

\Rightarrow $v_b = v_{bd} + v_b$ $v_b = v_{bd} + e$ $J =$ ^{power} independent sources
 $e =$ ^{power} independent sources

if all branches have admittances then $v_{bd} = Y_{b \times b} \cdot v_{bd}$

$$i_{bd} = \begin{bmatrix} G_2 & 0 & 0 & 0 & 0 & 0 \\ 0 & CA & 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 & 0 & 0 \\ 0 & 0 & g_m & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 & 1/AL & 0 \\ 0 & 0 & 0 & 0 & 0 & G_1 \end{bmatrix} \begin{bmatrix} v_{bd1} \\ v_{b2} \\ v_{b3} \\ v_{b4} \\ v_{b5} \\ -v_{b6} \end{bmatrix}; \quad J = \frac{\partial}{\partial v_b}, \quad e = \begin{bmatrix} v_1 \\ 0 \\ 0 \\ 0 \\ 0 \\ 0 \end{bmatrix} \quad R = \frac{d(e)}{dt}$$

$$i_{bd} \approx i_b - J \quad v_{bd} \approx v_b - e \quad \left. \begin{matrix} i_{bd} \approx Y_{b \times b} \cdot v_{bd} \\ v_{bd} \approx v_b - e \end{matrix} \right\} \text{Power of components}$$

Power for connection, KVL, KCL: $v_b = e^T v_f$, $i_b = \sigma^T i_f$

$$i_b - J = Y_{b \times b} (v_b - e) \Rightarrow \sigma^T i_f - J = Y_{b \times b} (e^T v_f - e) \quad \text{ⓐ}$$

$$\text{also } P_{in}(t) = \mathbf{V}_b^T \mathbf{i}'_b = (\mathbf{v}_f^T, (\mathbf{e}^T)^T), \mathbf{e}^T \cdot \mathbf{i}'_b = \mathbf{v}_f^T \mathbf{e} \mathbf{e}^T \cdot \mathbf{i}'_b = 0$$

for any circuit with the same graph, choose independent voltage sources in the tree and independent current sources in the links $\Rightarrow \mathbf{v}_f$ can be anything as can be $\mathbf{i}'_b \Rightarrow \mathbf{e} \mathbf{e}^T = \mathbf{0}_{f \times R}$

multiply \otimes by $\mathbf{e} =$ output matrix

$$\mathbf{e} \mathbf{e}^T \cdot \mathbf{i}'_b - \mathbf{e} \mathbf{J} = \mathbf{e} \mathbf{Y}_{b \times b} \mathbf{e}^T \mathbf{v}_f - \mathbf{e} \mathbf{Y}_{b \times b} \cdot \mathbf{e}$$

Options $\Rightarrow \mathbf{e} \mathbf{Y}_{b \times b} \mathbf{e} - \mathbf{e} \mathbf{J} = \mathbf{e} \mathbf{Y}_{b \times b} \mathbf{e}^T \mathbf{v}_f$ if $\mathbf{e} \mathbf{Y}_{b \times b} \mathbf{e}^T$ is nonsingular
 eg. current $\mathbf{v}'_N \Rightarrow \mathbf{v}'_f = (\mathbf{e} \mathbf{Y}_{b \times b} \mathbf{e}^T)^{-1} \mathbf{i}'_N \Rightarrow \mathbf{v}_b = \mathbf{e}^T \mathbf{v}_f$

$$\Rightarrow \mathbf{v}_{bd} = \mathbf{Y}_{b \times b} \mathbf{v}_{bd} \Rightarrow \mathbf{v}_b = \mathbf{v}_{bd} - \mathbf{J} \Rightarrow \text{restore the circuit}$$

$E Y_{6 \times 6} E^T$ is a TXT matrix

$$E Y_{6 \times 6} E^T = \begin{bmatrix} 1 & 0 & 1 & 1 & 0 & 1 \\ 0 & 1 & -1 & -1 & 0 & -1 \\ 0 & 0 & 0 & 0 & -1 & -1 \end{bmatrix}$$

$$\begin{bmatrix} G_i & 0 & 0 & 0 & 0 & 0 \\ 0 & C_A & 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & g_m & 0 & 0 \\ 0 & 0 & 0 & 0 & Y_{A_2} & 0 \\ 0 & 0 & 0 & 0 & 0 & G_L \end{bmatrix} \begin{bmatrix} 1 & 0 & 0 \\ 0 & 1 & 0 \\ 0 & -1 & 0 \\ 0 & -1 & 0 \\ 1 & -1 & -1 \\ 0 & 0 & 1 \\ 1 & -1 & -1 \end{bmatrix}$$

$$= \underbrace{\begin{bmatrix} G_i & 0 & g_m & 0 & 0 & G_L \\ 0 & C_A & -g_m & 0 & 0 & -G_L \\ 0 & 0 & -g_m & 0 & Y_{A_2} & -G_L \end{bmatrix}}_{E Y_{6 \times 6}} \begin{bmatrix} 1 & 0 & 0 \\ 0 & 1 & 0 \\ 0 & -1 & 0 \\ 1 & -1 & -1 \\ 0 & 0 & 1 \\ 1 & -1 & -1 \end{bmatrix} = \begin{bmatrix} G_i + g_m + G_L & -g_m - G_L & -G_L \\ -G_i - g_m & C_A + g_m + G_L & G_L \\ -g_m - G_L & g_m + G_L & \frac{1}{R_L} + G_L \end{bmatrix}$$

$$\text{gives } i_T \text{ and } v_T \Rightarrow i_T \rightarrow i_N = e Y_{ox} \cdot R = \begin{bmatrix} G_i & 0 & g_m & 0 & 0 & G_L \\ 0 & C_A & -g_m & 0 & 0 & -G_L \\ 0 & 0 & -g_m & 0 & Y_{a2} & -G_{a2} \end{bmatrix} \begin{bmatrix} v_1 \\ v_2 \\ v_3 \\ v_4 \\ v_5 \end{bmatrix}$$

$$= \begin{bmatrix} G_i v_1 \\ 0 \\ 0 \end{bmatrix} = \begin{bmatrix} G_i v_i \\ 0 \end{bmatrix} = \begin{bmatrix} Y_{11} & Y_{12} \\ Y_{21} & Y_{22} \end{bmatrix} \begin{bmatrix} v_{b1} \\ v_{b2} \\ v_{b5} \end{bmatrix}$$

$$\begin{bmatrix} 0 \\ 0 \end{bmatrix} = Y_{21} v_{b1} + Y_{22} \begin{bmatrix} v_{b2} \\ v_{b5} \end{bmatrix} \Rightarrow -Y_{22}^{-1} Y_{21} v_{b1} = \begin{bmatrix} v_{b2} \\ v_{b5} \end{bmatrix}$$

$$\Rightarrow G_1 v_1 = Y_{11} v_{b1} + Y_{12} (-Y_{22}^{-1} Y_{21}) v_{b1} = \underbrace{[Y_{11} - Y_{12} Y_{22}^{-1} Y_{21}]}_{Y_{11}'} v_{b1}$$

$$\therefore \text{need } Y_{22}^{-1} = \begin{bmatrix} C_a + g_m + G_L & G_L \\ g_m + G_L & \frac{1}{R_L} + G_L \end{bmatrix}^{-1}$$

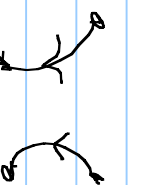
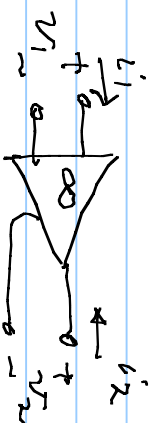
$$\begin{bmatrix} \frac{1}{R_L} + G_L & G_L \\ -G_L & C_a + g_m + G_L \end{bmatrix}$$

$$Y_{11} = Y_{12} Y_{22}^{-1} Y_{21} \Rightarrow [G_L + g_m + G_L] - [-g_m - G_L] \begin{bmatrix} \frac{1}{R_L} + G_L & G_L \\ -G_L & C_a + g_m + G_L \end{bmatrix} \begin{bmatrix} G_L - g_m \\ -g_m - G_L \end{bmatrix}$$

$$(C_a + g_m + G_L) \left(\frac{1}{R_L} + G_L \right) - G_L (g_m + G_L)$$

Examples of no-admittance devices:

ideal op-amp



$$b_1 = 0 \quad \left. \begin{array}{l} \\ \\ \end{array} \right\} \text{law}$$

$$v_1 = 0$$

can't write $i = Yv \Rightarrow$ ~~we~~ need more general

$$\begin{bmatrix} 1 & 0 \\ 0 & 0 \end{bmatrix} \begin{bmatrix} i_1 \\ i_2 \end{bmatrix} = \begin{bmatrix} 0 & 0 \\ 1 & 0 \end{bmatrix} \begin{bmatrix} v_1 \\ v_2 \end{bmatrix} \Rightarrow Av = B i$$