

EE610  
10/15/09

Richards' function, p. 361

Lossless 1-port synthesis, p. 342-

$$y(s) = \frac{s(s^2+6)}{(s^2+4)(s^2+9)} = \frac{s^3+6s}{s^4+13s^2+36}$$

For 2nd Case; continued fraction expansion about  $s=0$

$$\frac{1}{y(s)} = z(s) = \frac{36+13s^2+s^4}{6s+s^3} \Rightarrow z(s) = \frac{6s+s^3}{36+13s^2+s^4} = \frac{1}{\frac{6}{s} + \frac{7s^2+24}{6s+s^3}}$$

$$= \frac{1}{\frac{6}{s} + \frac{1}{\frac{6s+s^3}{7s^2+24}}}$$

$$= \frac{1}{\frac{6}{s} + \frac{1}{\frac{6s+\frac{6}{7}s^3}{\frac{49}{s} + \frac{1}{7s}}}}$$

$$= \frac{1}{\frac{6}{s} + \frac{1}{\frac{6s+\frac{6}{7}s^3}{\frac{49}{s} + \frac{1}{7s}}}}$$

$$y(s) = \frac{1}{\frac{6}{s} + \frac{1}{\frac{6s+s^3}{7s^2+24}}} = \frac{6s+s^3}{36+13s^2+s^4}$$

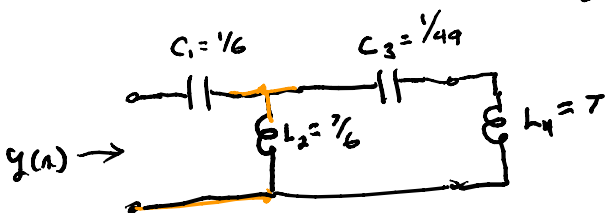
$$z_1(s) = \frac{6}{s}$$

$$z_2(s) = \frac{6}{7s}$$

$$z_3(s) = \frac{49}{s}$$

$$z_4(s) = \frac{1}{7s}$$

= continued fraction expansion about  $s=0$



2nd Case synthesis of  $y(s) = \frac{s^3+6s}{s^4+13s^2+36}$

∴ have 4 syntheses of this  $y(s)$ , each of which uses the minimum number, the degree of  $y(s)$ , in each  $S[y(s)] = \text{degree}$  (here = highest power of  $s$  seen in  $y(s)$ )  
[each step reduced the degree by 1, so synthesis of a finite lossless  $y(s)$  by 1st or 2nd Foster or Cauer]

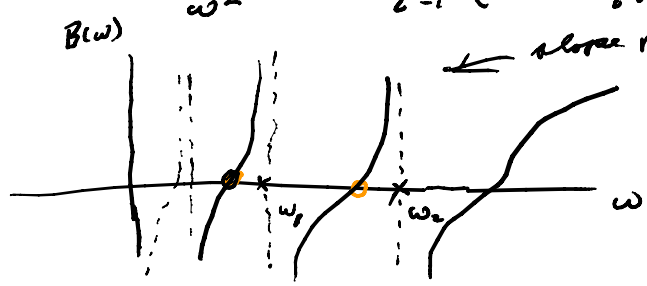
$$y(s) = -y(-s) = \frac{k_0}{s} + k_{\infty}s + \sum_{i=1}^n \frac{2k_i s}{s^2 + \omega_i^2} \quad k_i \geq 0$$

For  $s = j\omega$ ;  $y(j\omega) = j B(\omega)$ ;  $\frac{dB(\omega)}{d\omega}$

$$B(\omega) = \frac{y(j\omega)}{j} = \frac{-k_0}{\omega} + k_{\infty}\omega + \sum_{i=1}^m \frac{2k_i\omega}{-\omega^2 + \omega_i^2}$$

$$\frac{dB(\omega)}{d\omega} = \frac{k_0}{\omega^2} + k_{\infty} + \sum_{i=1}^m \left\{ \frac{2k_i}{-\omega^2 + \omega_i^2} - \frac{2k_i\omega(-2\omega)}{(-\omega^2 + \omega_i^2)^2} \right\}$$

$$= \frac{k_0}{\omega^2} + k_{\infty} + \sum_{i=1}^m \frac{2k_i(\omega^2 + \omega_i^2)}{(-\omega^2 + \omega_i^2)^2} \geq 0$$



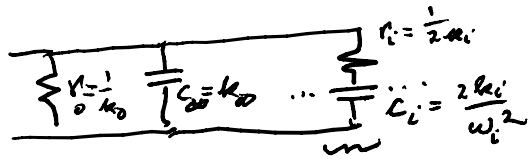
← slope positive  
shows between every pole there is a zero  
(allows us to easily create reactance functions)

Look at RC PR functions  $\Rightarrow$  had  $y_{RC}(s)$   
given one can replace R's by L's  $\Rightarrow$  LC circuit  
synthesize the LC circuit by its  $y_{LC}(s)$  then replace every L by an R gives  $y_{RC}$

$$y_{LC} = \frac{k_0}{s} + k_{\infty}s + \sum_{i=1}^m \frac{2k_i s}{s^2 + \omega_i^2} = \frac{k_0}{s} + k_{\infty}s + \sum_{i=1}^m \frac{1}{\frac{1}{2k_i} + \frac{\omega_i^2}{2k_i s}}$$

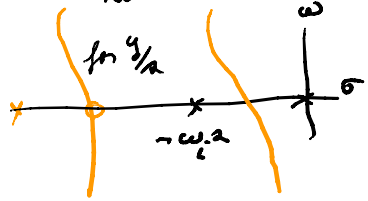
$\uparrow$   $L = 1/k_0$                        $\uparrow$   $L = \frac{1}{2k_i}$

$$y_{RC} = k_0 + k_{\infty}s + \sum_{i=1}^m \frac{1}{\frac{1}{2k_i} + \frac{\omega_i^2}{2k_i s}} = k_0 + k_{\infty}s + \sum_{i=1}^m \frac{2k_i s}{s + \omega_i^2}$$



$$\frac{y_{RC}(s)}{s} = \frac{k_0}{s} + k_{\infty} + \sum_{i=1}^m \frac{2k_i}{s + \omega_i^2}$$

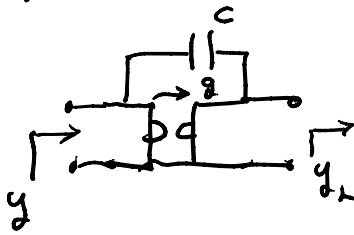
is now a partial fraction expansion



$$\frac{dy(s)/s}{ds} = -\frac{k_0}{s^2} + \sum \frac{-2k_i}{(s + \omega_i^2)^2} \leq 0$$

poles and zeros will alternate & are on the negative s axis

# Synthesis via the Richards' function



$$Y_{comp} = \begin{bmatrix} \alpha C & -\alpha C + g \\ -\alpha C - g & \alpha C \end{bmatrix} j$$

$$-y_L \cdot v_2 = \begin{bmatrix} i_1 \\ i_2 \end{bmatrix} = \begin{bmatrix} \alpha C & -\alpha C + g \\ -\alpha C - g & \alpha C \end{bmatrix} \begin{bmatrix} v_1 \\ v_2 \end{bmatrix}$$

$$(-\alpha C - g_L) v_2 = (-\alpha C - g) v_1 \Rightarrow v_2 = \frac{\alpha C + g}{\alpha C + g_L} v_1$$

$$i_1 = y \cdot v_1 = \left[ \alpha C + (-\alpha C + g) \frac{(\alpha C + g)}{\alpha C + g_L} \right] \cdot v_1$$

$$y = \frac{\alpha^2 C^2 + \alpha C g_L - \alpha^2 C^2 + g^2}{\alpha C + g_L} = \frac{\alpha C g_L + g^2}{\alpha C + g_L} = y$$

$$\Rightarrow \alpha C g_L + g^2 = \alpha C y + g_L y \Rightarrow (\alpha C - y) g_L = \alpha C y - g^2$$

$$g_L = \frac{\alpha C y - g^2}{\alpha C - y} = g^2 \left( \frac{1 - \frac{\alpha C}{g} \cdot \frac{y(s)}{g}}{y - \alpha C} \right) = g \left( \frac{1 - \frac{\alpha C}{g} \frac{y(s)}{g}}{\frac{y(s)}{g} - \frac{\alpha C}{g}} \right)$$

$$= g \left( \frac{\frac{\alpha C}{g} \frac{y(s)}{g} - 1}{\frac{\alpha C}{g} - \frac{y(s)}{g}} \right)$$

The Richards function:

p. 361

$$R(\alpha) = \frac{k y(\alpha) - \alpha y(k)}{k y(\alpha) - \alpha y(k)}$$

use

$$R_y(\alpha) = \frac{k y(\alpha) - \alpha y(k)}{k y(\alpha) - \alpha y(k)}$$

these are PR  
if  $k$  is real &  $> 0$   
&  $y(\alpha)$  is PR

$k$  has a pole at  $\alpha = k$   
& a zero & they cancel

I want  $y_L$  to look like  $R_y(\alpha)$

$$R_y(\alpha) = \frac{k y(\alpha) \left[ 1 - \frac{\alpha}{k} \cdot \frac{y(\alpha)}{y(\alpha)} \right]}{k y(\alpha) \left[ \frac{y(\alpha)}{y(\alpha)} - \frac{\alpha}{k} \right]} \sim g \left[ \frac{1 - \frac{\alpha}{(g/c)} \cdot \frac{y(\alpha)}{g}}{\frac{y(\alpha)}{g} - \frac{\alpha}{(g/c)}} \right] = y_L$$

form  $\frac{R_y(\alpha)}{g} \triangleq g = y(k), k = g/c$        $g = y(k), C = \frac{g}{k} = \frac{y(k)}{k}$

Choose  $k$  so that it is a zero of the even part of  $y(\alpha)$   
for lossless any  $k$  is such a zero.