

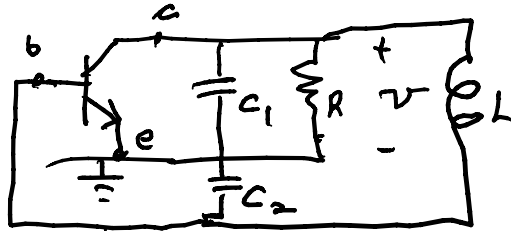
Colpitts Oscillator

pp. 1179-1182

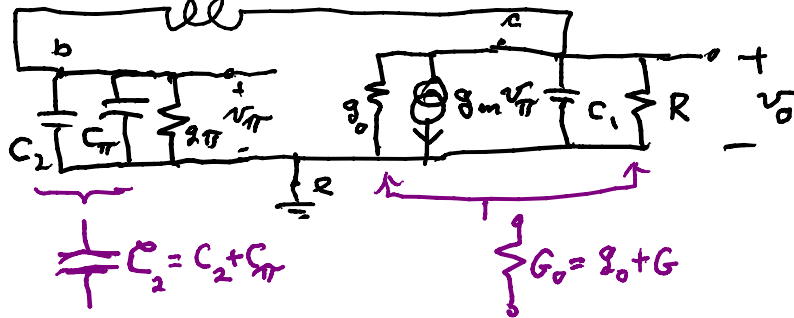
" ac circuit

EE303
Tu 10/07/08

RFC = radio
frequency
choke



$i_L = \frac{1}{sL} (v_o - v_\pi)$ \downarrow small signal equivalent circuit p. 1180



KVL
KCL
laws of devices

KCL @ b

$$i_L = sC_2 v_\pi + g_\pi v_\pi = \frac{1}{sL} (v_o - v_\pi)$$

$$[1 + sL(\alpha C_2 + g_\pi)] v_\pi = v_o$$

2 eqs
in
2
unkns.

$$\text{KCL @ c: } -i_L = G_o v_o + \alpha C_1 v_o + g_m v_\pi$$

$$= -(\alpha C_2 + g_\pi) v_\pi$$

$$(-g_m - g_\pi - \alpha C_2) v_\pi = (G_o + \alpha C_1) v_o$$

$$-\frac{(g_m + g_\pi + \alpha C_2) \cdot v_o}{[1 + sL(\alpha C_2 + g_\pi)]} = (G_o + \alpha C_1) v_o$$

$$\left\{ [G_o + \alpha C_1] [1 + sL(\alpha C_2 + g_\pi)] + (g_m + g_\pi + \alpha C_2) \right\} v_o = 0$$

$$\left\{ \begin{array}{l} G_o + G_o s^2 L^2 C_2 + G_o sL g_\pi \\ + \alpha C_1 + s^3 C_1 L C_2 + s^2 L C_1 g_\pi \\ + \alpha C_2 + (g_m + g_\pi) \end{array} \right\} v_o = 0$$

$$\{C_1 C_2 L s^3 + [G_0 L C_2 + L C_1 g_m] s^2 + [G_0 L g_m + C_1 + C_2] s + [G_0 + g_m + g_m]\} v_o = 0$$

Let a solution $v_o(t) = V_o e^{st}$ if $\{ \cdot \} = 0$

$$P(s) = s^3 + \frac{1}{L C_1 C_2} [G_0 L C_2 + g_m C_1] s^2 + \frac{1}{L C_1 C_2} [G_0 L g_m + C_1 + C_2] s + \frac{[G_0 + g_m + g_m]}{L C_1 C_2}$$

$P(s), v_o(t) = 0$ [use initial condition to work]

desire s such that $P(s) = 0$

Here are designing to get sinusoidal outputs

$$V_o e^{j\omega t} = v_o (\cos \omega t + j \sin \omega t)$$

\therefore desire $P(j\omega_0) = 0$

$$1) -j\omega^3 + j \frac{1}{L C_1 C_2} [G_0 L g_m + C_1 + C_2] \omega = 0 \quad (\text{set imaginary \& real part to } 0)$$

$$2) \frac{-1}{C_1 C_2} [G_0 L C_2 + g_m C_1] \omega^2 + \frac{[G_0 + g_m + g_m]}{L C_1 C_2} = 0$$

from 1) $\omega^2 = \frac{[G_0 L g_m + C_1 + C_2]}{L C_1 C_2}$ if $g_m = 0$

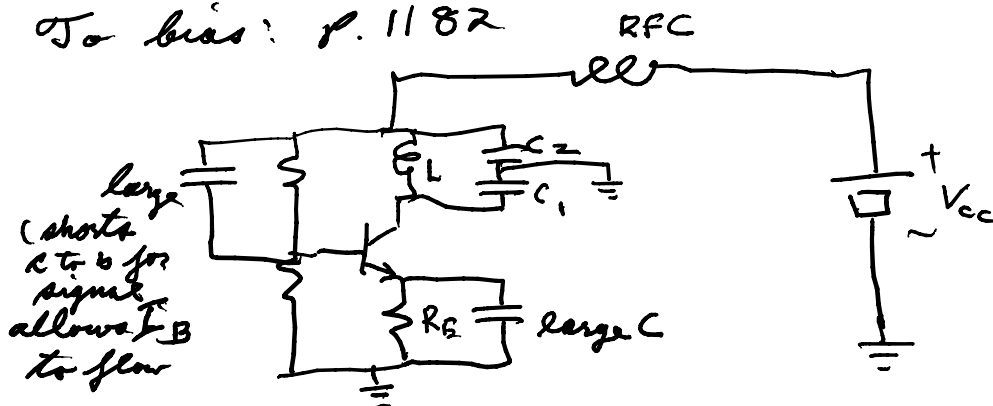
$$\omega_0 = \sqrt{\frac{C_1 + C_2}{C_1 C_2 L}}$$

from 2) $G_0 + g_m + g_m = L [G_0 L C_2 + g_m C_1] \omega^2$

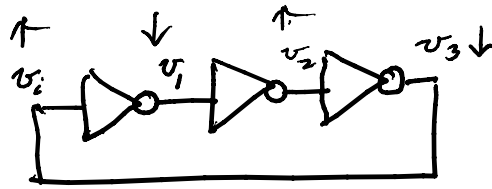
$$= \sqrt{\frac{1}{L C_1}}$$

$$\Rightarrow g_m = L [G_0 L C_2 + g_m C_1] \left[\frac{G_0 L g_m + C_1 + C_2}{L C_1 C_2} \right] - G_0 - g_m$$

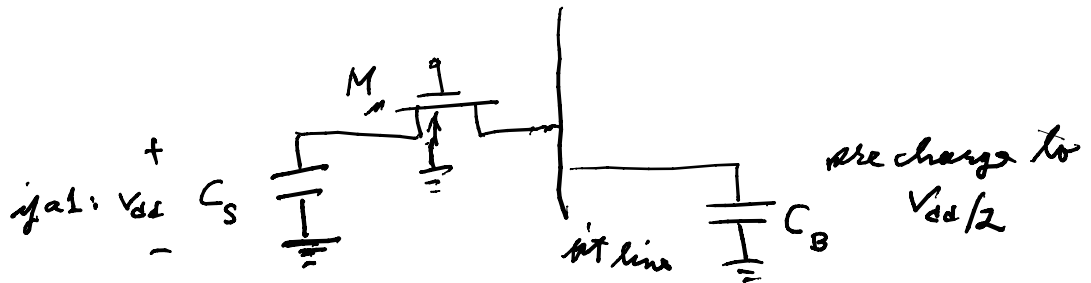
To bias: p. 1182



Ring oscillators



Dynamic Memory P. 1036



$$Q_S = C_S V_{dd} \quad Q_B = C_B \cdot V_{dd}/2$$

after connect with a readout pulse on the sense transistor!

$$V_B = \frac{Q_{SB}}{C_S + C_B} =$$

$$Q_{SB} = C_S V_{dd} + C_B V_{dd}/2 = \text{total charge, preserved before \& after connect}$$

after connect

$$V_B = \frac{1}{C_S + C_B} [C_S + C_B/2] V_{dd}$$

$$\begin{aligned} -\Delta V_{on C_B} &= \frac{V_{dd}}{2} - \frac{C_S + C_B/2}{C_S + C_B} V_{dd} = \frac{\left(\frac{C_S + \cancel{C_B/2}}{2} - C_S - \frac{\cancel{C_B}}{2}\right) V_{dd}}{C_S + C_B} \\ &= -\frac{C_S/2}{C_S + C_B} V_{dd} \end{aligned}$$

here $C_B \gg C_S \Rightarrow \Delta V_{on C_B} = +\frac{C_S}{2C_B} V_{dd}$

sign so $V_{C_B \text{ new}} = V_{C_B \text{ old}} + \Delta V_{C_B}$

