

setting up state variable equations  
for a single input single output  
given the transfer function

EE 610  
10/29/09

$$\frac{y(s)}{u(s)} = \frac{m(s)}{d(s)} = \frac{m_{s-1}s^{s-1} + \dots + m_1s + m_0}{s^s + d_{s-1}s^{s-1} + \dots + d_1s + d_0} = T(s)$$

$$E \dot{x} = Ax + Bu \quad ; \quad E = I_s, \quad s = \text{degree of } T(s) \quad x = \begin{bmatrix} x_1 \\ \vdots \\ x_s \end{bmatrix}$$

$$y = Cx$$

$$T(s) = C(sI_s - A)^{-1}B$$

Here we can choose

$$\dot{x} = \begin{bmatrix} 0 & 1 & 0 & \dots & 0 \\ 0 & 0 & 1 & \dots & 0 \\ \vdots & \vdots & \vdots & \ddots & \vdots \\ 0 & \dots & \dots & \dots & 1 \\ -d_0 & -d_1 & \dots & -d_{s-2} & -d_{s-1} \end{bmatrix} x + \begin{bmatrix} 0 \\ \vdots \\ 0 \\ 1 \end{bmatrix} u \quad ; \quad y = [m_0, m_1, \dots, m_{s-1}]$$

Ex:  $T(s) = \frac{3s^2 + 5s + 6}{s^3 + 2s^2 + 3s + 4}$

$$x = \begin{bmatrix} x_1 \\ x_2 \\ x_3 \end{bmatrix}, \quad s[T(s)] = s = 3$$

$$\dot{x} = \begin{bmatrix} 0 & 1 & 0 \\ 0 & 0 & 1 \\ -4 & -3 & -2 \end{bmatrix} x + \begin{bmatrix} 0 \\ 0 \\ 1 \end{bmatrix} u \quad ; \quad y = [6 \ 5 \ 3]$$

$$T(s) = [6 \ 5 \ 3] \begin{bmatrix} s & -1 & 0 \\ 0 & s & -1 \\ 4 & 3 & s+2 \end{bmatrix}^{-1} \begin{bmatrix} 0 \\ 0 \\ 1 \end{bmatrix} ; \quad \det[sI_3 - A] =$$

$$s(s(s+2)+3) + 4(1-0) = s^3 + 2s^2 + 3s + 4$$

$$= \frac{[6 \ 5 \ 3]}{s^3 + 2s^2 + 3s + 4} \begin{bmatrix} \text{last} \\ \text{column} \\ \text{of} \\ (sI_3 - A)^{-1} \\ \times \det \end{bmatrix} ; \quad \begin{array}{l} \text{entries of last column} \\ (1,3) = (-1) \det \begin{bmatrix} -1 & 0 \\ s & -1 \end{bmatrix} = 1 \\ (2,3) = (-1) \det \begin{bmatrix} s & 0 \\ 0 & -1 \end{bmatrix} = s \\ (3,3) = (-1) \det \begin{bmatrix} s & -1 \\ 0 & s \end{bmatrix} = s^2 \end{array}$$

$$= [6 \ 5 \ 3] \begin{bmatrix} 1 \\ s \\ s^2 \end{bmatrix} \times \frac{1}{s^3 + 2s^2 + 3s + 4}$$

$$= \frac{6 + 5s + 3s^2}{s^3 + 2s^2 + 3s + 4}$$

for general case the numerator will be

$$\underbrace{[m_0, m_1, \dots, m_{s-1}]}_C \times \begin{bmatrix} \text{last column} \\ \text{of } (sI_s - A)^{-1} \\ \times \det(sI_s - A) \end{bmatrix} = C \begin{bmatrix} 1 \\ a_1 \\ a_2 \\ \vdots \\ a_{s-1} \end{bmatrix} = m_0 + m_1 s + \dots + m_{s-1} s^{s-1}$$

for the last column of  $(sI_s - A)^{-1}$

$$\begin{bmatrix} a & -1 & 0 & \dots & 0 \\ 0 & a & -1 & 0 & \dots & 0 \\ & & & & & 0 \\ & & & & a & -1 \\ d_0 & d_1 & \dots & d_{s-2} & d_{s-1} & 0 \end{bmatrix}$$

for the  $(i, s)$  entry

$$= (-1)^{i+s} M(i, s \text{ col}, s \text{ row})$$

$$\det \begin{bmatrix} a & a & 0 \\ & \ddots & \\ 0 & -1 & \dots \\ & & \ddots & \\ & & & a & -1 \end{bmatrix}$$

$$= (-1)^{i+s} a^{i-1} (-1)^{(s-1)-(i+1)} = a^{i-1} (-1)^{s-2} = a^{i-1}$$

*minimal*  
∴ this gives a "realization"

$$\dot{x} = Ax + Bu$$

$$y = Cx \quad \delta^{-1} \text{ of } T(s) = \left( \sum_{i=0}^{s-1} n_i s^i \right) / \left( a s^s + \sum_{i=0}^{s-1} d_i s^i \right)$$

(here  $T(\infty) = 0$  at  $s = \infty$ )

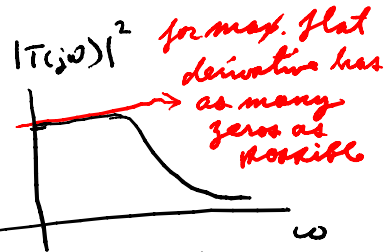
*minimal*

$$\text{size of } X = \delta [T(s)] = s$$

How to get transfer functions  
Maximally flat, low pass

$$T(s) \rightarrow |T(j\omega)|^2 = T(j\omega) T^*(j\omega) = T(j\omega) T(-j\omega) \text{ if "real"}$$

$$\Rightarrow |T(j\omega)|^2 \text{ is even in } \omega \text{ (i.e. } |T(j\omega)|^2 = |T(-j\omega)|^2)$$



$$|T(j\omega)|^2 = f(\omega^2) = \frac{1}{\omega^{2n} + a_{2n-2} \omega^{2n-2} + \dots + a_2 \omega^2 + a_0}$$

$$f(x) = \frac{1}{x^n + a_{2n-2} x^{n-1} + \dots + a_2 x + a_0} \quad ; \quad \frac{d f(\omega^2)}{d \omega^2} = \frac{d f(x)}{d x}$$

$$\frac{d f(x)}{d x} = -\frac{1}{D(x)^2} \cdot \frac{d D(x)}{d x} \Rightarrow \text{for maximally flat desire as many derivatives of } D(x) \text{ to be zero at } x=0 \uparrow \text{ in a Taylor series expansion}$$

$$= -\frac{1}{D(x)^2} \cdot [a_2 + 2a_3 x + \dots + (n-1)a_n x^{n-2} + n a_n x^{n-1}]$$

↑  
set to zero to get zeros of Taylor series

$$S(x) = \frac{1}{x^n + a_0} \Rightarrow T(j\omega)T(-j\omega) = \frac{1}{\omega^{2n} + a_0}$$

can normalize  $a_0 > 0$  to 1 by dividing  $\frac{1/a_0}{(\omega/a_0^{1/2n})^{2n} + 1} \Rightarrow \frac{1}{\omega^{2n} + 1}$

$$T(j\omega) \cdot T(-j\omega) = \frac{1}{\omega^{2n} + 1} \quad ; \quad j = \sqrt{-1}$$

$$\text{let } \omega \rightarrow s/j \Rightarrow T(s) \cdot T(-s) = \frac{1}{\left(\frac{s}{j}\right)^{2n} + 1} = \frac{1}{s^{2n}(-1)^n + 1} = \frac{1}{(-1)^n s^{2n} + 1}$$

now have an analytic function in  $s$ ,  $T(s)T(-s)$  has  $2n$  poles, the roots of  $(-1)^n s^{2n} = -1 \Rightarrow s^{2n} = (-1)^{n+1}$

$\therefore$  derive the  $2n$  roots of  $-1$  or  $+1$

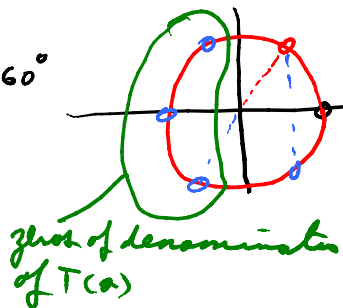
$$\text{case 1: } n \text{ odd} \quad s^{2n} = 1 = e^{j2\pi k} \quad , k = 0, 1, \dots$$

$$\Rightarrow s_k = e^{j\frac{2\pi k}{2n}} \quad , k = 0, 1, \dots, 2n-1$$

$$n=3$$

$$\frac{360^\circ}{6} = 60^\circ$$

$$6=2n$$



← zeros of denominator of  $T(s)T(-s)$

need for stability to choose left half plane zeros as in denominator

$$T(s) = \frac{1}{(s+1) \left( s + \cos 60^\circ + j \sin 60^\circ \right) \left( s + \cos 60^\circ - j \sin 60^\circ \right)}$$

$\frac{1}{2} \quad \frac{\sqrt{3}}{2}$

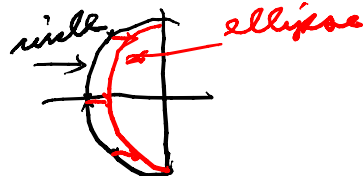
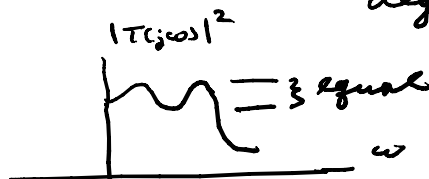
$$= \frac{1}{(s+1) \left( s + \frac{1}{2} + j\frac{\sqrt{3}}{2} \right) \left( s + \frac{1}{2} - j\frac{\sqrt{3}}{2} \right)}$$

$$= \frac{1}{(s+1) \left( s^2 + \frac{1}{4} + \frac{3}{4} + \left(\frac{1}{2} + j\frac{\sqrt{3}}{2}\right)s \right)} = \frac{1}{(s+1)(s^2 + s + 1)}$$

$$= \frac{1}{s^3 + 2s^2 + 2s + 1}$$

← universal low-pass maximally flat degree 3 "filter"

Equal ripples



for high pass set  $Q \rightarrow 1/Q$   
" band pass set  $Q \rightarrow \frac{p}{\omega_0} + \frac{\omega_0}{p}$