

Cavities

ENEE 681

Role of Cavities

Cavities are resonant structures: Support EM modes at specific frequencies.

Used in:

- Filters

- Oscillators

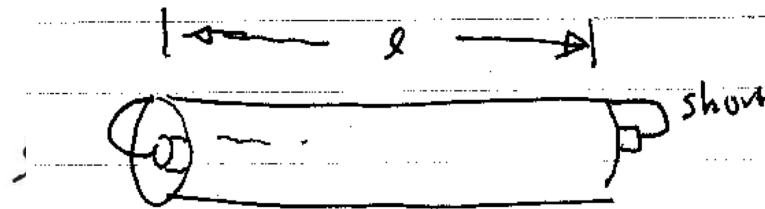
- Amplifiers

- Measurement of material properties

Resonance

natural frequency of oscillation of fields

Example:



Transmission line

Terminated in short circuits

$$kl = n\pi$$

$$\cancel{\omega = kv} \quad k = \frac{\omega}{v}$$

$$\omega_n = \frac{n\pi v}{l}$$

$$n = 1, 2, \dots$$

$$\frac{d^2}{dx^2} V(x) + \frac{\omega^2}{v^2} V(x) = 0$$

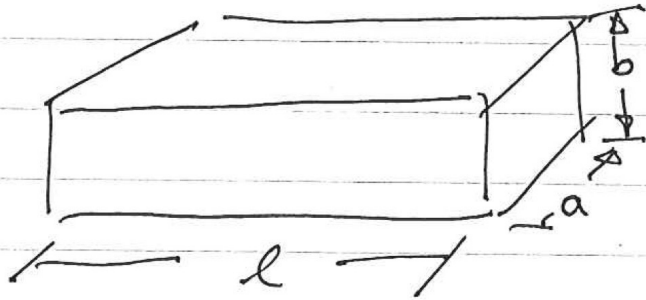
$$V(x) = V_+ e^{i\omega x/v} + V_- e^{-i\omega x/v}$$

$$V(x=0) = 0$$

$$V(x=l) = 0$$

Enclosed Rectangular Prism

~~enclosed~~ enclosed box

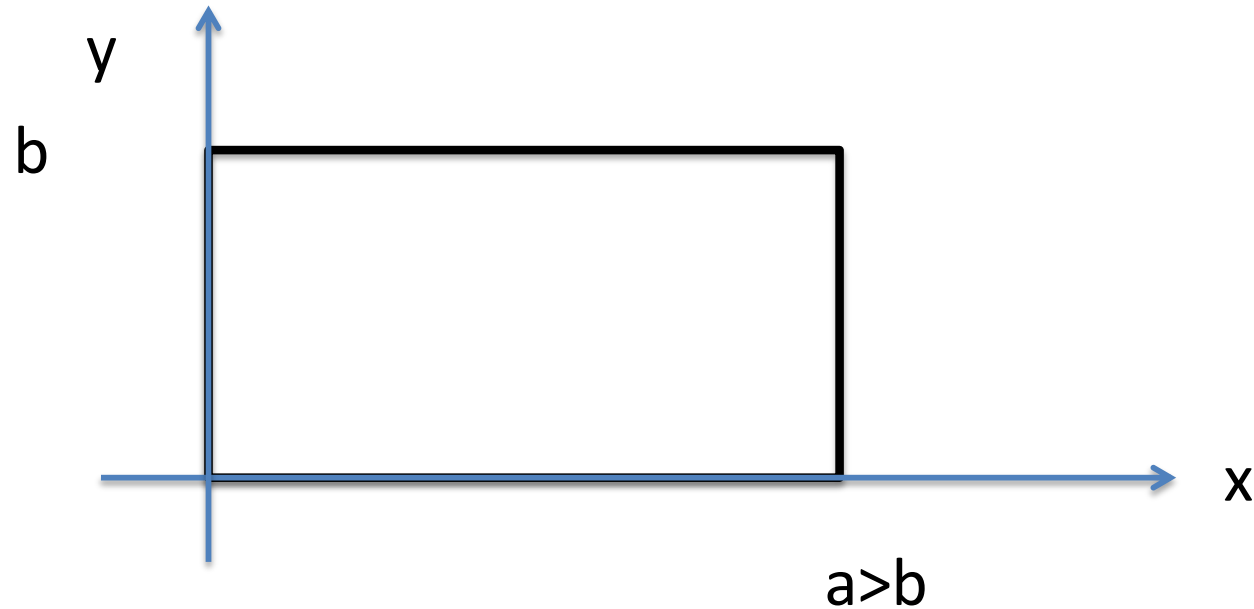


TE_{nm} or TM_{nm}

$$\omega_{nmp} = V \sqrt{\left(\frac{n\pi}{a}\right)^2 + \left(\frac{m\pi}{b}\right)^2 + \left(\frac{p\pi}{l}\right)^2}$$

n, m, p
are
integers

Modes of a Rectangular WG



$$\text{TM}_{nm}: \hat{E}_z = E_0 \sin(k_x x) \sin(k_y y) \quad k_x = \frac{n\pi}{a}, \quad k_y = \frac{m\pi}{b}: \quad n, m = 1, 2, 3, \dots$$

$$\text{TE}_{nm}: \hat{H}_z = H_0 \cos(k_x x) \cos(k_y y) \quad k_x = \frac{n\pi}{a}, \quad k_y = \frac{m\pi}{b}: \quad n, m = 0^*, 1, 2, 3, \dots$$

* one or the other, but not both

Cut-Off frequencies

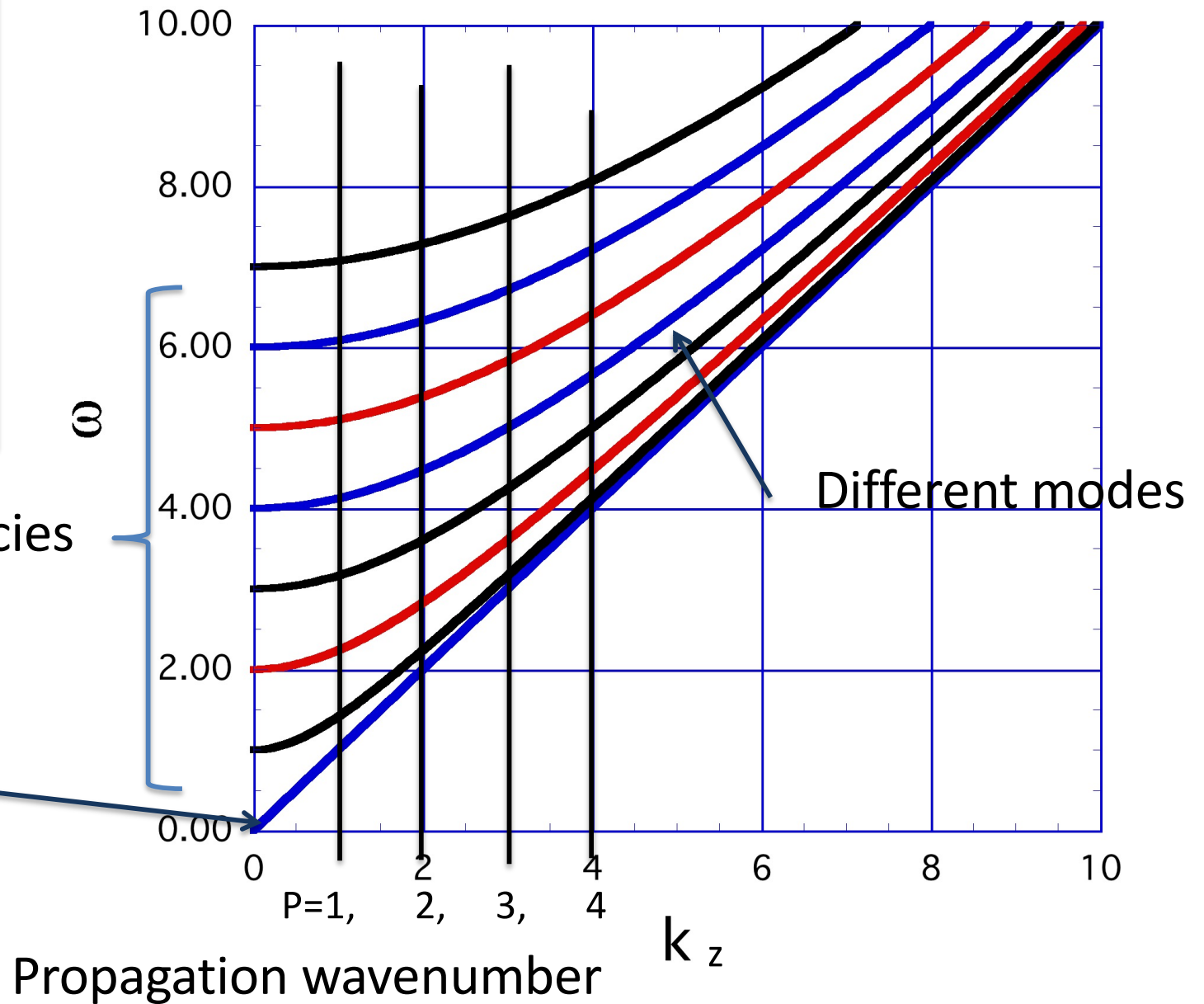
$$\omega_{c,n,m} = v \sqrt{\left(\frac{n\pi}{a}\right)^2 + \left(\frac{m\pi}{b}\right)^2}$$

WG Dispersion Relations

$$\frac{\omega^2}{v^2} = k_{\perp}^2 + k_z^2$$
$$k_z = p \frac{\pi}{L}$$

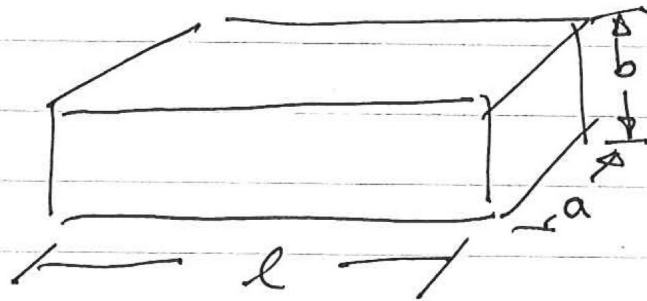
Cut off frequencies

Transmission
Lines Only



Enclosed Rectangular Prism

~~an~~ enclosed box



TE_{nm} or TM_{nm}

$$\omega_{nmp} = v \sqrt{\left(\frac{n\pi}{a}\right)^2 + \left(\frac{m\pi}{b}\right)^2 + \left(\frac{p\pi}{l}\right)^2}$$

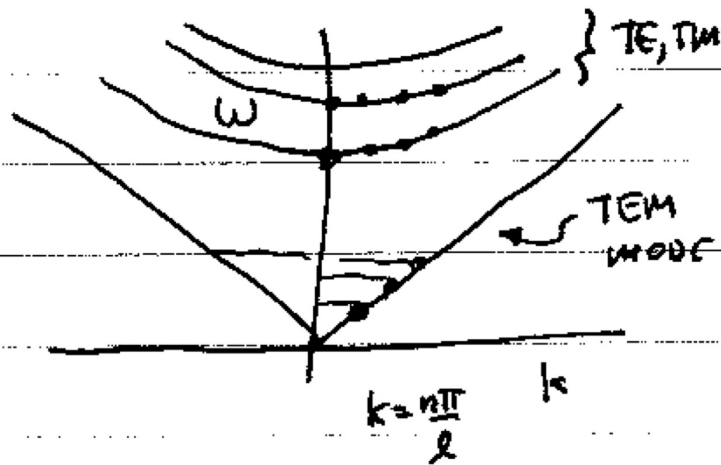
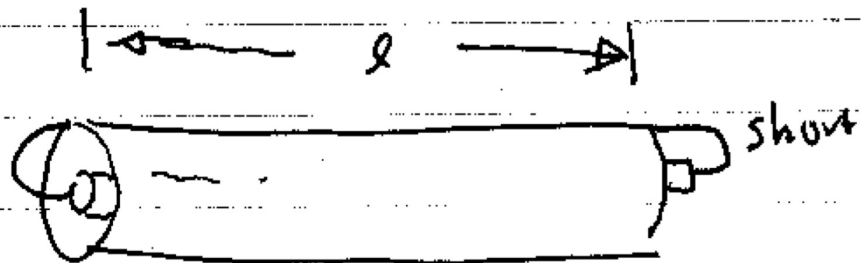
n, m, p
are
integers

$$\omega^2 = k_z^2 v^2 + \omega_{c,nm}^2$$

$$\omega_{c,nm} = v \sqrt{\left(\frac{n\pi}{a}\right)^2 + \left(\frac{m\pi}{b}\right)^2}$$

$$k_z = p \frac{\pi}{L}$$

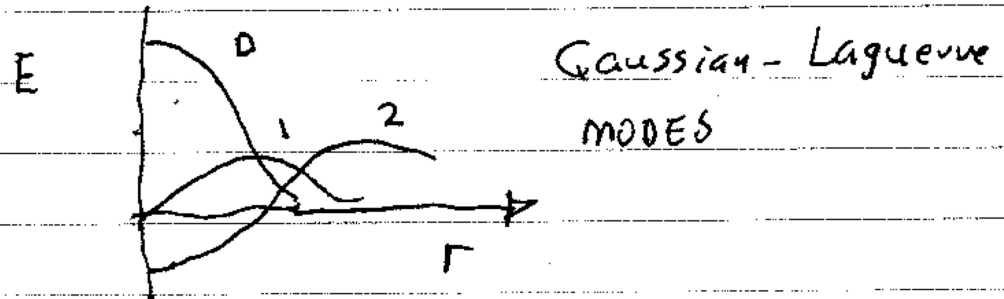
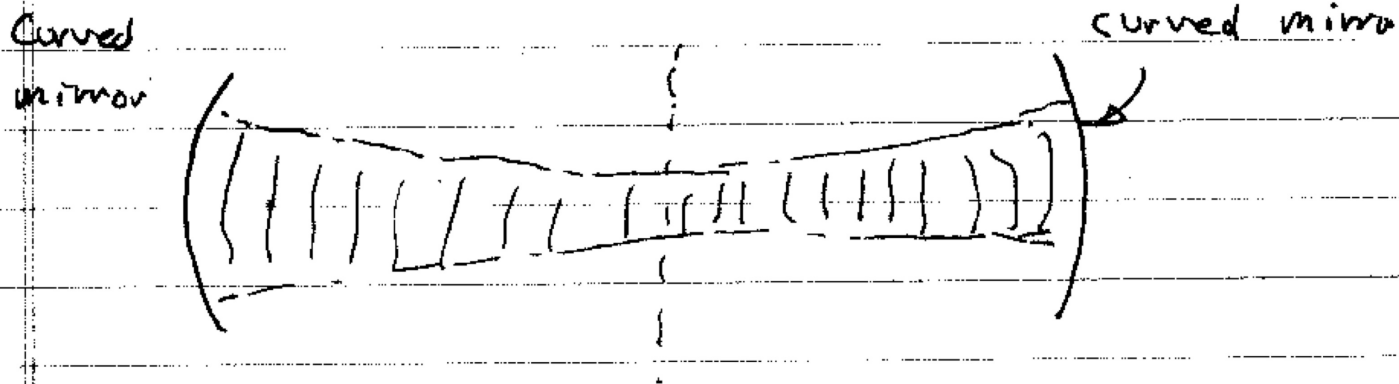
Transmission Line – TEM mode



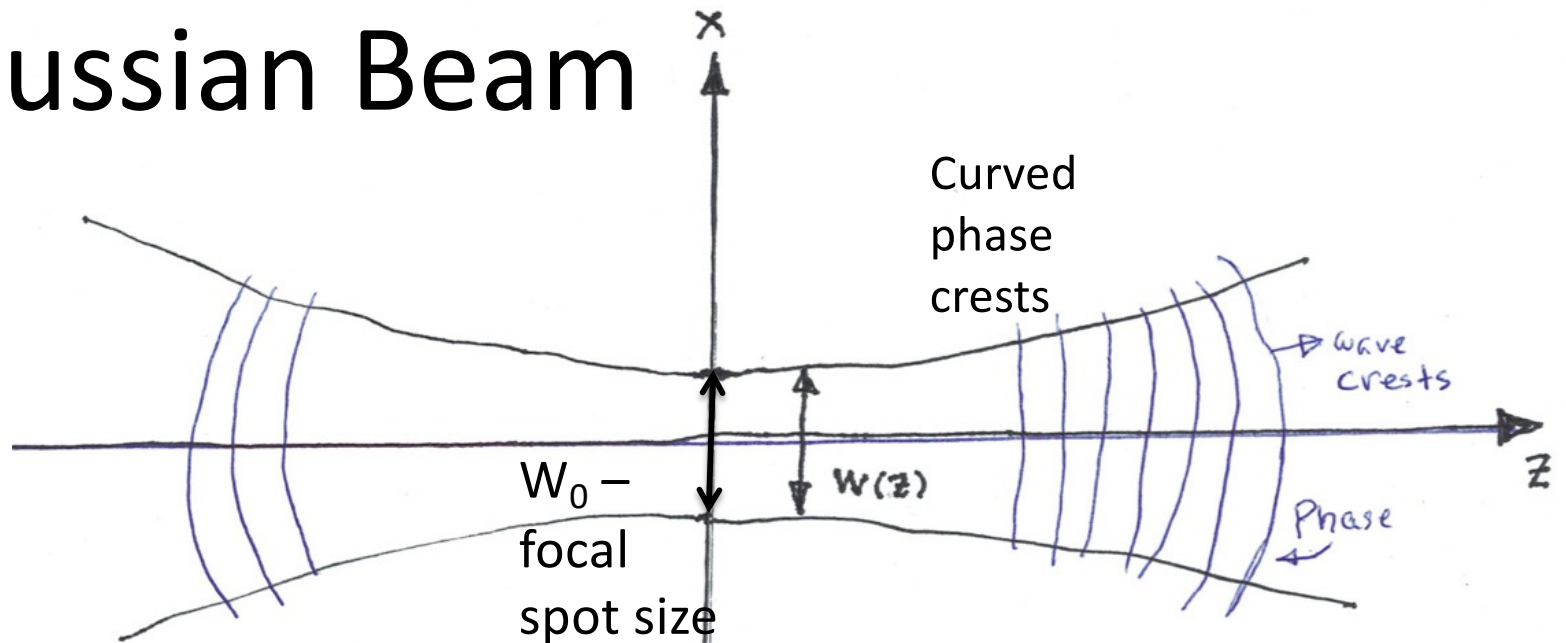
Operate at frequencies well below TE and TM cut-off

Fabry-Perot Cavity

Fabry-Perot Cavity



Gaussian Beam



$$E_{x,y}(x,y,z) = \frac{E_0}{1 + iz/Z_R} \exp \left[-\frac{(x^2 + y^2)}{W_0^2 (1 + iz/Z_R)} + ikz \right]$$

$$W(z) = W_0 \sqrt{1 + z^2/Z_R^2} \quad Z_R = \frac{1}{2} kW_0^2 \quad \text{Rayleigh Length}$$

$$E_{x,y}(0,0,z) = \frac{E_0}{1 + iz/Z_R} \exp[+ikz] = \frac{E_0}{\sqrt{1 + (z/Z_R)^2}} \exp[+ikz + i\phi(z)]$$

$$\text{Guoy Phase} \quad \tan \phi = -z/Z_R$$

Gaussian Beam

$$E_{x,y}(x,y,z) = \frac{E_0}{1 + iz/Z_R} \exp \left[-\frac{(x^2 + y^2)}{W_0^2 (1 + iz/Z_R)} + ikz \right]$$

$$= \underbrace{\frac{E_0}{\sqrt{1 + z^2/Z_R^2}} \exp \left[-\frac{(x^2 + y^2)}{W_0^2 (1 + z^2/Z_R^2)} \right]}_{\text{Amplitude}} \exp \left\{ i \underbrace{\left[kz + \frac{z(x^2 + y^2)}{Z_R W_0^2 (1 + z^2/Z_R^2)} + \phi_G \right]}_{\text{Phase}} \right\}$$

Amplitude

Phase

Pick parameters such that phase is constant on surface of mirror.

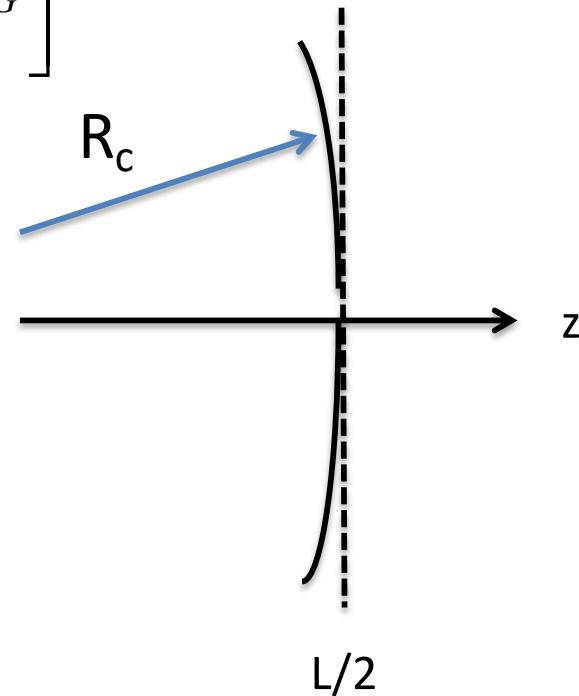
The pick k such that the phase changes by $p\pi$ in going from one mirror to the next

Wavebeam phase $\left[kz + \frac{z(x^2 + y^2)}{Z_R W_0^2 (1 + z^2 / Z_R^2)} + \phi_G \right]$

Surface of mirror $z \simeq \frac{L}{2} - \frac{(x^2 + y^2)}{2R_c}$

Will match if $\frac{L}{2R_c} = \frac{(L/2)^2}{Z_R^2 + (L/2)^2} < 1$

Phase change mirror to mirror $2 \left(k_p \frac{L}{2} + \phi_G \right) = p\pi$



For a given L and R_c Z_R is determined above.
Hence W_0 focal spot determined.

$$Z_R = \frac{1}{2} k W_0^2$$

Design a Fabrey-Perot resonator

Requirements:

Wavelength 1 micron = 10^{-6} m.

Focal spot size = 100 microns = 10^{-4} m.

Spot size on mirrors = 300 microns = 3×10^{-4} m

Find L and R_c

Super bonus: How big must the mirrors be to keep “spill over” below 10%

Quality Factor

Quality factor measures losses Q

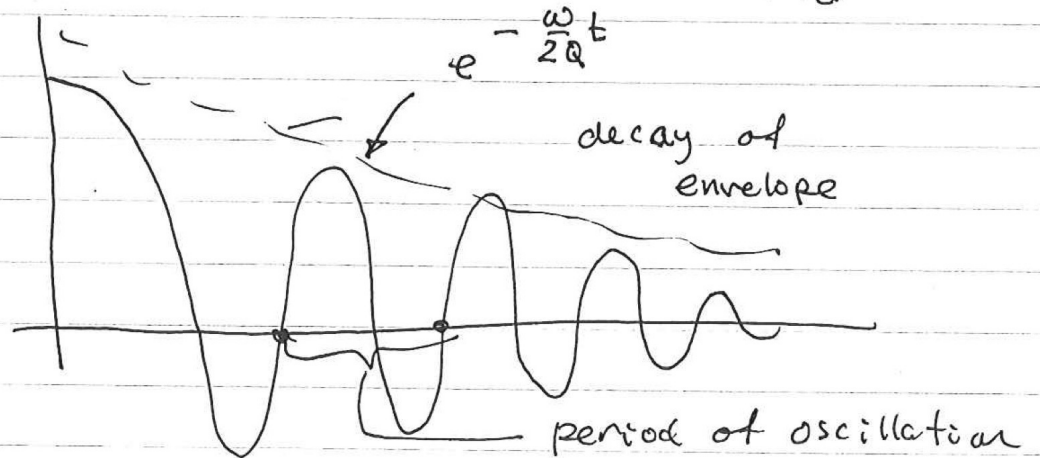
big Q low loss low Q high loss

Time domain

$$P_{\text{in}} \left(\frac{\omega U}{P_d} \right) = Q$$

$V(t)$

$E(t)$



$$\frac{1}{Q} \equiv \frac{\text{Power Dissipated}}{\omega \text{ Energy Stored}}$$

Field decay rate

$$E, H \sim \exp\left(-\frac{\omega}{2Q}t\right)$$

Frequency Domain

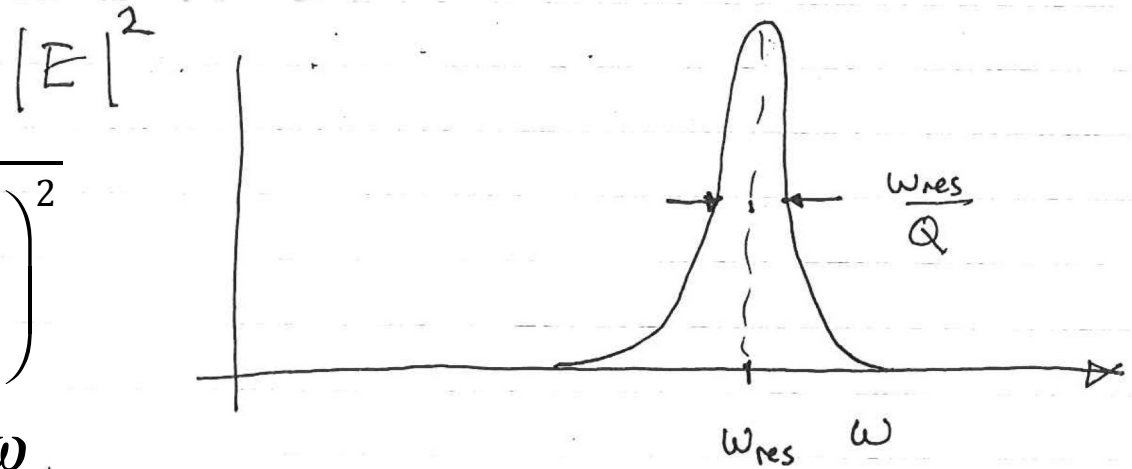
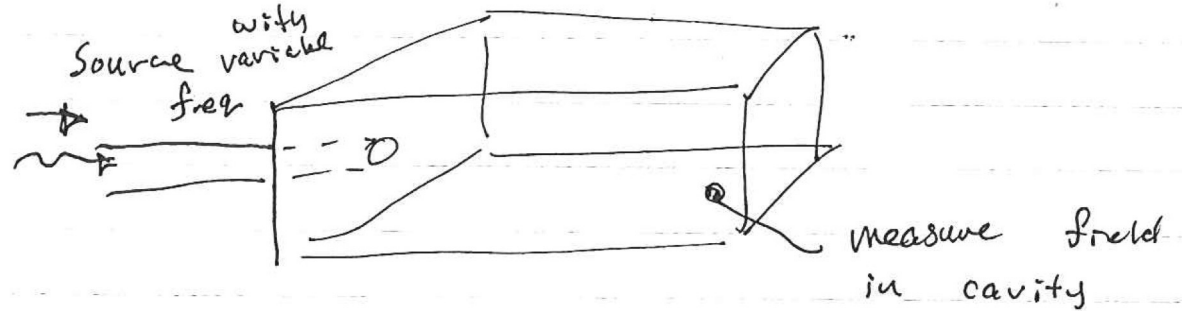
Steady state field response

$$E, H \sim \frac{\text{Source}}{\omega - \omega_{res} + i \frac{\omega_{res}}{2Q}}$$

$$|E|^2, |H|^2 \sim \frac{1}{(\omega - \omega_{res})^2 + \left(\frac{\omega_{res}}{2Q}\right)^2}$$

$$\text{Half maximum } \omega - \omega_{res} = \pm \frac{\omega_{res}}{2Q}$$

$$\text{Full Width at Half Maximum (FWHM)} = \frac{\omega_{res}}{Q}$$



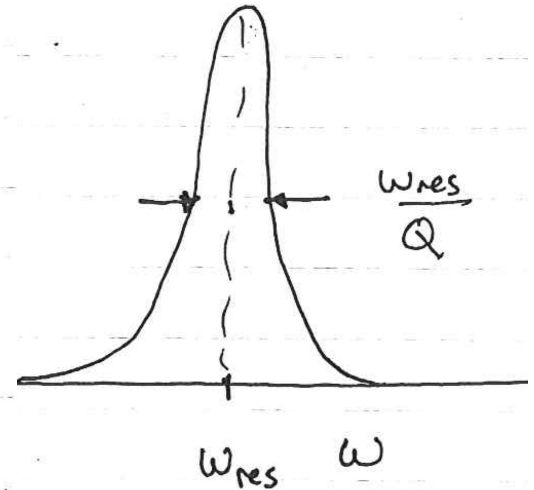
Response Function

$$|E|^2 \propto$$

$$\frac{1}{(\omega - \omega_{res})^2 + \left(\frac{\omega_{res}}{2Q}\right)^2}$$

when $\omega = \omega_{res} \pm \frac{\omega_{res}}{2Q}$

$|E|^2$ is $\frac{1}{2}$ of peak value



Full width at half max of Power
(FWHM) $\rightarrow \frac{\omega_{res}}{Q}$

Multiple Contributions to Loss

if. losses are small ($Q \gg 1$)

losses are additive

different
loss mechanism

$$\frac{1}{Q} = \frac{P_d}{\omega U} = \frac{P_{d1}}{\omega U} + \frac{P_{d2}}{\omega U} + \dots$$

$$= \frac{1}{Q_1} + \frac{1}{Q_2} + \dots$$

Reciprocals of Q add.

Dielectric and Conductor Loss

Q due to lossy dielectric

$$\epsilon = \epsilon' - j\epsilon''$$

$$Q = \frac{\epsilon'}{\epsilon''}$$

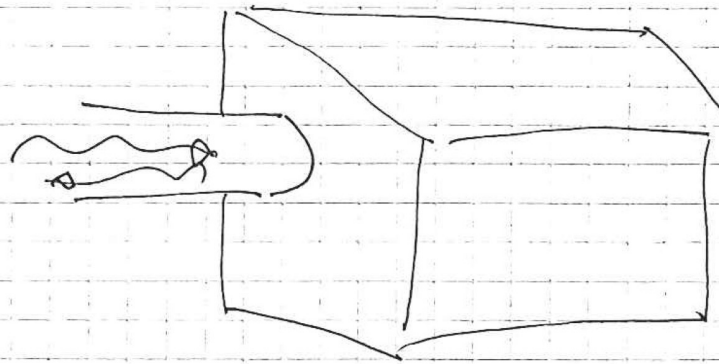
ϵ
must completely
fill cavity

Losses due to conductors

$$Q = \frac{\omega}{R_s} \left(\frac{\mu}{R_s} \right) \frac{\int d^3x |\hat{H}|^2 \leftarrow \text{energy stored}}{\int da |\hat{H}_t|^2 \leftarrow \text{losses}}$$

Coupling to Cavities

A closed box is useless

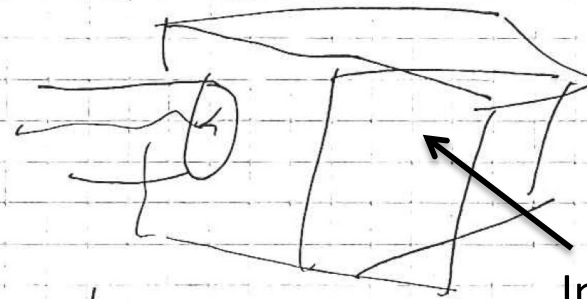


need to be able to get power in
it

Adding a hole (or coupling port
does two things) power can
come in and power can go
out.

Coupling Also Characterized by Q

Modifies Q



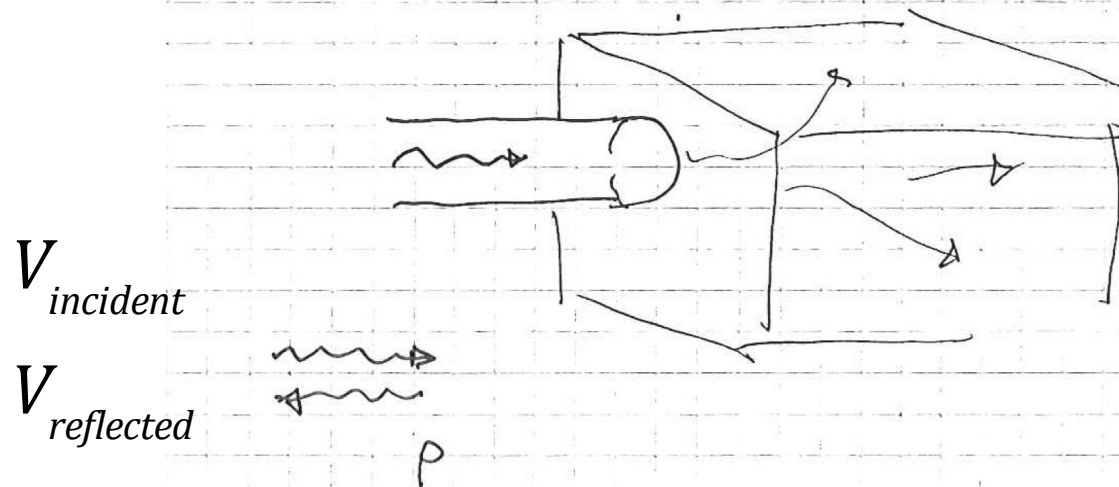
Internal losses

$$\frac{1}{Q_T} = \frac{1}{Q_{\text{internal}}} + \frac{1}{Q_{\text{coupling}}}$$

Q_{internal}

Critically Coupled

$$Q_{\text{coupling}} = Q_{\text{internal}}$$



Voltage reflection coefficient

$$\rho = - \frac{i \left(\frac{\omega}{\omega_{\text{res}}} - 1 \right) - \left(\frac{1}{2Q_i} - \frac{1}{2Q_e} \right)}{i \left(\frac{\omega}{\omega_{\text{res}}} - 1 \right) - \left(\frac{1}{2Q_i} + \frac{1}{2Q_e} \right)}$$
~~$$\rho = \frac{V_{\text{reflected}}}{V_{\text{incident}}}$$~~

Universal Response

$$\rho = \frac{i\left(\frac{\omega}{\omega_{res}} - 1\right) - \left(\frac{1}{2Q_i} - \frac{1}{2Q_e}\right)}{i\left(\frac{\omega}{\omega_{res}} - 1\right) - \left(\frac{1}{2Q_i} + \frac{1}{2Q_e}\right)} \quad \rho = \frac{V_{reflected}}{V_{incident}}$$

Knowing Q_i , Q_e and ω_{res} determines

far from resonance $\rho \approx -1$

Reflectivity at resonance

0 - if $Q_i = Q_e$

$$|\rho|_{res}^2 = \left(\frac{Q_i^{-1} - Q_e^{-1}}{Q_i^{-1} + Q_e^{-1}} \right)^2$$

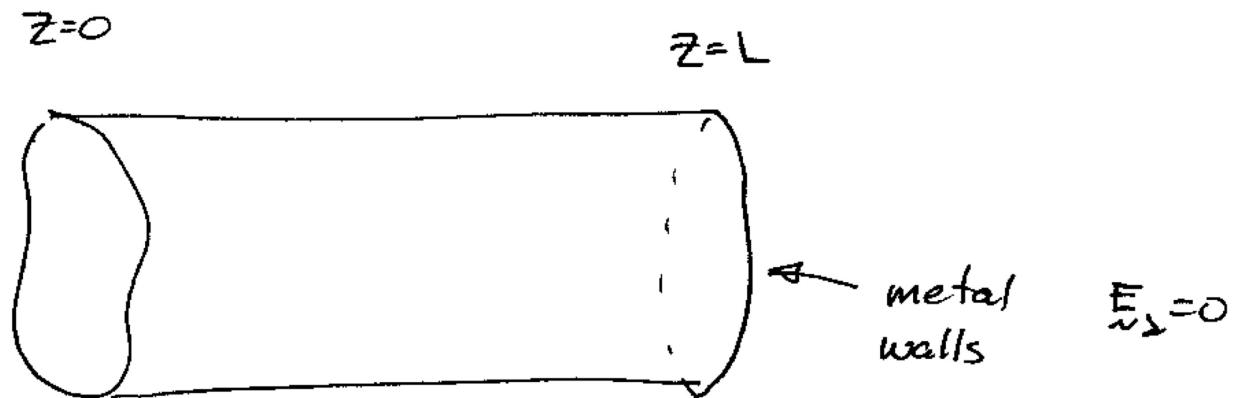
damping rate for fields $\rho \rightarrow \infty$

$$\frac{\omega}{\omega_{res}} = 1 - i \frac{1}{2Q_r}$$

$$\frac{1}{Q_r} = \frac{1}{Q_i} + \frac{1}{Q_e}$$

Waveguide Cavities

CAVITIES constructed from cylindrical
waveguides will have normal modes



Waveguide fields

$$\mathbf{E} = \text{Re} \left\{ \hat{\mathbf{E}}(x, y) \exp \left[i \left(k_z z - \omega t \right) \right] \right\}$$

$$\mathbf{H} = \text{Re} \left\{ \hat{\mathbf{H}}(x, y) \exp \left[i \left(k_z z - \omega t \right) \right] \right\}$$

$$\frac{\partial^2 \hat{H}_z}{\partial x^2} + \frac{\partial^2 \hat{H}_z}{\partial y^2} = - \left((\omega / v)^2 - k_z^2 \right) \hat{H}_z$$

$$\frac{\partial^2 \hat{E}_z}{\partial x^2} + \frac{\partial^2 \hat{E}_z}{\partial y^2} = - \left((\omega / v)^2 - k_z^2 \right) \hat{E}_z$$

$$\hat{\mathbf{E}}_{\perp} = \frac{i}{(\omega / v)^2 - k_z^2} \left[k_z \nabla_{\perp} \hat{E}_z - \omega \mu \hat{\mathbf{z}} \times \nabla_{\perp} \hat{H}_z \right]$$

$$\hat{E}_z \Big|_{\text{wall}} = 0, \quad \mathbf{n} \cdot \nabla_{\perp} \hat{H}_z \Big|_{\text{wall}} = 0$$

$$\hat{\mathbf{H}}_{\perp} = \frac{i}{(\omega / v)^2 - k_z^2} \left[k_z \nabla_{\perp} \hat{H}_z + \omega \epsilon \hat{\mathbf{z}} \times \nabla_{\perp} \hat{E}_z \right]$$

Forward and Backward Waves

TM modes

$$\hat{E}_z = \hat{E}_{z,nm}(x,y) \left(A_+ \exp(ik_z z) + A_- \exp(-ik_z z) \right)$$

$$\hat{\mathbf{E}}_{\perp} = \frac{ik_z \nabla_{\perp} \hat{E}_{z,nm}}{(\omega/v)^2 - k_z^2} \left(A_+ \exp(ik_z z) - A_- \exp(-ik_z z) \right)$$

BC: $\hat{\mathbf{E}}_{\perp}(z=0,L) = 0$

$$A_+ = A_- \quad A_+ = A_- e^{-2ik_z L}$$

$$k_z L = p\pi, \quad p = 0, 1, 2, \dots$$

TE modes

$$\hat{H}_z = H_{z,nm}(x,y) \left(A_+ \exp(ik_z z) + A_- \exp(-ik_z z) \right)$$

$$\hat{\mathbf{E}}_{\perp} = \frac{-i\omega\mu\hat{\mathbf{z}} \times \nabla_{\perp} \hat{H}_{z,nm}}{(\omega/v)^2 - k_z^2} \left(A_+ \exp(ik_z z) + A_- \exp(-ik_z z) \right)$$

$$A_+ = -A_- \quad A_+ = -A_- e^{-2ik_z L}$$

$$k_z L = p\pi, \quad p = 1, 2, \dots$$

Resonant Frequencies

$$k_{11} = \frac{\pi p}{L}$$

$p \neq 0$ not allowed

all fields zero

$$\frac{\omega^2}{c^2} \epsilon \mu = k_c^2 + \left(\frac{\pi p}{L} \right)^2$$

determined by cross section

$$\omega_{\text{Res}} = \frac{c}{\sqrt{\epsilon \mu}} \sqrt{k_c^2 + \left(\frac{\pi p}{L} \right)^2}$$

Rectangular cross section

$T E_{nmp}$
cavity mode

$$\omega_{\text{Res}} = \omega_{nmp} = \frac{c}{\sqrt{\epsilon \mu}} \sqrt{\left(\frac{n\pi}{a} \right)^2 + \left(\frac{m\pi}{b} \right)^2 + \left(\frac{p\pi}{L} \right)^2}$$

TM Modes

$$E_z = \frac{1}{2} \left\{ \hat{E}_{||}(x_{\perp}) \left[A_+ e^{i(k_z z - \omega t)} + A_- e^{-i k_z z - i \omega t} \right] + c.c. \right\}$$

$$\vec{E}_{\perp} = \frac{1}{2} \left\{ \frac{i k_{||} \nabla_{\perp} \hat{E}_{||}}{k_c^2} \left[A_+ e^{i(k_z z - \omega t)} - A_- e^{-i k_z z - i \omega t} \right] \right\}$$

at $z=0$ $A_+ - A_- = 0$ $A_+ = A_-$

at $z=L$ $A_+ e^{i k_z L} = A_- e^{-i k_z L}$

again

$$2k_{||}L = 2\pi p \quad p=0 \quad o.k.$$

Cavity Losses

$$Q = \frac{\omega U}{P_d}$$

← energy stored
← power dissipated

Quality Factor

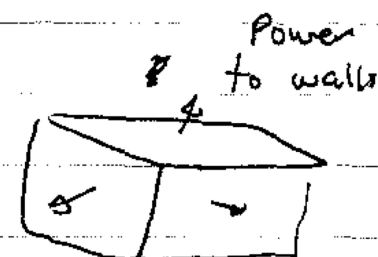
Poynting's Theorem

$$\frac{\partial}{\partial t} \int d^3x \frac{1}{4} (\epsilon |\hat{E}|^2 + \mu |\hat{H}|^2) + \int_S dA \hat{n} \cdot \frac{1}{2} \text{Re} \{ \hat{E} \times \hat{H}^* \} = 0$$

For cavity modes

$$\int d^3x \frac{\epsilon |\hat{E}|^2}{4} = \int d^3x \frac{\mu |\hat{H}|^2}{4}$$

average energy
stored in E & H
equal



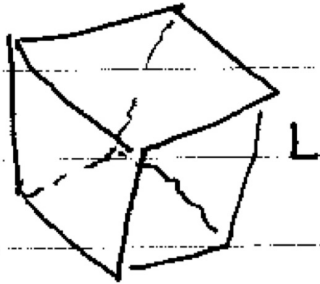
$$P_d = \int dA \frac{1}{2} R_s |\hat{H}_t|^2$$

$$Q = \frac{\omega}{c} \sqrt{\frac{\mu}{\epsilon}} \frac{1}{R_s} \frac{\int d^3x |\hat{H}|^2}{\int dA |\hat{H}_t|^2}$$

Weyl's Formula

How many modes in a cavity of volume V
have $\omega_{res} < \omega$?

Consider a cubic cavity of side L $V = L^3$



Resonant Frequencies

$$\omega_{\underline{n}} = \frac{\pi c}{L} \sqrt{n_x^2 + n_y^2 + n_z^2}$$

$$\underline{n} = (n_x, n_y, n_z)$$

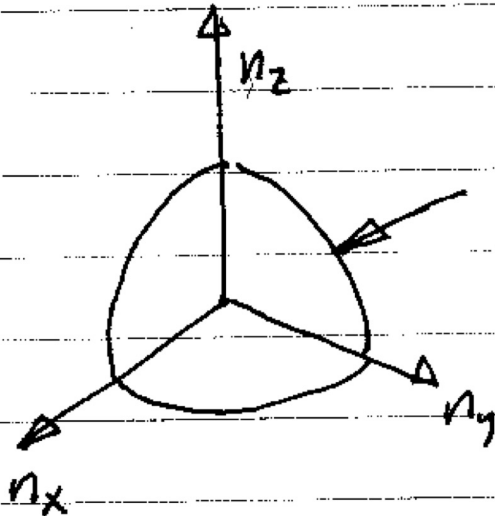
Estimate of Number of Modes

How many combinations of integers (n_x, n_y, n_z) have

$$n_x^2 + n_y^2 + n_z^2 < \left(\frac{\omega L}{\pi c}\right)^2 = \left(\frac{kL}{\pi}\right)^2$$

$$\omega_n = \frac{\pi c}{L} \sqrt{n_x^2 + n_y^2 + n_z^2}$$

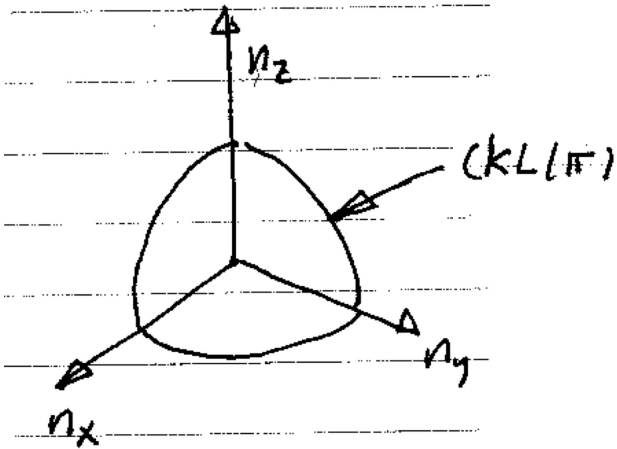
$$n = (n_x, n_y, n_z)$$



Spherical surface
of radius (kL/π)

Each combination
occupies a cube of
unit volume

Volume Inside Spherical Surface



$$N = \frac{1}{8} \frac{4}{3} \pi \left(\frac{kL}{\pi} \right)^3$$

fraction of sphere

$$N(k) = \frac{(kL)^3}{6\pi^2} = \frac{k^3 V}{6\pi^2}$$

But wait!

For each set of integers

there are 2 polarizations

For EM modes

$$N(k) = \frac{1}{3} \frac{k^3 V}{\pi^2}$$

Example

$$\text{Volume} = 1 \text{ m}^3$$

$$f = 1 \text{ GHz}$$

$$k = \frac{2\pi f}{c} = \frac{2\pi \times 10^9}{3 \times 10^8} = 21 \text{ m}^{-1}$$

$$N(k) \approx 310$$

What is the typical spacing

$$\delta k = \text{spacing in } k = \omega/c$$

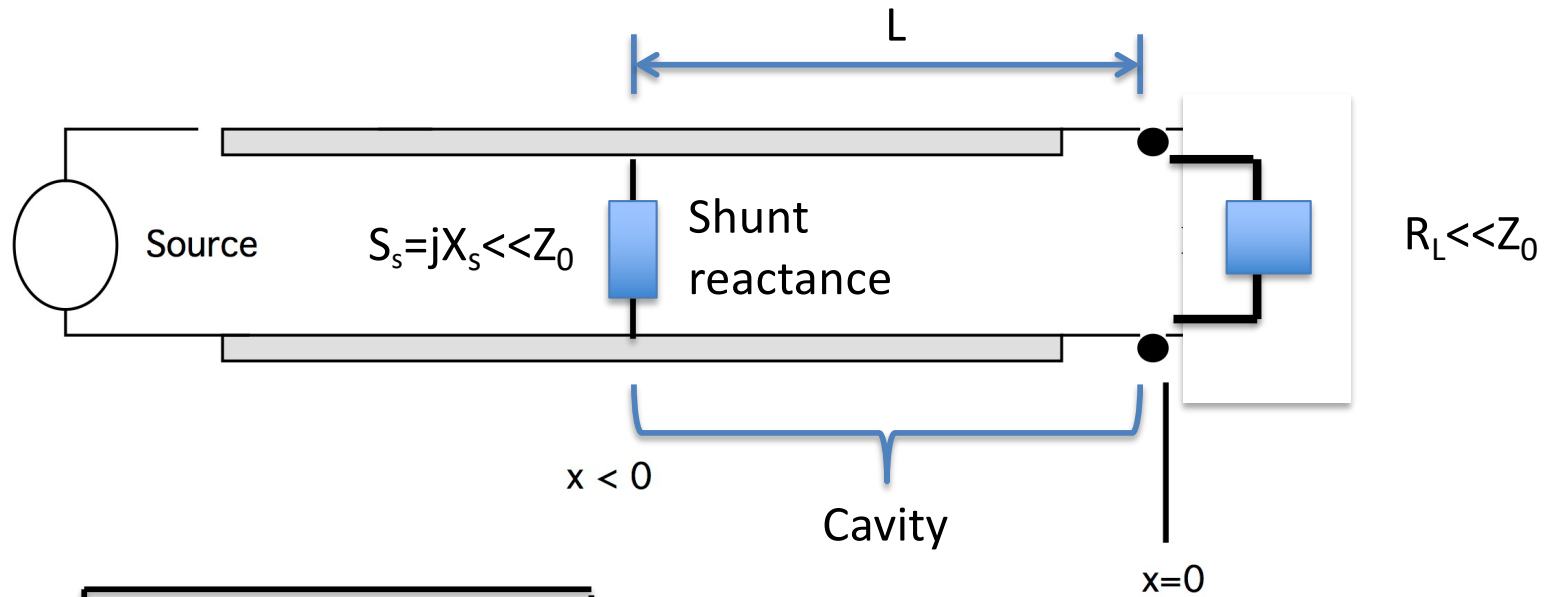
$$N(k + \delta k) = N(k) + 1$$

$$N(k + \delta k) \approx N(k) + \delta k \frac{dN}{dk} \quad \frac{dN}{dk} = \frac{k^2 V}{\pi^2}$$

$$\text{fractional spacing} \quad \frac{\delta k}{k} = \frac{1}{k \frac{dN}{dk}} = \frac{\pi^2}{k^3 V} \approx 1.07 \times 10^{-3}$$

High Q cavity model

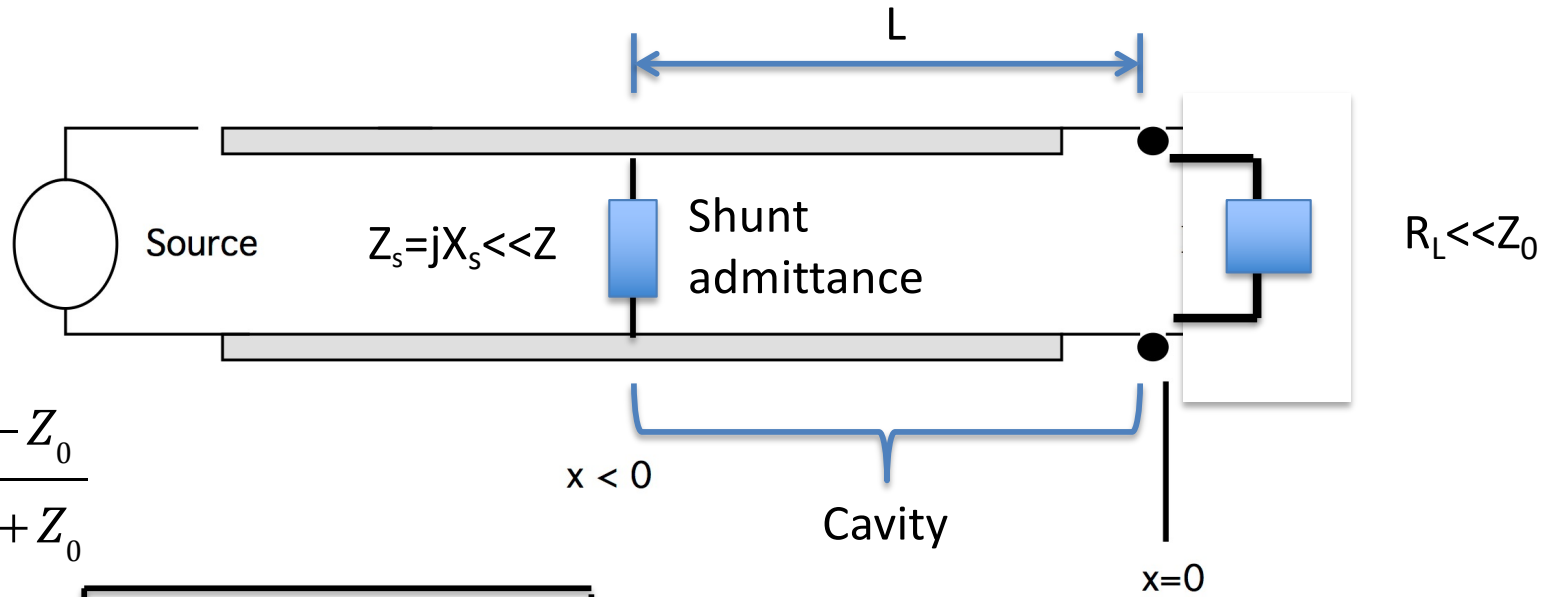
$$e^{-i\omega t} \rightarrow e^{j\omega t}$$



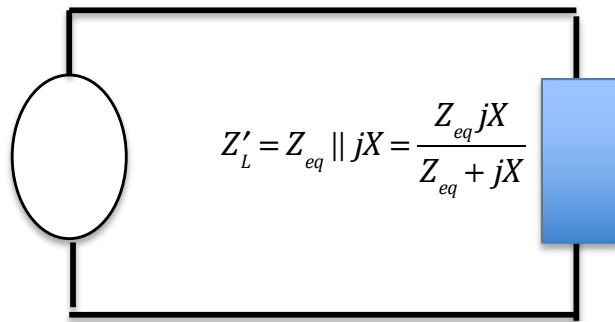
$$Z'_L = Z_{eq} \parallel jX = \frac{Z_{eq} jX}{Z_{eq} + jX}$$

$$\rho = \frac{Z'_{eq} - Z_0}{Z'_{eq} + Z_0}$$

High Q Cavity Model



$$\rho_{cav} = \frac{(Z_{eq} \parallel jX_s) - Z_0}{(Z_{eq} \parallel jX_s) + Z_0}$$

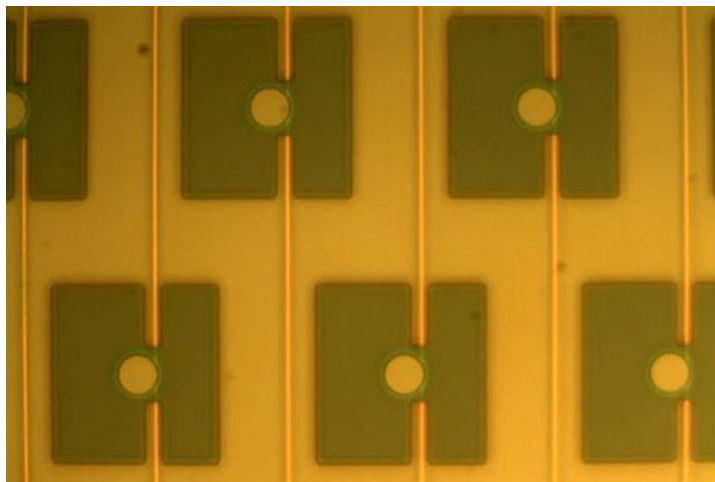
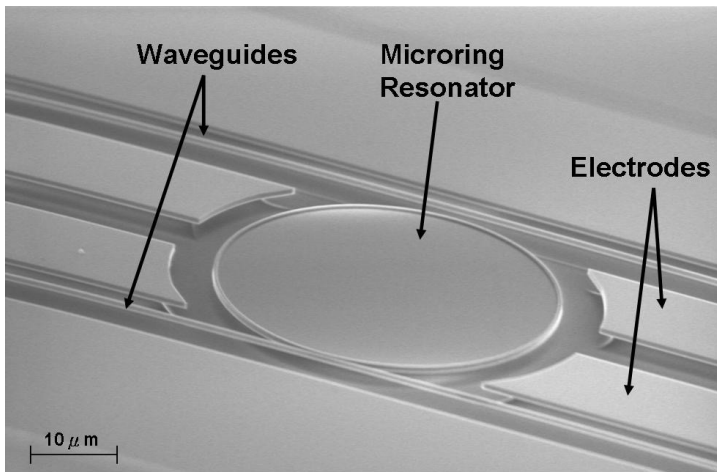


$$Z_{eq} = Z_0 \frac{R_L \cos(kL) + jZ_0 \sin(kL)}{jR_L \sin(kL) + Z_0 \cos(kL)} \approx jZ_0 \tan(kL) + R_L$$

$$\rho_{cav} \approx -\frac{j(Z_0 \tan(kL) + X_s) + R_L - X_s^2 / Z_0}{j(Z_0 \tan(kL) + X_s) + R_L + X_s^2 / Z_0} = -\frac{j2(\omega - \omega_c) / \omega_c + Q_{int}^{-1} - Q_{ext}^{-1}}{j2(\omega - \omega_c) / \omega_c + Q_{int}^{-1} + Q_{ext}^{-1}}$$

Integrated photonics

(Courtesy Edo Waks)

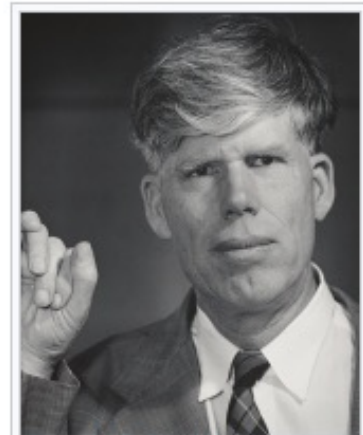


Klystron – Beam Driven HPM source

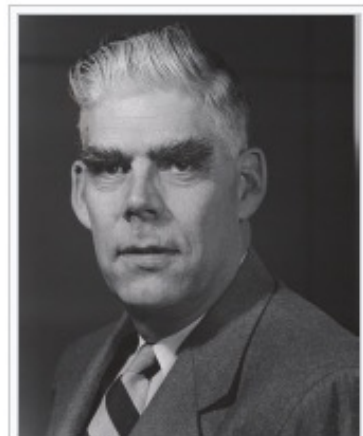
Klystron: invented in 1937 by the Varian brothers. One of the first Palo Alto High Tech. firms.

High Power Source of Microwaves

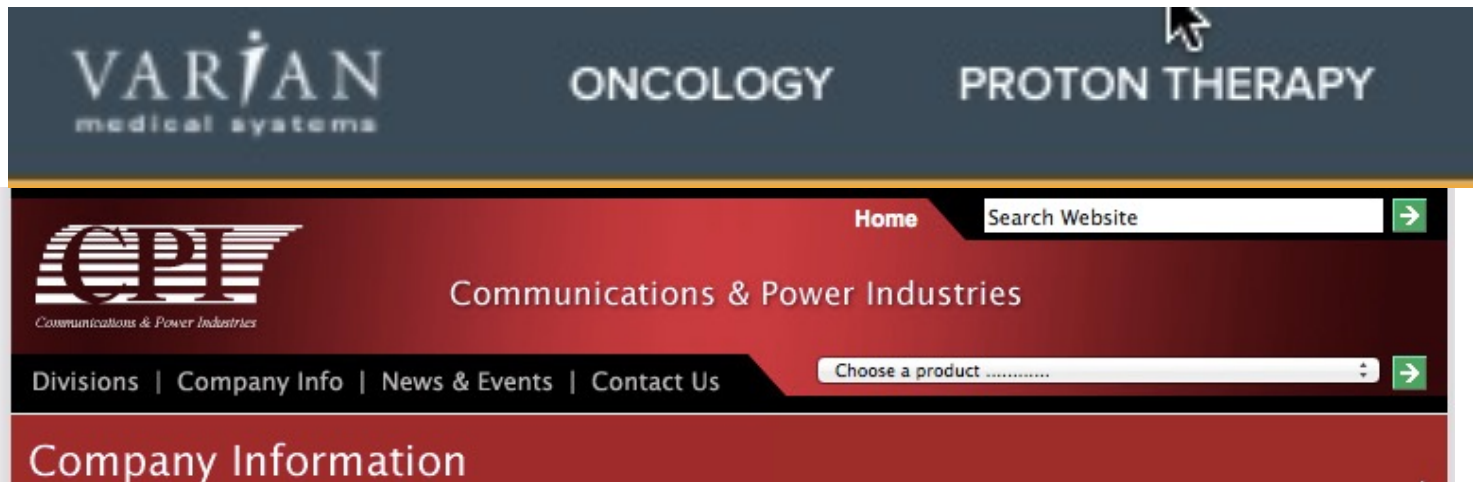
Radar, Particle Accelerators, (LHC 16 x 300 kW), etc



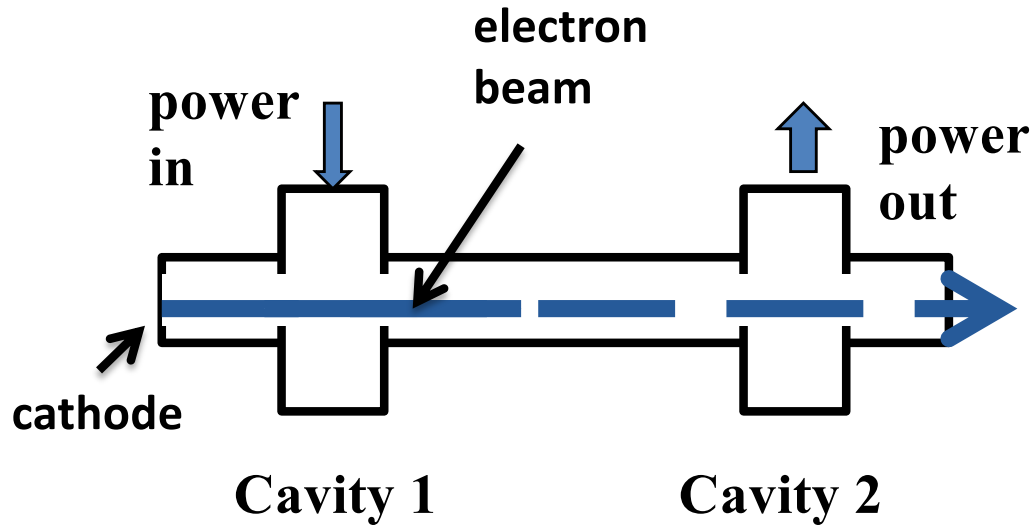
Russell Varian (1898–1959). Photograph by Ansel Adams.



Sigurd Varian (1901–1961) Photograph by Ansel Adams.

A screenshot of the Varian medical systems website. The top navigation bar is dark blue with the Varian logo on the left and 'ONCOLOGY' and 'PROTON THERAPY' on the right. Below this is a red banner with the 'CPI' logo and the text 'Communications & Power Industries'. A search bar is located in the top right of the red banner. Below the red banner is a dark blue navigation bar with links for 'Divisions', 'Company Info', 'News & Events', and 'Contact Us'. A product selection dropdown menu is also visible. The bottom section of the screenshot is a red banner with the text 'Company Information'.

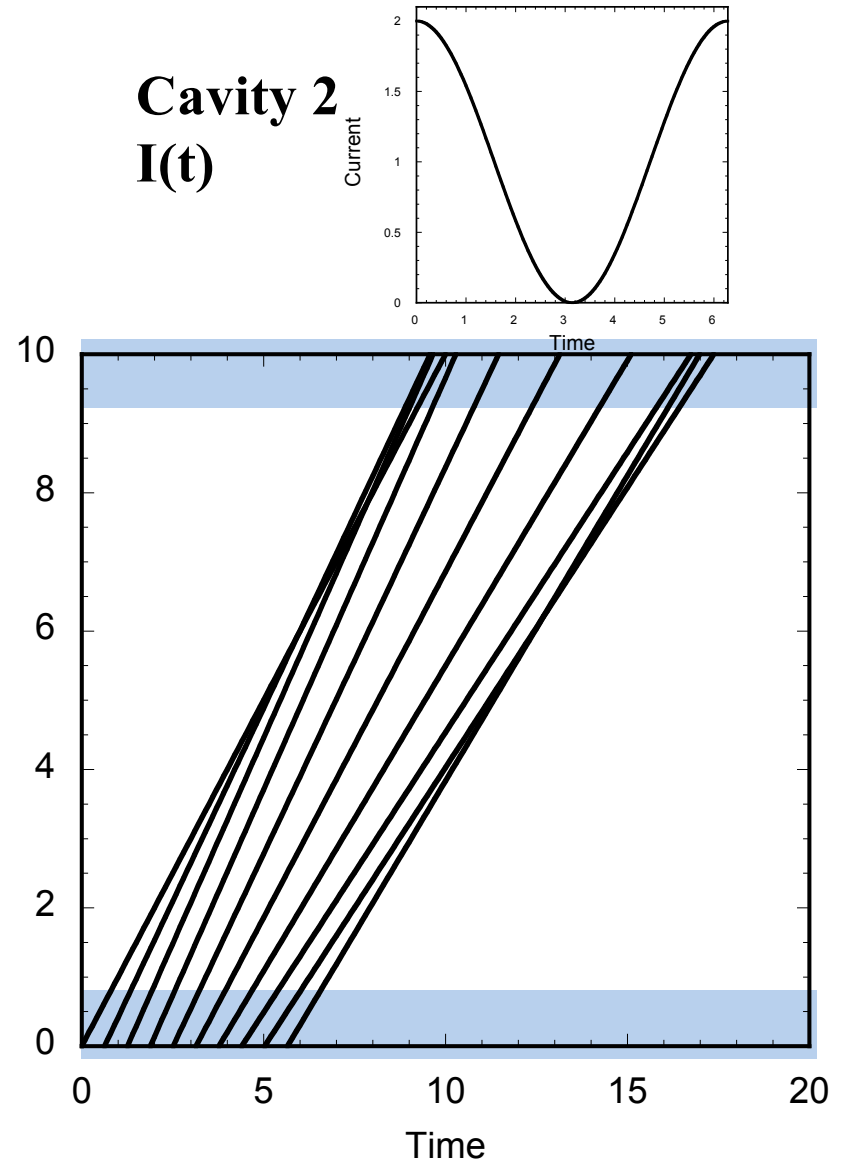
Velocity Modulation Ballistic Bunching



Field in cavity 1 gives small time dependent velocity modulation

Fast electrons catch up to slow electrons giving large current modulation.

Cavity 1



Examples

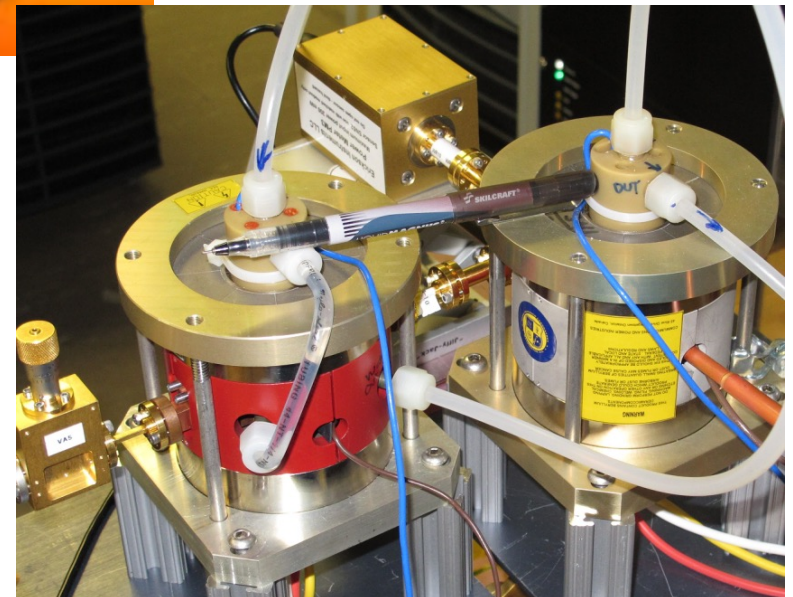


**Monica
Blank**

170 GHz CPI Gyrotron
IEEE IVEC
<http://ieeexplore.ieee.org>



**L3 Ka Band
Power Module**
<http://www.linkmicrotek.com>



Experimental high power set-up showing the CPI 218.4 GHz EIK driving the compact NRL Serpentine Waveguide (SWG) TWT.