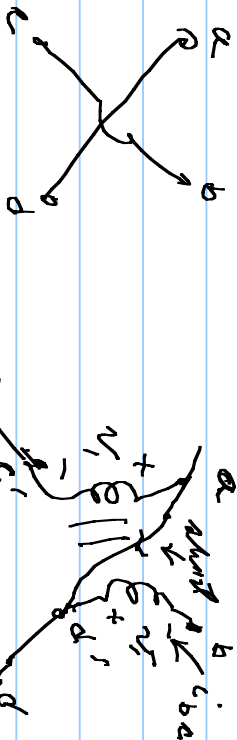


EE 610

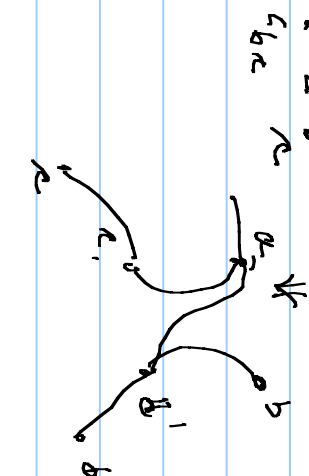
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How to get a planar graph:



$$v_{c'} = v_b \text{ as } v_1 + (1 - v_1) = 0$$

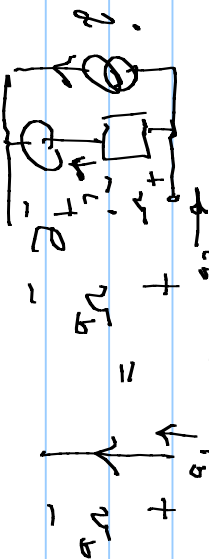
current in a goes to d'



no crossing lead

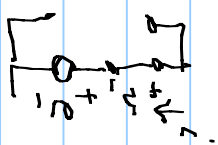
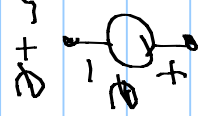
⇒ always a graph with no crossing leads

What to do if a branch is



$$v_b = v + e$$

$$i_b = i + j$$



$$v = 0$$

$$Av = Bc'$$

$$[A][v] = [0][c']$$

$$v = \begin{bmatrix} v_1 \\ v_2 \\ \vdots \\ v_n \end{bmatrix} \quad v_b = \begin{bmatrix} v_{1b} \\ v_{2b} \\ \vdots \\ v_{nb} \end{bmatrix} = \begin{bmatrix} v_1 \\ v_2 \\ \vdots \\ v_n \end{bmatrix} \begin{bmatrix} v_b \\ b \end{bmatrix}$$

if linear then
 $v_b = e^T v$
 $v_b = \sigma^T v$
 $A v = B v$

$$v = v_b - e \quad ; \quad A v = A v_b - A e = B v = B v_b - B e$$

$$v = v_b - e \quad ; \quad A v_b - B v_b = A e - B e$$

$$A e^T v - B \sigma^T v = A e - B e$$

$$\begin{bmatrix} A e^T \\ \vdots \\ -B \sigma^T \end{bmatrix} \begin{bmatrix} v \\ \vdots \\ v \end{bmatrix} = A e - B e \Rightarrow a \text{ b-vector}$$

a b-vector

adding matrices

$v^i \Rightarrow$ independent voltages, $v^i =$ selected voltages

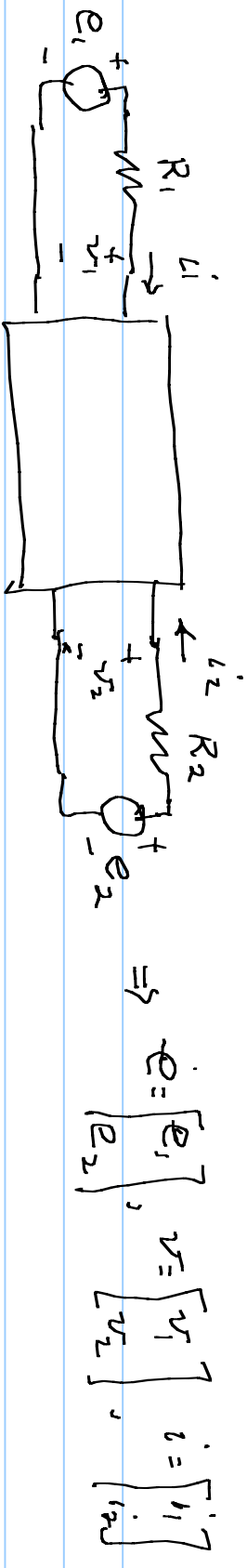
$$2 v^i = v + R i \quad ; \quad 2 v^i = v - R i$$

$$2 v^i = 2 v^i + 2 v^i \Rightarrow v = v^i + v^i$$

$$2 R i = 2 v^i - 2 v^i \Rightarrow R i = v^i - v^i$$

$$i = G(v^i - v^i)$$

$R \sim Z_0$ of a
Transmission
line



$$e_1 = v_1 + R_1 i_1$$

$$e_2 = v_2 + R_2 i_2$$

$$e = v + \begin{bmatrix} R_1 & 0 \\ 0 & R_2 \end{bmatrix} i$$

$$\Rightarrow v + R i = 2v \Rightarrow v = e/2$$

$$v = S v = \begin{bmatrix} R_{11} & R_{12} \\ R_{21} & R_{22} \end{bmatrix} \begin{bmatrix} v_1 \\ v_2 \end{bmatrix} = \begin{bmatrix} v_1 \\ v_2 \end{bmatrix}$$

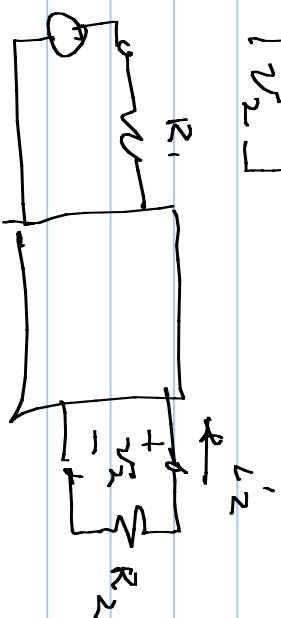
$$\text{If } v_2 = 0 \Rightarrow e_2 = 0 \Rightarrow \text{short circuit}$$

$$i_2 = -G_2 v_2$$

$$v_2 = R_{21} v_1 = \frac{v_2 - R_2 i_2}{2}$$

$$= \frac{v_2 + i_2 \cdot G_2 v_2}{2} = \frac{2v_2}{2} = v_2 = R_{21} v_1 = R_{21} \frac{e_1}{2}$$

$$R_{21} = \frac{v_2}{v_1} = \frac{v_2}{e_1/2} \Rightarrow A_v = \frac{R_{21}}{2} = \text{voltage gain}$$

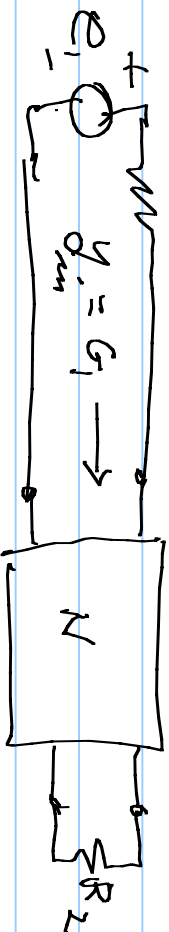


$$R_{11} = \frac{v_1^N}{v_1^i} \Big|_{v_2^i = 0} \quad v_2^i = 0 = v_2 + R_2 i_2 \Rightarrow v_2 = -R_2 i_2$$

↑
same case

$$= E_2 = 0$$

$= \frac{(v_1 - R_1 i_1)/2}{(v_1 + R_1 i_1)/2} = \text{reflection coefficient} = 0$ if $i_1 = G_1 v_1$, $G_1 = 1/R_1$, well matched



$$v_1 \cdot i_1 + v_2 \cdot i_2 = P_{in}(t) = v_1^{(t)} \cdot i_1^{(t)}$$

$$= (v_1^i + v_1^N)^T G (v_1^i - v_1^N)$$

$$= v_1^{i,T} G v_1^i - v_1^{i,T} G v_1^N + v_1^{N,T} G v_1^i - v_1^{N,T} G v_1^N$$

$$= v_1^{i,T} G v_1^i - v_1^{N,T} G v_1^N + v_1^{N,T} G v_1^i - v_1^{N,T} G v_1^N$$

$$= v_1^{i,T} G v_1^i - v_1^{N,T} G v_1^N$$

$(v_1^{i,T} G v_1^N)^T = v_1^{N,T} G v_1^i$, $G^T = G$
 as diagonal

if passive then energy in should be ≥ 0

if normality $G = I_2 \Rightarrow R_m(t) = v^T v - v^T v$

$$E_N(t) = \int_{-\infty}^t P_m(z) dz = \int_{-\infty}^t [v^T(z)v(z) - v^T(z)v(z)] dz$$

$dE_N(t) \geq 0 \Rightarrow v^T \in h_2 \Rightarrow v^T \in h_2$ \Downarrow Parseval

$$V_N^T = S(\alpha) V(\alpha) \Rightarrow \int_{-\infty}^{\infty} [V^T(\omega) V(\omega) - V^T(\omega) V(\omega)] d\omega \geq 0$$

$$\int [V^T(\omega) V(\omega) - V^T(\omega) S(\omega) S(\omega) V(\omega)] d\omega \geq 0 \text{ if } \text{Parseval's theorem} \text{ is valid}$$

$$= \int V^T(\omega) [I_2 - S(\omega) S(\omega)] V(\omega) d\omega \geq 0 \Rightarrow I_2 - S(\omega) S(\omega) \geq 0$$

$S(\alpha)$ is real for real α \Rightarrow is positive semi-definite

(if rational all coefficients are real)
 $S(\alpha)$ is analytic in $\sigma > 0$, $\alpha = \sigma + j\omega$

$$I_x - S(A)S(A)^T \succeq 0 \text{ and } S \succeq 0$$

↓
positive semidefinite

These 3 conditions define $S(A)$ as a bounded real matrix