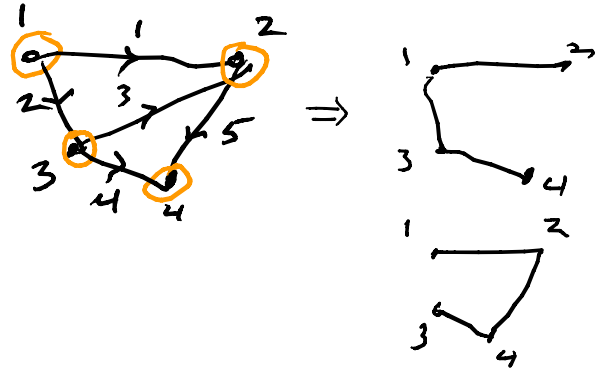


incidence matrix  
KCL @ nodes

$$\begin{bmatrix} 0 \\ 0 \\ 0 \\ 0 \end{bmatrix} = \begin{bmatrix} 1 & 1 & 0 & 0 & 0 \\ -1 & 0 & -1 & 0 & 1 \\ 0 & -1 & 1 & 1 & 0 \\ 0 & 0 & 0 & -1 & -1 \end{bmatrix} \begin{bmatrix} i_1 \\ i_2 \\ i_3 \\ i_4 \\ i_5 \end{bmatrix}$$

↑  
augmented  
incidence matrix  $A_a$ , singular;  $A_a A_a^T$  is  $4 \times 4$



Semi-state

$$E \dot{x} = Ax + Bu$$

$$y = Cx$$

for the differentiator  $T(s) = Q = C(AE - A)^{-1}B$

$$\begin{bmatrix} 0 & 0 \\ 1 & 0 \end{bmatrix} \dot{x} = \begin{bmatrix} 1 & 0 \\ 0 & 1 \end{bmatrix} x + Bu$$

$$y = Cx, \quad (AE - A)^{-1} = \begin{bmatrix} -1 & 0 \\ a & -1 \end{bmatrix}^{-1} = \frac{1}{1} \begin{bmatrix} -1 & 0 \\ -a & -1 \end{bmatrix}$$

$$y = Qu = [c_1, c_2] \begin{bmatrix} -1 & 0 \\ -a & -1 \end{bmatrix} \begin{bmatrix} b_1 \\ b_2 \end{bmatrix} u = [0 \ -1] \begin{bmatrix} -1 & 0 \\ -a & -1 \end{bmatrix} \begin{bmatrix} 1 \\ 0 \end{bmatrix} u = a \cdot u$$

$$\begin{bmatrix} 0 & 0 \\ 1 & 0 \end{bmatrix} \dot{x} = \begin{bmatrix} 1 & 0 \\ 0 & 1 \end{bmatrix} x + \begin{bmatrix} 1 \\ 0 \end{bmatrix} u$$

$$y = [0 \ -1]x \Rightarrow T(s) = a$$

If  $E$  is nonsingular then  $\hat{x} = E^{-1}Ax + E^{-1}Bu = \hat{A}x + \hat{B}u$

$$y = Cx + Du$$

$$\Rightarrow T(s) = C(AI_n - \hat{A})^{-1}\hat{B} + D$$

$$\text{as } a \rightarrow \infty \Rightarrow T(s) \Rightarrow \frac{c_1 b_1}{a} \Rightarrow 0 \text{ no pole at } \infty$$

Passive circuits  $P_{in} \geq 0$ ;  $E(t) = \int_0^t p(\tau) d\tau + E(0) \geq 0$   
 $E(t) = \int_0^t v^T(\tau) i(\tau) d\tau \geq 0$  for all  $t > 0$  means passive  
 & all  $v(t) \geq i(t)$  supported by a circuit

When does  $y(s)$  describe a passive circuit

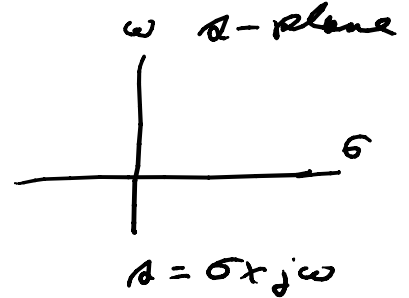
$y(s)$  needs to be positive-real

conditions are

$\text{Re } y(s) \geq 0$  in  $\sigma > 0$  (passive)

no singularities in  $\sigma > 0$  (stable)

$y(\sigma)$  is real for  $\sigma > 0$  (real elements)



Ex: 1)  $y(s) = \sqrt{s}$   $\infty$  long passive RC line

note if  $s = \sigma = -1$  then  $y(-1) = \sqrt{-1} = j$  not real

but  $y(\sigma) = \sqrt{\sigma}$  is real if  $\sigma > 0$

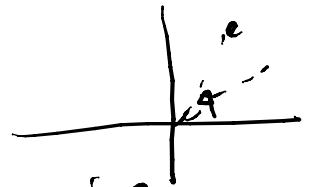
is not analytic at  $s=0$  as  $\frac{dy}{ds} = \frac{1/2}{\sqrt{s}} \rightarrow \infty$  @  $s=0$

only singularity is at  $s=0$

$$\text{Re } y(s) = \frac{\sqrt{s} + \sqrt{s^*}}{2} = \frac{|s|^{1/2} e^{j\Delta/2} + |s|^{1/2} e^{-j\Delta/2}}{2}$$

$$= |s|^{1/2} \cos(\Delta/2) > 0$$

for  $s$  in  $\sigma > 0$



2)  $y(s) = N(s)/D(s)$ ,  $N(s)$  &  $D(s)$  polynomial

rational  $y(s)$ : conditions

all positive-real rational PR

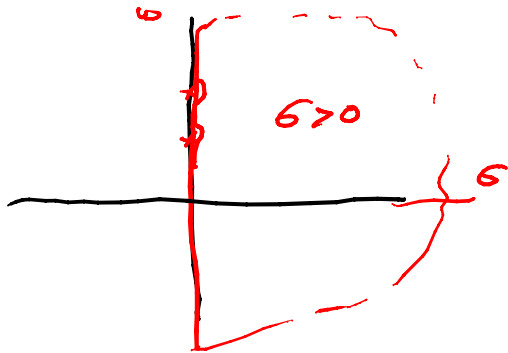
Real for real  $s$ ,  $\sigma > 0 \Rightarrow$  real coefficients  
 ( $y(\sigma)$  real for  $\sigma > 0$ )

analytic in  $\sigma > 0$ ; can only have poles for singular points  $\Rightarrow$  no poles in  $\sigma > 0$

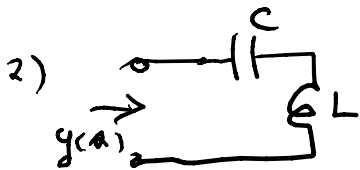
but poles on  $j\omega$  axis are simple & with a positive real residue

Check  $\text{Re } y(s)$  in  $\sigma > 0 \Rightarrow \text{Re } y(j\omega) \geq 0$

where we choose the boundary of  $\sigma > 0$  as the  $j\omega$  axis closed at  $\infty$  on the right



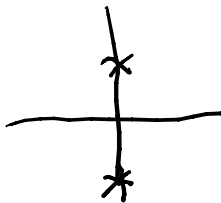
Ex: 1)  $Y(s) = G = 1/R$       $\int_{-\infty}^{\infty} R \Rightarrow R \text{ real} \geq 0$



$Z(s) = Ls + \frac{1}{Cs}$  ;  $Y(s) = \frac{1}{Z(s)} = \frac{Cs}{1 + LCs^2}$

need  $LC \geq C$  real  $\Rightarrow L \ \& \ C$  real

poles  $Y(s) \Rightarrow 0 = 1 + LCs^2 \Rightarrow s^2 = -1/LC$  ;  $s = \pm j\sqrt{1/LC}$



$$Y(s) = \frac{s/L}{(s^2 + \frac{1}{LC})} = \frac{(\frac{1}{2})s}{(s + j\sqrt{1/LC})(s - j\sqrt{1/LC})}$$

$$= \frac{k_1}{s + j\sqrt{1/LC}} + \frac{k_2}{s - j\sqrt{1/LC}}$$

$k_1 =$  residue at  $s = -j\sqrt{1/LC}$  multiply by  $s + j\sqrt{1/LC}$

$$(s + j\sqrt{1/LC})Y(s) = \frac{1}{L} s / (s - j\sqrt{1/LC}) = k_1 + \frac{k_2 s + j\sqrt{1/LC}}{s - j\sqrt{1/LC}}$$

$$k_1 = \left. \frac{1}{L} \frac{s}{s - j\sqrt{1/LC}} \right|_{s = -j\sqrt{1/LC}} = \frac{1}{L} \frac{(-j\sqrt{1/LC})}{-2j\sqrt{1/LC}} = \frac{1}{2L} \Rightarrow L > 0 \text{ for } k_1 \text{ positive}$$

$k_2 = 1/2L$

if  $Y(s)$  is PR then so is  $Z(s) = 1/Y(s)$

here  $Z(s) = \frac{1}{Cs} + Ls$

$\Rightarrow C > 0$  for  $Y(s)$  to be PR

