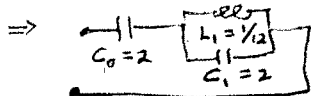


#1.  $y(s) = \frac{s(s^2+6)}{s^2+3} \Rightarrow z(s) = \frac{1}{y(s)} = \frac{s^2+3}{s(s^2+6)}$

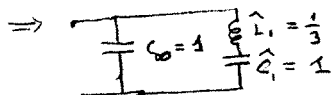
1st. Poles:  $z(s) = \frac{k_0}{s} + \frac{k_1 s}{s^2+6}$ ,  $k_0 = s z(s)|_{s=0} = \frac{s^2+3}{s^2+6}|_{s=0} = \frac{3}{6} = \frac{1}{2}$

$k_1 = \frac{s^2+6}{s} z(s)|_{s^2=-6} = \frac{s^2+3}{s^2}|_{s^2=-6} = \frac{-6+3}{-6} = \frac{-3}{-6} = \frac{1}{2}$   
 $= \frac{1/2}{s} + \frac{1/2 s}{s^2+6} = \frac{1}{2s} + \frac{1}{2s+12}$



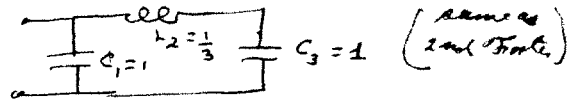
2nd. Poles:  $y(s) = k_{\infty} s + \frac{\hat{k}_1 s}{s^2+3}$ ,  $k_{\infty} s = \frac{s^3}{s^2} \Rightarrow k_{\infty} = 1$

$\hat{k}_1 = \frac{s^2+3}{s} y(s)|_{s^2=-3} = \frac{s^2+6}{s^2-3}|_{s^2=-3} = \frac{-3+6}{-3-3} = 3$   
 $= s + \frac{3s}{s^2+3} = s + \frac{1}{\frac{1}{3}s + \frac{1}{s}}$



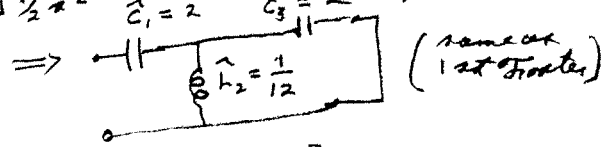
1st. Zeros:  $\frac{s^2+3}{s^3+6s} = \frac{s^2+3}{s^2+3s} = \frac{\frac{1}{3}s}{\frac{1}{3}s} = \frac{1}{3}$

$\Rightarrow y = s + \frac{1}{\frac{1}{3}s + \frac{1}{s}}$



2nd. Zeros:  $\frac{6s + s^3}{3 + s^2} = \frac{2s}{3 + \frac{1}{2}s^2} = \frac{12/s}{\frac{6s + s^3}{s}}$

$\Rightarrow z = \frac{1}{2s} + \frac{1}{\frac{12}{s} + \frac{1}{2s}}$



3. Richards' sections

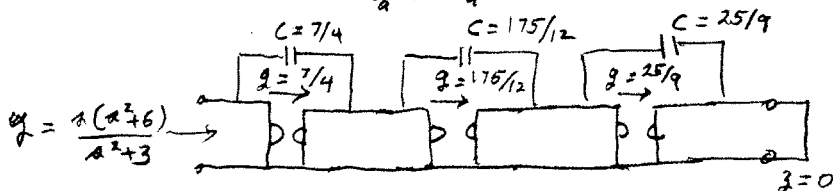
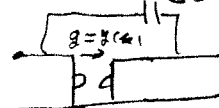
$y_2 = y(s) \left[ \frac{1/y(s) - 1/y_1(s)}{1/y_1(s) - 1/y(s)} \right]$ ,  $k=1$

$y_2(s) = \frac{1}{4} \left[ \frac{7/4 - \frac{s^2(s^2+6)}{s^2+3}}{\frac{s^2(s^2+6)}{s^2+3} - 7/4} \right] = \frac{1}{4} \left[ \frac{-s^4 - \frac{17}{4}s^2 + \frac{21}{4}}{s^2(-\frac{3}{4}s^2 + \frac{3}{4})} \right] = \frac{1}{4} \left[ \frac{-(s^2-1)(s^2+21/4)}{-3/4(s^2-1)} \right] = \frac{1}{3} (s^2+21/4)$

$y_{LL}(s) = \frac{1}{3} \cdot \frac{25}{4} \left[ \frac{\frac{7}{3} \cdot \frac{25}{4} - \frac{7}{3} (s^2+21/4)}{\frac{7}{3} (s^2+21/4) - \frac{7}{3} \cdot \frac{25}{4}} \right] = \frac{1}{3} \cdot \frac{25}{4} \left[ \frac{-s^2+1}{-\frac{21}{4}s^2 + \frac{21}{4}} \right] = \frac{1}{3} \cdot \frac{25}{4} \cdot \frac{4}{21} = \frac{25}{9}$

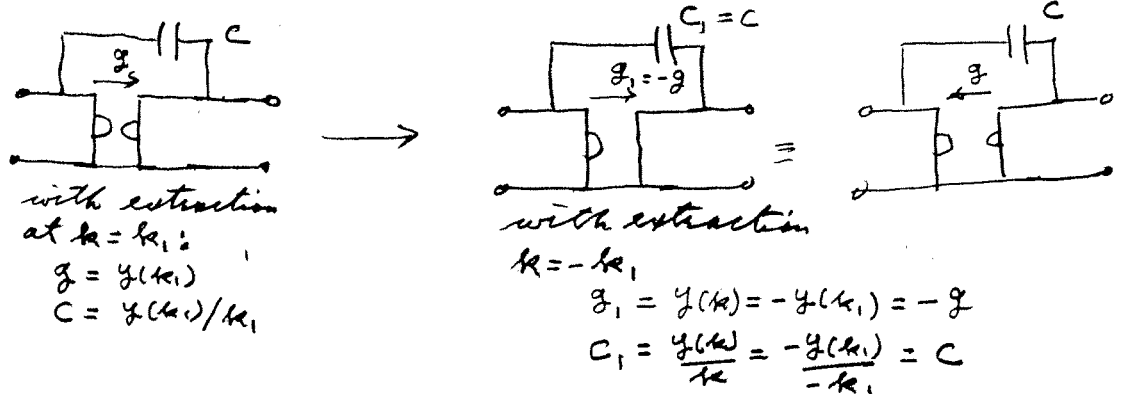
$y_{LLL}(s) = \frac{25}{9} \left[ \frac{25/9 - \frac{25}{9} s^2}{\frac{25}{9} s^2 - \frac{25}{9}} \right] \Rightarrow z_{LLL} = 0$

Each section



#2. a) as  $k$  is replaced by  $-k$ ,  $g_{old} = y(k) \rightarrow g_{new} = y(-k) = -y(k)$  so  $g_{new} = -g_{old}$   
 $g$  is replaced by its negative.  $C = y(k)/k \rightarrow C_{new} = \frac{y(-k)}{-k} = -\frac{y(k)}{-k} = C_{old}$

so the capacitor retains its old value and the gyrator is turned around



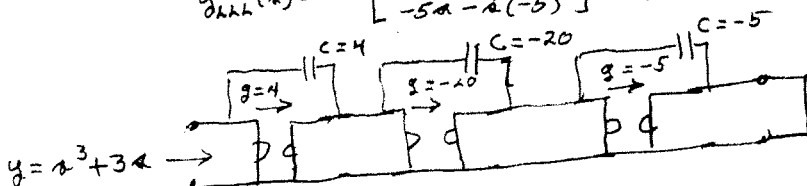
b) as each Richards' function extraction of an even part zero also extracts the negative of the zero, so each extraction removes 2 even part zeros. It also lowers the degree by one. so at a minimum  $2\delta[y(s)]$  even part zeros are removed. In the case of lossless  $y(s)$ , there are an infinite number of even part zeros, so no maximum.

c) as the Richards' function lowers the degree of a rational  $y(s)$  irrespective of it being PR it can be used for active synthesis (the C's in the sections will not all be  $> 0$ !) so  $g(s) = s^3 + 3s$ ,  $y(s) = -y(-s)$  so any three real  $k$ 's can be used. Choose  $k_1 = k_2 = k_3 = 1$

$$y_2(s) = 4 \left[ \frac{4 - s^2(s^2 + 3)}{s(s^2 + 3) - 4s} \right] = \frac{4}{s} \left[ \frac{-(s^2 - 1)(s^2 + 4)}{s^2 - 1} \right] = -\frac{4(s^2 + 4)}{s}$$

$$y_{2H}(s) = -20 \left[ \frac{-20 - (-4(s^2 + 4))}{-4(s^2 + 4) - s(-20)} \right] = -20s \left[ \frac{4 - 4s^2}{-16s^2 + 16} \right] = -5s$$

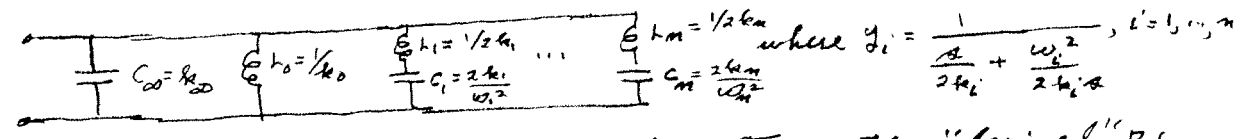
$$y_{2HL}(s) = -5 \left[ \frac{-5 - s(-5s)}{-5s - s(-5)} \right] = \frac{-25(s^2 - 1)}{0} \Rightarrow z_{2HL} = 0$$



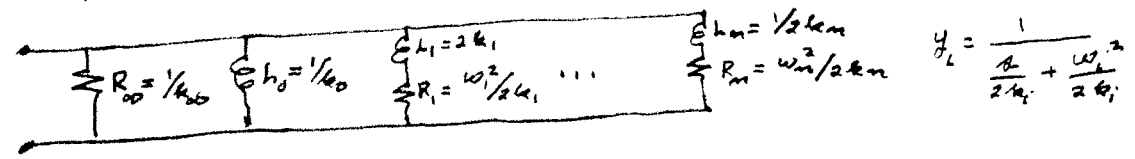
#3. From an LC circuit we know if  $y_{LC}(s)$  exists (i.e.  $\neq 0$ ), by 2nd Foster's,

$$y_{LC}(s) = k_{\infty} s + \frac{k_0}{s} + \sum_{i=1}^n \frac{2k_i s}{s^2 + \omega_i^2} \quad \text{all } k_i > 0, i=0, \dots, n, \infty$$

in which case the circuit is



Then for LR, replace each C by  $G = 1/R$ . Then the "derived" RL circuit is



This has

$$y_{LR}(s) = k_{\infty} + \frac{k_0}{s} + \sum_{i=1}^n \frac{2k_i s}{s^2 + \omega_i^2}$$

If we set  $s = p^2$  & multiply by  $p$  we get

$$p y_{LR}(p^2) = k_{\infty} p + \frac{k_0}{p} + \sum_{i=1}^n \frac{2k_i p}{p^2 + \omega_i^2} = y_{LC}(p)$$

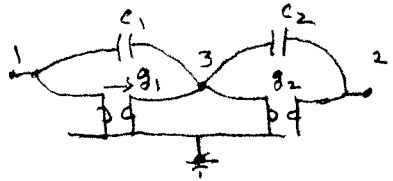
setting  $p$  back to  $s$ ,  $s y_{LR}(s^2) = y_{LC}(s)$

necessity  $\therefore s y_{LR}(s^2) = -[-s y_{LR}(1-s^2)]$   $\leftarrow$  necessarily true

sufficiency since any PR  $y(s)$  for which  $s y(s^2)$  is LC it can be synthesized by the above circuit the condition is sufficient

$\therefore$  a PR  $y(s)$  is for an LR circuit if and only if  $s y(s^2)$  is lossless.

#4,



$$Y_{\text{nodal}}^{(a)} = \begin{bmatrix} \kappa c_1 & 0 & -\kappa c_1 + g_1 \\ 0 & \kappa c_2 & -\kappa c_2 + g_2 \\ -\kappa c_1 - g_1 & -\kappa c_2 - g_2 & \kappa c_1 + \kappa c_2 \end{bmatrix}$$

By inspection of components attached to the nodes.

We wish to eliminate node 3  $\Rightarrow i_3 = 0 \Rightarrow$

$$\begin{bmatrix} -\kappa c_1 - g_1 & -\kappa c_2 - g_2 \end{bmatrix} \begin{bmatrix} v_1 \\ v_2 \end{bmatrix} = - \begin{bmatrix} \kappa c_1 + \kappa c_2 \end{bmatrix} v_3$$

$$\Rightarrow v_3 = \begin{bmatrix} \frac{\kappa c_1 + g_1}{\kappa c_1 + \kappa c_2} & \frac{\kappa c_2 + g_2}{\kappa c_1 + \kappa c_2} \end{bmatrix} \begin{bmatrix} v_1 \\ v_2 \end{bmatrix}$$

$$\therefore Y_{\text{2port}} = \begin{bmatrix} \kappa c_1 & 0 \\ 0 & \kappa c_2 \end{bmatrix} + \begin{bmatrix} -\kappa c_1 + g_1 \\ -\kappa c_2 + g_2 \end{bmatrix} \begin{bmatrix} \frac{\kappa c_1 + g_1}{\kappa c_1 + \kappa c_2} & \frac{\kappa c_2 + g_2}{\kappa c_1 + \kappa c_2} \end{bmatrix}$$

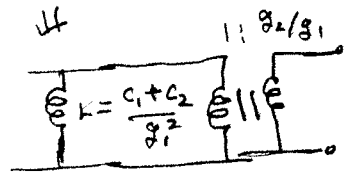
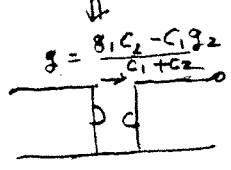
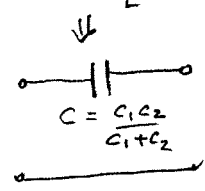
$$\kappa(c_1 + c_2) \cdot y_{11} = \kappa c_1 \cdot \kappa(c_1 + c_2) + (-\kappa c_1 + g_1)(\kappa c_1 + g_1) = \kappa^2 c_1 c_2 + g_1^2$$

$$\kappa(c_1 + c_2) \cdot y_{12} = (-\kappa c_1 + g_1)(\kappa c_2 + g_2) = -\kappa^2 c_1 c_2 + \kappa(g_1 c_2 - c_1 g_2) + g_1 g_2$$

$$\kappa(c_1 + c_2) \cdot y_{21} = (-\kappa c_2 + g_2)(\kappa c_1 + g_1) = -\kappa^2 c_1 c_2 - \kappa(g_1 c_2 - c_1 g_2) + g_1 g_2$$

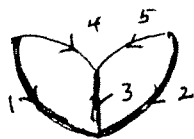
$$\kappa(c_1 + c_2) \cdot y_{22} = \kappa c_2 \cdot \kappa(c_1 + c_2) + (-\kappa c_2 + g_2)(\kappa c_2 + g_2) = \kappa^2 c_1 c_2 + g_2^2$$

$$\therefore Y_{\text{2port}}^{(a)} = \frac{\kappa c_1 c_2}{c_1 + c_2} \begin{bmatrix} 1 & -1 \\ -1 & 1 \end{bmatrix} + \frac{g_1 c_2 - c_1 g_2}{c_1 + c_2} \begin{bmatrix} 0 & 1 \\ -1 & 0 \end{bmatrix} + \frac{g_1^2}{\kappa(c_1 + c_2)} \begin{bmatrix} 1 & g_2/g_1 \\ g_2/g_1 & (g_2/g_1)^2 \end{bmatrix}$$



all in parallel

#5: a)



Cutset:  $Q_3 = \begin{bmatrix} 0 \\ 0 \\ 0 \end{bmatrix} = \begin{bmatrix} 1 & 0 & 0 & 1 & 0 \\ 0 & 1 & 0 & 0 & -1 \\ 0 & 0 & 1 & -1 & -1 \end{bmatrix} \begin{bmatrix} l_1 \\ l_2 \\ l_3 \\ l_4 \\ l_5 \end{bmatrix}$

$\Rightarrow e = \begin{bmatrix} 1 & 0 & 0 & 1 & 0 \\ 0 & 1 & 0 & 0 & -1 \\ 0 & 0 & 1 & -1 & -1 \end{bmatrix}$

Tie set:  $Q_2 = \begin{bmatrix} 0 \\ 0 \end{bmatrix} = \begin{bmatrix} -1 & 0 & 1 & 1 & 0 \\ 0 & -1 & 1 & 0 & 1 \end{bmatrix} \begin{bmatrix} v_1 \\ v_2 \\ v_3 \\ v_4 \\ v_5 \end{bmatrix}$

$\Rightarrow T = \begin{bmatrix} -1 & 0 & 1 & 1 & 0 \\ 0 & -1 & 1 & 0 & 1 \end{bmatrix}$

b)  $Av = Bl$  gives

$$\begin{bmatrix} 1 & 0 & 0 & 0 & 0 \\ 0 & 1 & 0 & 0 & 0 \\ 0 & 0 & 1 & 0 & 0 \\ 0 & 0 & 0 & 0 & g \\ 0 & 0 & 0 & -g & 0 \end{bmatrix} \begin{bmatrix} v_1 \\ v_2 \\ v_3 \\ v_4 \\ v_5 \end{bmatrix} = \begin{bmatrix} 0 & 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 & 0 \\ 0 & 0 & 2L & 0 & 0 \\ 0 & 0 & 0 & 1 & 0 \\ 0 & 0 & 0 & 0 & 1 \end{bmatrix} \begin{bmatrix} l_1 \\ l_2 \\ l_3 \\ l_4 \\ l_5 \end{bmatrix}$$

$v_b = v + e, e = \begin{bmatrix} v_1 \\ v_2 \\ 0 \\ 0 \\ 0 \end{bmatrix}$   
 $l_b = l$

$l_b = T^T l_r = l = \begin{bmatrix} -1 & 0 \\ 0 & -1 \\ 1 & 1 \\ 0 & 1 \end{bmatrix} \begin{bmatrix} l_4 \\ l_5 \end{bmatrix}$

use  $v_b = e^T v_r = v + e \Rightarrow v = e^T v_r - e$

$\therefore Ae^T v_r - Ae = \begin{bmatrix} 1 & 1 & 1 & 0 & 0 \\ 0 & 0 & 0 & 1 & 0 \\ 0 & 0 & 0 & 0 & 1 \\ 0 & 0 & 0 & -g & 0 \end{bmatrix} \begin{bmatrix} v_1 \\ v_2 \\ v_3 \\ v_4 \\ v_5 \end{bmatrix} - \begin{bmatrix} 1 & 0 & 0 & 0 & 0 \\ 0 & 1 & 0 & 0 & 0 \\ 0 & 0 & 1 & 0 & 0 \\ 0 & 0 & 0 & 0 & g \\ 0 & 0 & 0 & -g & 0 \end{bmatrix} \begin{bmatrix} v_1 \\ v_2 \\ 0 \\ 0 \\ 0 \end{bmatrix}$

$= \begin{bmatrix} 1 & 0 & 0 & 0 & 0 \\ 0 & 1 & 0 & 0 & 0 \\ 0 & 0 & 1 & 0 & 0 \\ 0 & 0 & 0 & 1 & 0 \\ -g & 0 & -g & 0 & 0 \end{bmatrix} v_r - \begin{bmatrix} v_1 \\ v_2 \\ 0 \\ 0 \\ 0 \end{bmatrix}$

$B^T l_r = \begin{bmatrix} 0 & 0 & 2L & 1 & 0 \\ 0 & 0 & 0 & 0 & 1 \end{bmatrix} \begin{bmatrix} l_4 \\ l_5 \end{bmatrix} = \begin{bmatrix} 0 & 0 \\ 0 & 0 \\ 2L & 2L \\ 1 & 0 \\ 0 & 1 \end{bmatrix} l_r$

$\therefore [Ae^T, -B^T] x = \begin{bmatrix} 1 & 0 & 0 & 0 & 0 & 0 & 0 \\ 0 & 1 & 0 & 0 & 0 & 0 & 0 \\ 0 & 0 & 1 & 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & 1 & 0 & -2L & -2L \\ 0 & 0 & 0 & 0 & 1 & 0 & 0 \\ -g & 0 & -g & 0 & 0 & 0 & -1 \end{bmatrix} x = \begin{bmatrix} 1 & 0 \\ 0 & 0 \\ 0 & 0 \\ 0 & 0 \\ 0 & 0 \\ 0 & 0 \end{bmatrix} u, x = \begin{bmatrix} v_1 \\ v_2 \\ v_3 \\ l_4 \\ l_5 \end{bmatrix} = \begin{bmatrix} v_r \\ l_r \end{bmatrix}$

$y = \begin{bmatrix} -1 \\ -1 \end{bmatrix} = \begin{bmatrix} 0 & 0 & 0 & -1 & 0 \\ 0 & 0 & 0 & 0 & -1 \end{bmatrix} x$  using  $l_b = T^T l_r$

$\therefore$  semistate equations

$$\begin{bmatrix} 0 & 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 & 0 \end{bmatrix} \frac{dx}{dt} = \begin{bmatrix} 1 & 0 & 0 & 0 & 0 \\ 0 & 1 & 0 & 0 & 0 \\ 0 & 0 & 1 & 0 & 0 \\ 0 & 0 & 0 & 1 & 0 \\ -g & 0 & -g & 0 & -1 \end{bmatrix} x + \begin{bmatrix} -1 & 0 \\ 0 & -1 \\ 0 & 0 \\ 0 & 0 \\ 0 & 0 \end{bmatrix} u$$

$y = \begin{bmatrix} 0 & 0 & 0 & -1 & 0 \\ 0 & 0 & 0 & 0 & -1 \end{bmatrix} x$