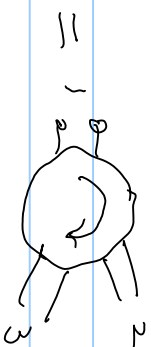
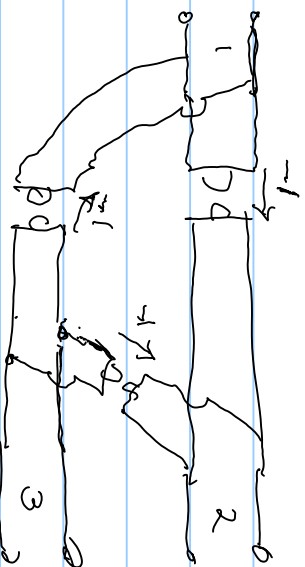
$$Y = (1_3 + S)^{-1} (1_3 - S) = \frac{1}{2} \begin{bmatrix} 0 & 2 & -2 \\ -2 & 0 & 2 \\ 2 & -2 & 0 \end{bmatrix}$$

$$= \begin{bmatrix} 0 & 1 & -1 \\ -1 & 0 & 1 \\ 1 & -1 & 0 \end{bmatrix}$$

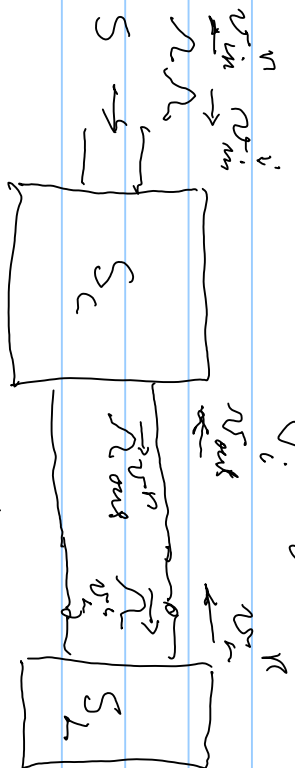
$$y_{12} = 1, y_{21} = -1$$

$$x = \begin{bmatrix} 0 & -g \\ +g & 0 \end{bmatrix} \quad g = -1$$

Can make with 3 gyrostators



Loaded restoring system



$v_k^n = v_{out}^i$ when normalizing
 $v_L^i = v_{out}^n$ for the same Z_0

$$S_C = \begin{bmatrix} S_{11} & S_{12} \\ S_{21} & S_{22} \end{bmatrix} = \begin{bmatrix} 0 & 0 & 1 \\ 1 & 0 & 0 \\ 0 & 1 & 0 \end{bmatrix}$$

$$S_L = \begin{bmatrix} S_a & 0 \\ 0 & S_b \end{bmatrix} \Rightarrow S = S_b \cdot S_C$$

$$S_C \begin{bmatrix} v_{in}^i \\ v_{out}^i \end{bmatrix} = \begin{bmatrix} v_{in}^n \\ v_{out}^n \end{bmatrix}$$

$$\begin{aligned} v_{in}^n &= S_{11} v_{in}^i + S_{12} v_{out}^i \\ v_{out}^n &= S_{21} v_{in}^i + S_{22} v_{out}^i \\ v_L^n &= v_{out}^i = S_{21} v_L^i = S_{21} v_{out}^n \end{aligned}$$

$$S_L v_{out}^n = v_{out}^i = S_L S_{21} v_{in}^i + S_L S_{22} v_{out}^i$$

$$\Rightarrow (1_L - S_L S_{22}) v_{out}^i = S_L S_{21} v_{in}^i$$

$$v_{out}^i = (1_L - S_L S_{22})^{-1} S_L S_{21} v_{in}^i$$

$$S v_{in}^i = \begin{bmatrix} S_{11} + S_{12} (1_L - S_L S_{22})^{-1} S_L S_{21} \\ S_{21} \end{bmatrix} v_{in}^i$$

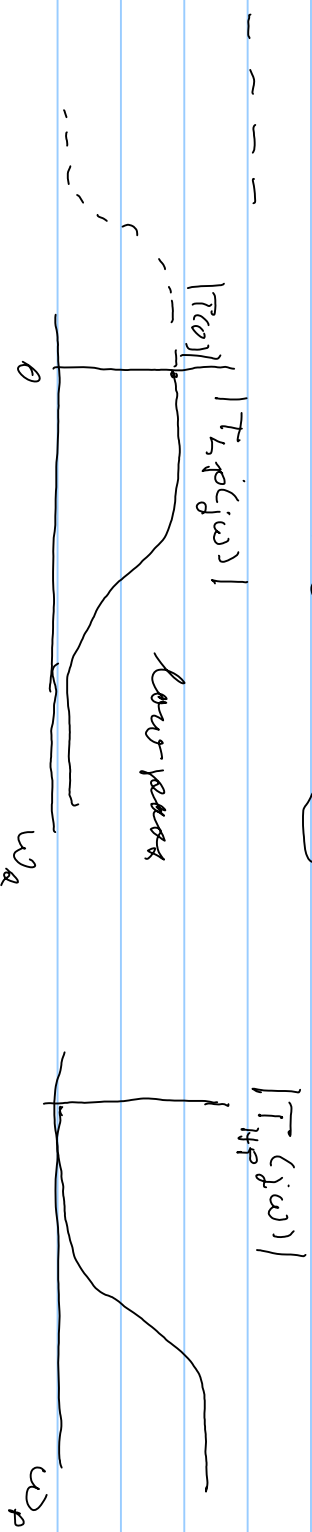
$$S = S_{11} + S_{12} (1_L - S_L S_{22})^{-1} S_L S_{21}$$

Ex

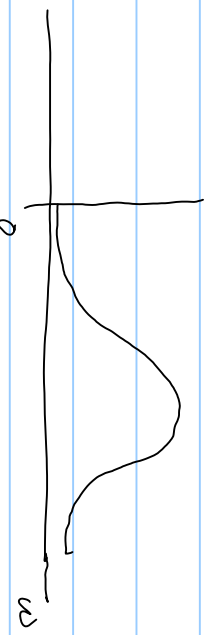
$$1_L - S_L S_{22} = \begin{bmatrix} 1 & 0 \\ 0 & 1 \end{bmatrix} - \begin{bmatrix} S_{21} & 0 \\ 0 & S_{22} \end{bmatrix} \begin{bmatrix} 0 & 0 \\ 1 & 0 \end{bmatrix} = \begin{bmatrix} 1 & 0 \\ -S_{21} & 1 \end{bmatrix} \Rightarrow (1_L - S_L S_{22})^{-1} = \begin{bmatrix} 1 & 0 \\ S_{21} & 1 \end{bmatrix}$$

$$S_k S_{k1} = \begin{bmatrix} S_a & 0 \\ 0 & S_b \end{bmatrix} \begin{bmatrix} 1 \\ 0 \end{bmatrix} = \begin{bmatrix} S_a \\ 0 \end{bmatrix},$$

$$S = S_{11} + S_{12} (I_2 - S_{11} S_{22})^{-1} S_{12} S_{21} = \begin{bmatrix} 0 \\ 0 \end{bmatrix} + \begin{bmatrix} 0 & 1 \\ 0 & 1 \end{bmatrix} \begin{bmatrix} 1 & 0 \\ 0 & 1 \end{bmatrix} \begin{bmatrix} S_a \\ 0 \end{bmatrix} = \begin{bmatrix} S_a \\ 0 \end{bmatrix} = S_b \cdot S_a$$



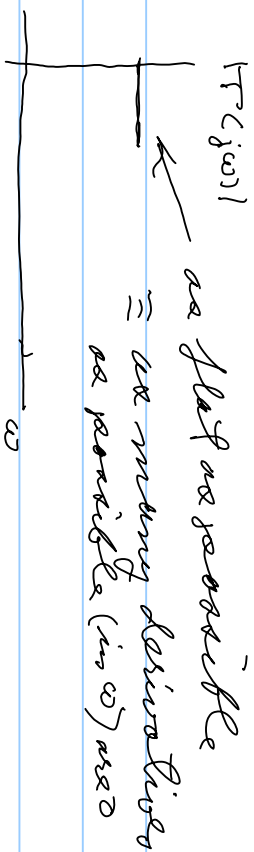
\Rightarrow max $Q = 1/p$
 (poles are PR properties)



$$p/\omega_0 + \frac{\omega_0}{p} = Q$$

Approximately flat LP filter

$$T(\omega) = \frac{K}{a^n + d_{n-1}a^{n-1} + \dots + d_1a + d_0}$$



$$T(j\omega) \cdot T^*(j\omega) = |T(j\omega)|^2 = T(j\omega)T(-j\omega)$$

as real coefficients at all ω are on a

$$\frac{d|T(j\omega)|^2}{d\omega} = 2|T(j\omega)| \cdot \frac{d|T(j\omega)|}{d\omega} \quad \text{as for } \frac{d|T(j\omega)|}{d\omega} = 0 \text{ then}$$

$$\text{max } \frac{d|T(j\omega)|^2}{d\omega} = 0$$

$$|T(j\omega)| = \frac{|K|}{|D(j\omega)|} \quad ; \quad \frac{d|T(j\omega)|}{d\omega} = \frac{-|K|}{|D(j\omega)|^2} \frac{d|D(j\omega)|}{d\omega}$$

$$\therefore \frac{d|T(j\omega)|}{d\omega} = 0 \Rightarrow \frac{d|D(j\omega)|}{d\omega} = 0 \Rightarrow \frac{d|D(j\omega)|^2}{d\omega} = 0$$

also know that $|D(j\omega)| \neq 0$ for $-\infty < \omega < \infty$

$$D(\omega) = a^n + d_{n-1}a^{n-1} + \dots + d_1a + d_0$$

$$D(j\omega) = j^n \omega^n + j^{n-1} d_{n-1} \omega^{n-1} + \dots + j d_1 \omega + d_0$$

$$\begin{aligned}
 D(g^i w) \cdot D(-g^i w) &= d_0^2 + \alpha_1 w^2 + \alpha_2 w^4 + \dots + \alpha_n w^{2n-1} \\
 &= d_0^2 + \frac{d|D(g^i w)|^2}{d\omega^2} \omega^2 + \dots + \frac{d|D(g^i w)|^2}{d\omega^{2n-1}} \omega^{2n} + \alpha_n w^{2n}
 \end{aligned}$$

$\begin{matrix} 1 & 1 \\ 0 & 0 \end{matrix}$

or many factors

$$= d_0^2 + \alpha_{2n} \omega^{2n} = d_0^2 + (g^i)^{2n} \omega^{2n} = d_0^2 + (-1)^n \omega^{2n}$$

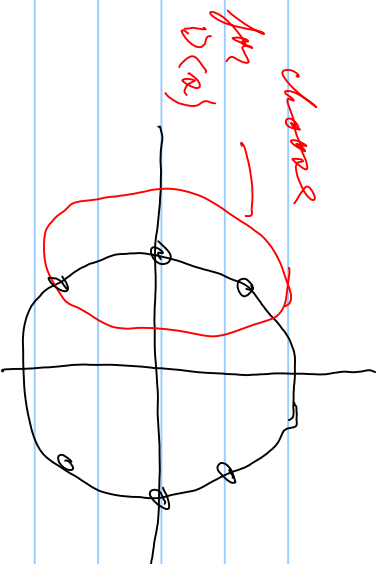
$$\omega = \alpha/g \quad |D(g^i w)|^2 \Big|_{\omega=\alpha/g} = d_0^2 + (-1)^n \left(\frac{\alpha}{g}\right)^{2n} = d_0^2 + \alpha^{2n}$$

normalizing for $d_0 = 1$ $P(\alpha) = \alpha^{2n} + 1 = D(\alpha)D(-\alpha)$

deviser to factors $P(\alpha)$ into $D(\alpha)D(-\alpha) \Rightarrow$ devisor the zeros of $P(\alpha)$ they are $2n$ th roots of $-1 \Rightarrow P(\alpha) = 0 \Rightarrow \alpha^{2n} = -1$

$$-1 = e^{j\pi} = e^{j(\pi + 2k\pi)} = \alpha^{2n} \quad k = 0, \pm 1, \dots$$

nth root is $\alpha_m = e^{j(\pi + 2m\pi)/2n}$



s -plane all roots are on the unit circle

for $D(s)$, as desired $T(s) = \frac{k}{D(s)}$ stable
we want left half plane roots