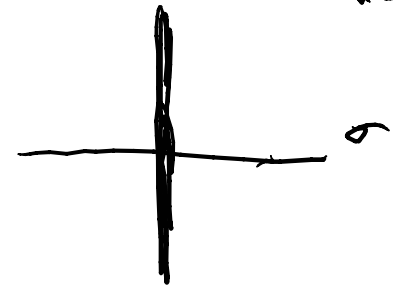
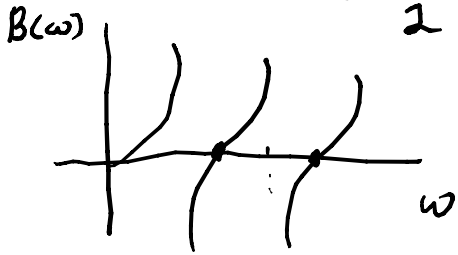


Reactance function $Z(s)$; rational + passive + $\frac{Z(s)+Z(-s)}{2} = 0$
 $s = \sigma + j\omega$

$P_{in}(j\omega) = 0$ for all ω

$\Rightarrow \frac{Z(s)+Z(-s)}{2} = 0 \Rightarrow$

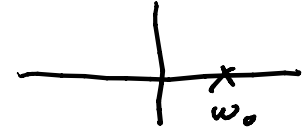


@ a pole $\approx \frac{ka}{a^2 + \omega_0^2}$

if passive, no poles in $\sigma > 0$

$E_r\left(\frac{ka}{a^2 + \omega_0^2}\right) = \frac{ka}{a^2 + \omega_0^2} + \frac{k(-a)}{(-a)^2 + \omega_0^2} = 0$

non passive $\frac{ka}{-a^2 + \omega_0^2}$ has $E_r = 0$



for passive no poles in $\sigma > 0$

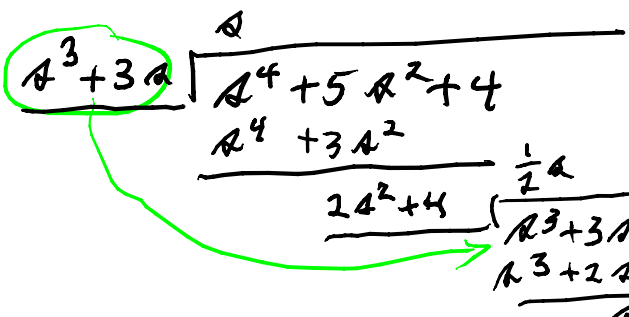
Ex: $Y(s) = \frac{(s^2+1)(s^2+4)}{s(s^2+3)}$; $Z(s) = \frac{s(s^2+3)}{(s^2+1)(s^2+4)}$ no pole at ∞
 $= \frac{s^3 + 3s}{s^4 + 5s^2 + 4}$

$Y(s) = \frac{s^4 + 5s^2 + 4}{s^3 + 3s}$

1st Cover, remove poles at ∞ (if none, there is a 0 at ∞)
 so turn it over & there is a pole)

divide highest power of den into num

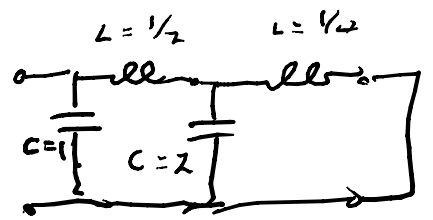
$Y(s) = s + \frac{2s^2 + 4}{s^3 + 3s} = Y_c + Y_{rem}$



$Y_{rem} = \frac{1}{Y_c} = \frac{s^3 + 3s}{2s^2 + 4}$ has a pole at ∞
 $Y_{rem} = \frac{1}{Y_c} = \frac{s^3 + 3s}{2s^2 + 4}$ has a pole at ∞

$$z(a) = \frac{1}{a + \frac{1}{\frac{1}{2}a + \frac{1}{2a + \frac{1}{\frac{1}{4}a}}}}$$

continued fraction expansion about $a = \infty$



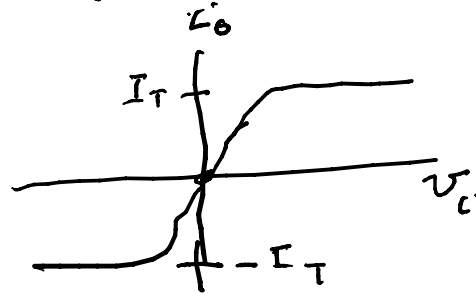
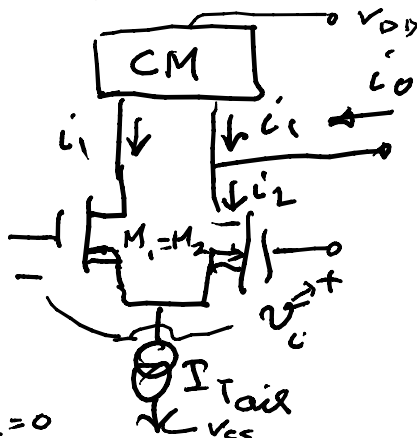
look at $y(s) = \frac{s(s^2+4)}{(s+1)(s^2+9)} = \frac{k_1}{(s+j3)} + \frac{k_2}{(s-j3)} + \frac{k_3}{s+1} + 1$

$$k_1 = \frac{s(s^2+4)(s+j3)}{(s+1)(s+j3)(s-j3)} \Big|_{s=-j3} = k_1 + \left[\frac{k_2(s+j3)}{(s-j3)} + \frac{k_3(s+j3)}{s+1} \right]$$

$$= \frac{-j3(-9+4)}{-j3+1} = \frac{+j15}{1-j3} = k_1 \neq \text{real}$$

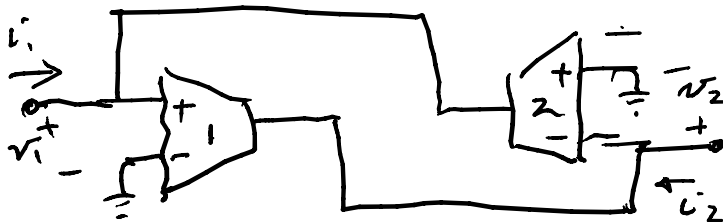
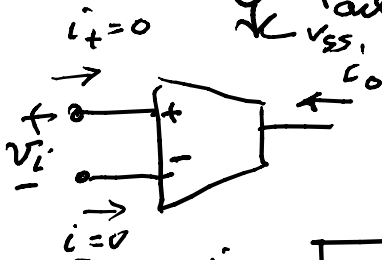
\therefore this $y(s)$ is not PR

OTA = operational transconductance amplifier



lineage @ $v_{i1} = 0 \Rightarrow g_m$

$$Y = \begin{bmatrix} 0 & 0 \\ g_m & 0 \end{bmatrix}$$



$$Y = \begin{bmatrix} 0 & -g_{m2} \\ g_{m1} & 0 \end{bmatrix}$$

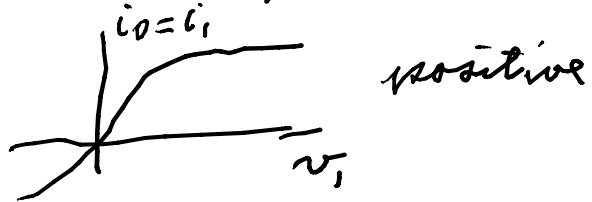
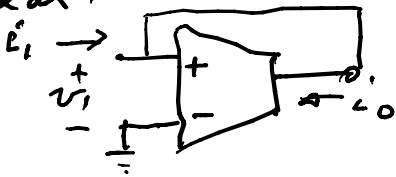
look at $\frac{V^T I + I^T V}{2}$; $V = \begin{bmatrix} v_1 \\ v_2 \end{bmatrix}$, $I = \begin{bmatrix} i_1 \\ i_2 \end{bmatrix}$

$$\frac{V^{TX} Y V + V^{TX} Y^{TX} V}{2} = \frac{V^{TX} (Y + Y^{TX}) V}{2} \text{ desire } \geq 0 \text{ for passive}$$

$$Y + Y^{TX} = Y + Y^T = \begin{bmatrix} 0 & -g_{m2} \\ g_{m1} & 0 \end{bmatrix} + \begin{bmatrix} 0 & g_{m1} \\ -g_{m2} & 0 \end{bmatrix} = \begin{bmatrix} 0 & g_{m1} - g_{m2} \\ g_{m1} - g_{m2} & 0 \end{bmatrix}$$

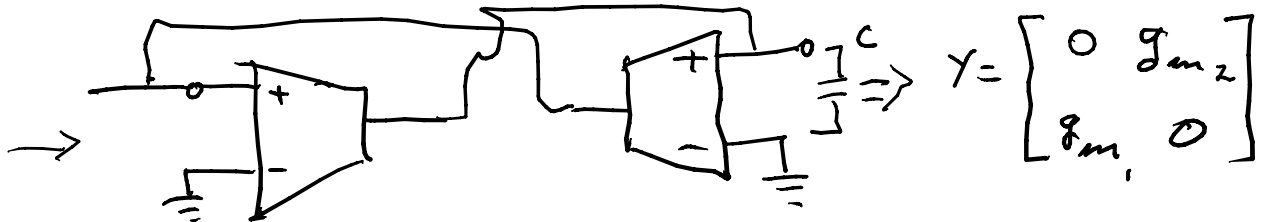
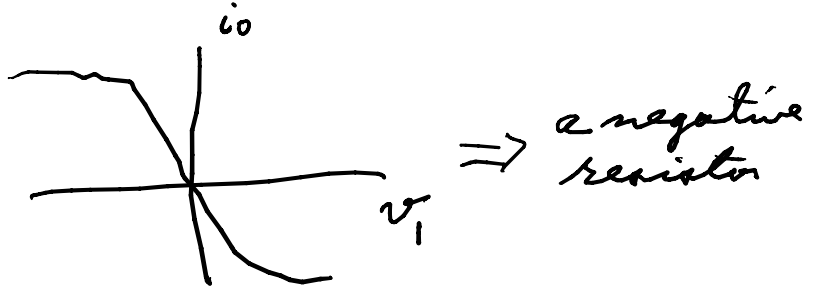
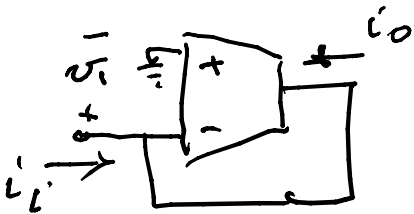
\therefore it is passive if Δ only if $g_{m1} = g_{m2} \Rightarrow$ lossless

look at:



$$\begin{bmatrix} v_1 & v_2 \end{bmatrix} \begin{bmatrix} 0 & g \\ g & 0 \end{bmatrix} \begin{bmatrix} v_1 \\ v_2 \end{bmatrix} = \begin{bmatrix} g v_2 & g v_1 \end{bmatrix} \begin{bmatrix} v_1 \\ v_2 \end{bmatrix} = 2g v_1 v_2$$

is < 0 if $v_1 = -v_2$



$$y_{in} = y_{11} - y_{12} \frac{1}{y_{22} + y_e} \cdot y_{21} = -g_{m2} \frac{1}{sC} \cdot g_{m1} =$$

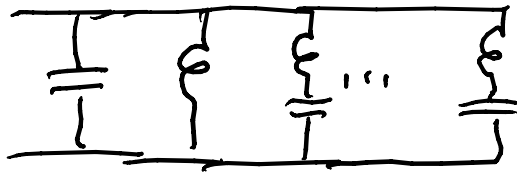
$$= \frac{-g_{m1} g_{m2}}{sC} \Rightarrow L_{eq} = \frac{-C}{g_{m1} g_{m2}} \Rightarrow \text{negative inductor}$$

$$= \frac{1}{L_{eq} s} \text{ residue @ the pole at } s=0 \text{ is } -\frac{g_{m1} g_{m2}}{C}$$

but if g_{m1} & g_{m2} have opposite signs the residue is positive if $C > 0 \Rightarrow y_{in}$ is PR

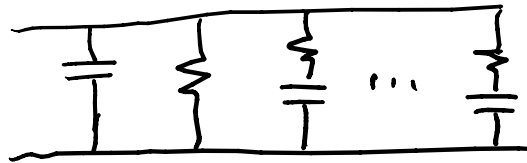
$\Sigma_k: C = 10^{-9} \text{Fd}, g_{m1} = 10^{-3} = -g_{m2}$
 $L_{eq} = \frac{10^{-9}}{10^{-3} \times 10^{-3}} = 10^{-3} \text{Hy} = 1 \text{mHy}$

LC circuit $y_{LC} = \sum \frac{2k_i s}{s^2 + \omega_i^2} + \frac{k_0}{s} + k_{\infty} s$



$y_{LC} = \frac{1}{\frac{1}{s} + \frac{1}{sC_i}}$
 \Downarrow
 $y_{RC} = \frac{1}{s_i + \frac{1}{sC_i}} = \frac{sC_i}{1 + s^2 C_i^2}$

can replace L by R & get all RC circuits have this



$\Rightarrow y_{RC} = k_0 + r_0 + \sum \frac{g_i s}{s + g_i/C_i}$

has all poles on negative σ axis

2nd Foster form

$\frac{y_{RC}(s)}{s} = k_{\infty} + \frac{r_0}{s} + \sum \frac{g_i}{s + g_i/C_i}$

here the g_i are residues & must be > 0 for a passive RC circuit

poles & zeros will alternate on σ axis