

$$P_{ave}(j\omega) = \operatorname{Re}(I^{*T}V) \Big|_{s=j\omega} \quad * = j \rightarrow -j$$

$$= \frac{(I^{*T}V + (I^{*T}V)^*)}{2} \Big|_{s=j\omega} = \frac{I^{*T}V + (I^T V^*)}{2} \Big|_{s=j\omega}$$

but $I = YV$

$$= \frac{(YV)^{*T} + ((YV)^T V^*)^T}{2} \Big|_{s=j\omega}$$

$$= \frac{V^{*T} Y^{*T} V + V^{*T} (Y)V}{2} \Big|_{s=j\omega} = \frac{V^{*T} [Y^{*T} + Y] V}{2} \Big|_{s=j\omega}$$

= 0 if lossless for all vectors V complex

$Y(j\omega)^{*T} + Y(j\omega) = 0$ matrix if $Y(s)$ is rational with real coefficients

$$= \left[Y(-s)^T + Y(s) \right] \Big|_{s=j\omega} = 0 \text{ matrix} \quad \text{then } Y(j\omega)^* = Y(-j\omega) = Y(-s) \Big|_{s=j\omega}$$

if stable $Y(s)$ is analytic in $\operatorname{Re} s > 0$ & here

$Y(-s)^T + Y(s)$ is zero on an infinite set & if $Y(s)$ is rational, by complex variable theory this sum is zero for all s .

$$Y(-s)^T + Y(s) \equiv 0 \quad \text{for all } s \leftarrow \text{lossless condition}$$

If $Y(s)$ is $1 \times 1 \Rightarrow$ driving point $Y(s) = y(s)$

$$2\operatorname{Re}(y(s)) = y(s) + y(-s) \equiv 0$$

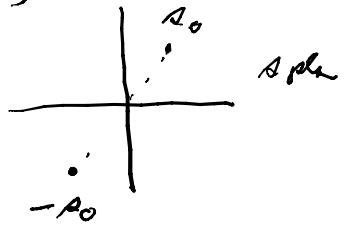
Even

ex: $y(s) = \frac{2s}{s^2 + 7} j$

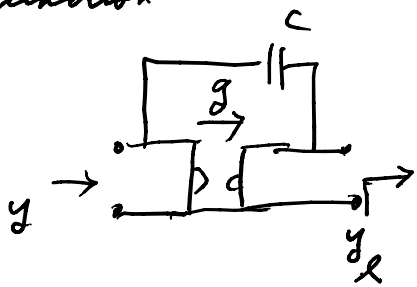
$$2\operatorname{Re} y(s) = y(s) + y(-s) = y + y^*$$

$$= \frac{2s}{s^2 + 7} + \frac{(-2s)}{(-s)^2 + 7} = 0$$

$s \rightarrow -s =$ Hurwitz conjugate



We can synthesize a lossless $y(s)$ by the Richards' function



$$y_L = \frac{1 - \left(\frac{C\alpha}{g}\right) \left(\frac{y_{in}(\alpha)}{g}\right)}{\left(\frac{y_{in}(\alpha)}{g}\right) - \left(\frac{C\alpha}{g}\right)} \quad \text{from 09/10/07}$$

$$\begin{aligned} \text{Y.361 } R(\alpha) &= \frac{kz(0) - \alpha z(k)}{kz(k) - \alpha z(0)} = \frac{ky(k) - \alpha y(0)}{ky(0) - \alpha y(k)} \\ &= \frac{\frac{1}{k} \frac{1}{y(0)} - \frac{\alpha}{k} \frac{1}{y(k)}}{\frac{1}{k} \frac{1}{y(k)} - \frac{\alpha}{k} \frac{1}{y(0)}} = \frac{ky(k) \left(1 - \frac{\alpha}{k} \frac{y(0)}{y(k)}\right)}{ky(0) \left(\frac{y(0)}{y(k)} - \frac{\alpha}{k}\right)} = \frac{1 - \left(\frac{\alpha}{k}\right) \left(\frac{y(0)}{y(k)}\right)}{\left(\frac{y(0)}{y(k)}\right) - \left(\frac{\alpha}{k}\right)} \end{aligned}$$

choose $g = y(k)$, $\frac{\alpha}{k} = \frac{C\alpha}{g} \Rightarrow \frac{1}{k} = \frac{C}{y(k)} \Rightarrow C = \frac{y(k)}{R}$

$\alpha = k$ gives a pole & zero cancellation & can get another
if $\alpha = -k$ also can be used to cancel
choose k to be a zero of $\text{Ev } y(s)$, then $y(-\alpha) = -y(\alpha)$

$$ky(k) - \alpha y(\alpha) \Big|_{\alpha = -k} = ky(k) - (-k)(+y(-k)) \text{ if } \alpha = -k \text{ is a zero of Ev } y(s)$$

$$\alpha = -k = k[y(k) + y(-k)] = 0 \quad \uparrow$$

$$\text{two } ky(\alpha) - \alpha y(k) \Big|_{\alpha = -k} = ky(-k) - (-k)y(k) = k[y(-k) + y(k)] = 0 \text{ if } k = \text{a zero of Ev } y(s)$$

But if $y(s)$ is lossless any real positive k is a zero of $\text{Ev } y(s)$.

Ex: $y(s) = \frac{2s}{s^2 + 7}$ choose any pos. real k , say $k = 2$

$$y(k) = y(2) = \frac{4}{4+7} = 4/11$$

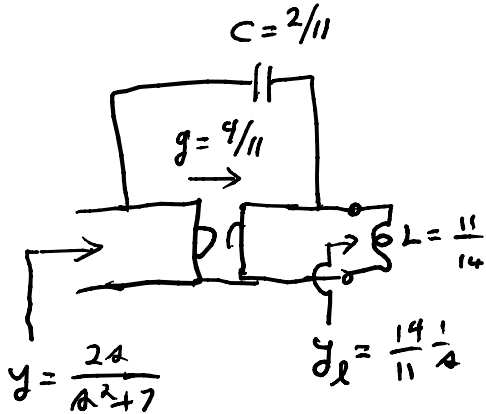
$$g = y(k) = 4/11 \quad C = \frac{y(k)}{k} = \frac{4/11}{2} = 2/11$$

$$\frac{y_2(z)}{z} = \frac{ky_2(k) - az_2(z)}{ky_2(z) - az_2(k)} = \frac{2 \times \frac{4}{11} - a \frac{2a}{a^2+7}}{2 \times \frac{2a}{a^2+7} - a \times \frac{4}{11}} = \frac{\frac{8}{11}a^2 + \frac{7 \times 8}{11} - 2a^2}{4a - \frac{4a^3}{11} - \frac{28a}{11}}$$

we know $(a-2)(a-(-2)) = a^2 - 4$ cancels in numerator and denominator.

$$\frac{y_2(z)}{z} = \frac{(8-22)a^2 + 7 \times 8}{-4a^3 + (44-28)a} = \frac{1}{a} \times \frac{-14 \left[a^2 - \frac{56}{14} \right]}{-4 \left[a^2 - \frac{16}{4} \right]} = \frac{1}{a} \cdot \left(\frac{7}{2} \right)$$

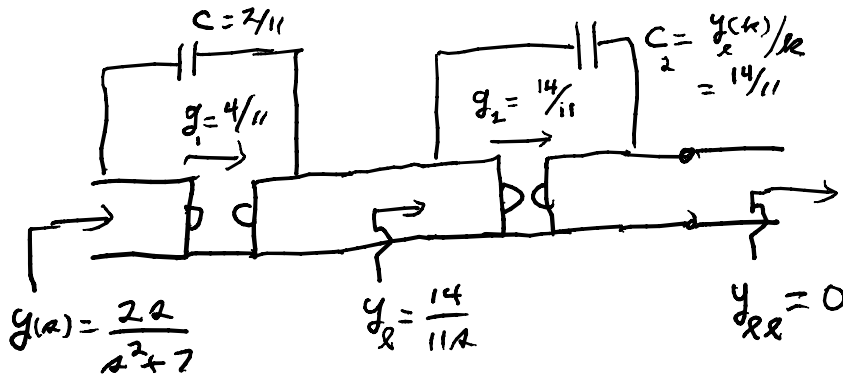
$$y_2(z) = \frac{4}{11} \times \frac{7}{2} \cdot \frac{1}{z} = \frac{14}{11} \cdot \frac{1}{z} \quad \left. \begin{array}{l} \text{ } \\ \text{ } \end{array} \right\} L = \frac{11}{14}$$



$y_2(z) = \frac{14}{11a}$, by Richards' function, choose a real, say $a=1$

$$\frac{y_{22}(z)}{y_2(z)} = \frac{ky_2(k) - az_2(z)}{ky_2(z) - az_2(k)} = \frac{1 \cdot \frac{14}{11} - \frac{14}{11}}{1 \cdot \frac{14}{11a} - a \cdot \frac{14}{11}} \equiv 0 \Rightarrow \text{open circuit}$$

as $i = yv = 0$



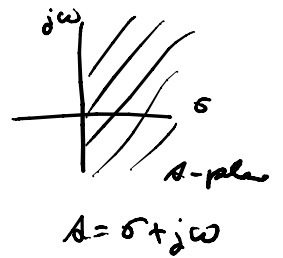
Positive-real
function
(for $Y(s)$ & $Z(s)$)

$f(s)$

1) analytic in $\text{Re } s > 0$
(stable circuit)

2) $f(s)$ is real
in $\text{Re } s > 0$
(real components)

3) $\text{Re } f(s) \geq 0$ in $\text{Re } s > 0$
(passive)



importance: any passive (real) circuit has $Y(s)$ & $Z(s)$
positive real