

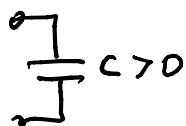
$S(s)$  is bounded-real;  $S_{n \times n}$   
if the circuit is passive

EE 610  
09/21/10 b

- real 1.  $S(s)$  is real for  $\sigma > 0$  (real coefficients if ratios of polynomials)
- stable 2.  $S(s)$  is analytic in  $\sigma > 0$   
& no poles on  $\sigma = 0$
- passive 3.  $\frac{1}{n} - S^T(s)S(s)$  is positive semi-definite in  $\sigma > 0$ ;  $s = \sigma + j\omega$

$$\Rightarrow a^{T*} (\frac{1}{n} - S^T(s)S(s)) a \geq 0 \text{ for all complex } a \text{ vectors}$$

Example:



$S(s)$

$$i(t) = C \frac{dv}{dt} \Rightarrow I(s) = C A V(s)$$

$$a = \frac{v + v^i}{2r} = \frac{v^i}{r}$$

$$b = \frac{v - v^i}{2r} = \frac{v^r}{r}$$

$$B(s) = S(s) A(s)$$

$$V^r = S(s) V^i$$

$$v = v^i + v^r$$

$$i = v^i - v^r$$

$$v^i - v^r = C A (v^i + v^r)$$

$$-v^r - C A v^r = C A v^i - v^i$$

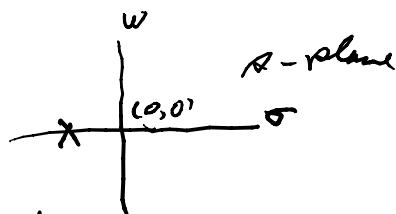
$$v^r = \frac{1}{1 + C A} (1 - C A) \cdot v^i \Rightarrow S(s) = \frac{1 - C A}{1 + C A}$$

1) true  $\Rightarrow$  coeff real

2) true  $\Rightarrow$  pole at  $\sigma = -1/C$

3)  $1 - \left( \frac{1 - C A^*}{1 + C A^*} \right) \left( \frac{1 - C A}{1 + C A} \right) = 1 - \frac{|1 - C A|}{|1 + C A|}$  in  $s = \sigma + j\omega$   
where  $\sigma > 0$

$$= 1 - \left| \frac{1 - C\sigma - jC\omega}{1 + C\sigma + jC\omega} \right| = 1 - \frac{((1 - C\sigma)^2 + C^2\omega^2)^{1/2}}{((1 + C\sigma)^2 + C^2\omega^2)^{1/2}} > 0$$



Note on  $S(j\omega)$ : For the capacitor  $S(j\omega) = \frac{1-j\omega C}{1+j\omega C}$

$$S^*(j\omega) = S(-j\omega) = \frac{1+j\omega C}{1-j\omega C} = \frac{1}{S(j\omega)}$$

$\therefore 1 - S^*(j\omega)S(j\omega) = 1 - 1 = 0$  for the capacitor

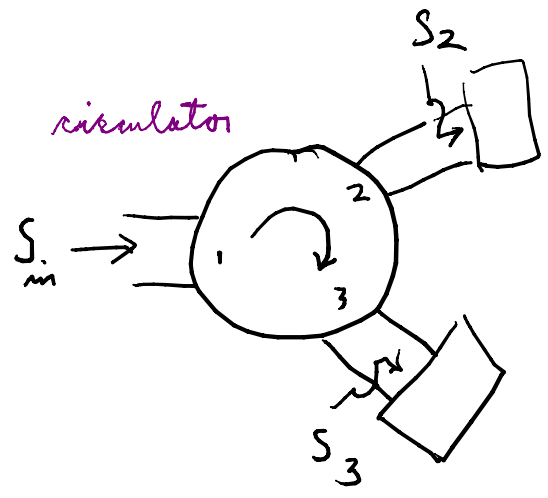
and  $S(-s) \Big|_{s=j\omega} = S^*(j\omega)$

poles of  $Y \neq Z$  can be on  $j\omega$  axis:

$$Y(s) = sC, \quad Z(s) = \frac{1}{sC}$$

$\uparrow$  pole at 0                       $\uparrow$  0 } on  $j\omega$  axis

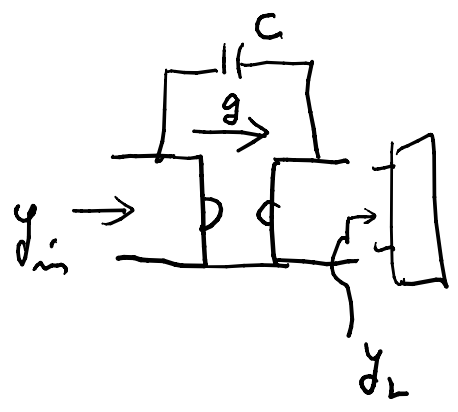
but for  $S(s)$  example:  $S(s) = \frac{1-sC}{1+sC}$  no pole in  $\sigma \geq 0$  if  $C > 0$



$$S_{in} = S_3 \cdot S_2$$

only if same  $V$

Note:  $V_{2_{in}} = V_2 - V I_{2_{in}} = V_2 + V I_{load} = V_{2_{load}}$   
 (here  $I_{2_{in}} = -I_{load}$ )



Richards' p. 361  
function

$Y_L$  is related to  $Y_{in}$  via the Richards' function (times  $Y_{in}$ )