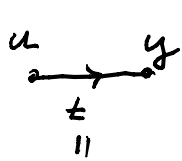


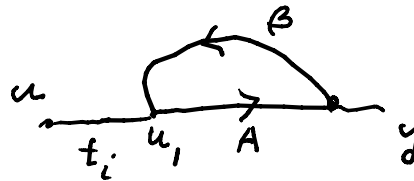
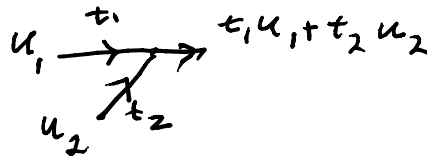
feedback, p. 805

Filters, p. 1261



transmittance

$$y = t u$$



$$y = A u_1$$

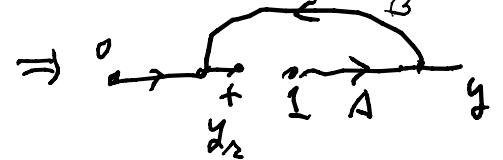
$$u_1 = t_1 u + \beta y$$

$$y = A(t_1 u + \beta y)$$

$$(1 - A\beta) y = A t_1 u$$

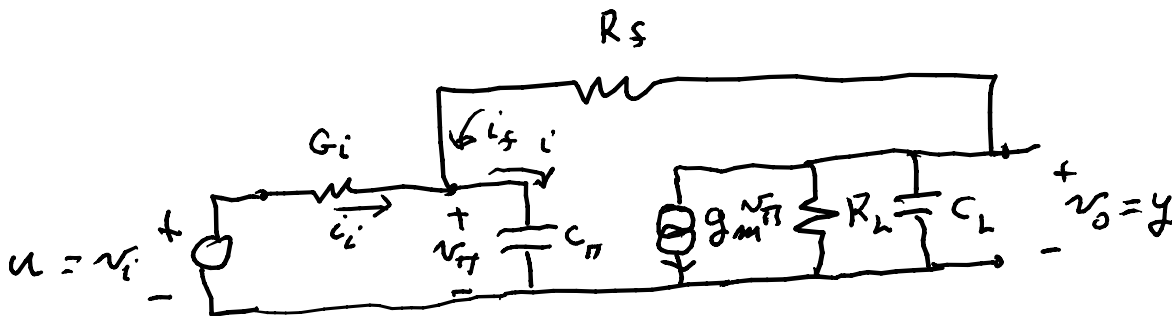
$$\frac{y}{u} = \text{transfer function} = \frac{A t_1}{1 - A\beta} = \text{return difference}$$

to get return ratio



" $A\beta = \text{return ratio}$

$$\text{return difference} = 1 - \text{return ratio} = 1 - A\beta$$



$$i = y_{\pi} v_{\pi} = \frac{1}{2 C_{\pi}} ; \quad A i = v_o$$

$$i = i_f + i_c$$

$$= G_f (v_o - v_{\pi}) + G_c (v_i - v_{\pi})$$

$$= G_f v_o + G_c v_i - (G_c + G_f) v_{\pi}$$

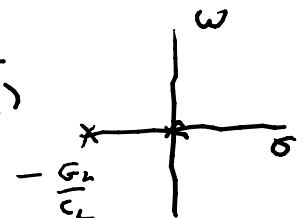
$$= G_f v_o + G_c v_i - (G_c + G_f) (i / y_{\pi})$$

$$(1 + (G_c + G_f) \frac{1}{y_{\pi}}) i = G_f v_o + G_c v_i$$

$$g_m \frac{i}{y_{\pi}} = -g_L v_o, \quad y_L = G_L + s C_L$$

$$A = -g_m / y_{\pi} y_L$$

$$= \frac{-g_m}{s C_{\pi} (G_L + s C_L)}$$



s-plane
 $s = \sigma + j\omega$

$$i = \frac{y_{\pi} G_S v_0 + y_{\pi} G_i v_i}{y_{\pi} + (G_i + G_S)} \Rightarrow v_0 = \frac{-g_m}{y_L y_{\pi}} \left[\frac{y_{\pi} G_S v_0 + y_{\pi} G_i v_i}{y_{\pi} + (G_i + G_S)} \right]$$

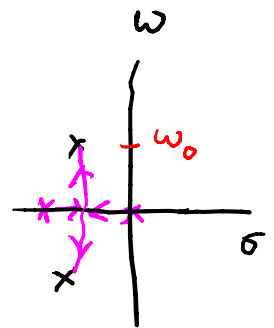
$$\left(1 + \frac{g_m G_S}{y_L [y_{\pi} + (G_i + G_S)]}\right) v_0 = \frac{-g_m G_i}{[y_{\pi} + (G_i + G_S)] y_L} v_i$$

$$\begin{aligned} \frac{v_0}{v_i}(s) &= \frac{-g_m G_i}{y_L y_{\pi} + y_L (G_i + G_S) + g_m G_S} = \frac{-g_m G_i}{(C_L + C_{C_L}) s C_{\pi} + (s C_L + G_L)(G_i + G_S) + g_m G_S} \\ &= \frac{-g_m G_i}{s^2 C_L C_{\pi} + s [C_L C_{\pi} + C_L (G_i + G_S)] + (G_L [G_i + G_S] + g_m G_S)} \\ &= \frac{-g_m G_i / C_L C_{\pi}}{s^2 + s \frac{1}{C_L C_{\pi}} (C_L C_{\pi} + C_L G_i + C_L G_S) + \frac{(G_L [G_i + G_S] + g_m G_S)}{C_L C_{\pi}}} \end{aligned}$$

denominator, $D(s) = s^2 + \frac{\omega_0}{Q} s + \omega_0^2$

$$(s + a_1)(s + a_2)$$

$$a_{1,2} = -\frac{\omega_0}{2Q} \pm \frac{1}{2} \sqrt{\frac{\omega_0^2}{Q^2} - 4\omega_0^2}$$



$$T(s) = \frac{v_0}{v_i}(s) = \frac{-\omega_0^2 A_0}{s^2 + \frac{\omega_0}{Q} s + \omega_0^2}$$

$$\Rightarrow \omega_0 \sqrt{1 - \frac{1}{4Q^2}}$$

if imaginary

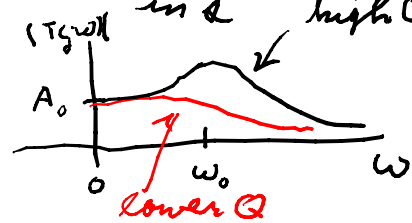
$T(s)$ \Rightarrow use for sinusoidal steady state

$$s = j\omega \quad T(j\omega) = \frac{-\omega_0^2 A_0}{(-\omega^2 + \omega_0^2) + j \frac{\omega_0 \omega}{Q}}$$

$$|T(j\omega)|^2 = T(j\omega) T^*(j\omega) = T(j\omega) T(-j\omega) = T(s) T(-s) \quad s = j\omega$$

$$T(s) T(-s) = T(-s) T(s) \Rightarrow T(s) T(-s) \Rightarrow \text{even function in } s \text{ "high Q"}$$

$$\text{here } |T(j\omega)|^2 = \frac{(-\omega_0^2 A_0)^2}{(\omega_0^2 - \omega^2)^2 + \left(\frac{\omega_0^2 \omega^2}{Q^2}\right)}$$



Maximally flat $T(s)$, low pass

$$T(s) = \frac{1}{s^m + a_{m-1}s^{m-1} + \dots + a_1s + a_0}$$

$$\frac{d|T(j\omega)|}{d\omega} = 0, \dots \text{ as many of } \frac{d|T(j\omega)|^m}{d\omega^m} = 0 \text{ as}$$

possible (limited #m is only possible)

look at $\frac{1}{|T(j\omega)|}$; $\frac{d|1/T(j\omega)|}{d\omega} = 0 = -\frac{1}{|T(j\omega)|^2} \cdot \frac{d|T(j\omega)|}{d\omega}$

\therefore use $\left| \frac{1}{T(j\omega)} \right| \Rightarrow \frac{1}{T(s)} = s^m + a_{m-1}s + \dots + a_0$

$$T(j\omega) T^*(j\omega) = T(j\omega) T(-j\omega) = |T(j\omega)|^2$$

$$\frac{d|T(j\omega)|^2}{d\omega} = 2|T(j\omega)| \frac{d|T(j\omega)|}{d\omega} \Rightarrow \text{same set } = 0 \text{ as setting } \frac{d|T(j\omega)|^2}{d\omega} = 0$$

$$\frac{1}{|T(j\omega)|^2} = \left| \frac{1}{s^m + a_{m-1}s^{m-1} + \dots + a_0} \right|_{s=j\omega}^2$$

$$= [(j\omega)^m + a_{m-1}(j\omega)^{m-1} + \dots] [(-j\omega)^m + a_{m-1}(-j\omega)^{m-1} + \dots + a_0]$$

even in ω (no j 's present)

$$= a_0^2 + \dots + \frac{a_{2m-2}}{2^{m-2}} \omega^{2m-2} + \omega^{2m} = a_0^2 + \omega^{2m}$$