

# Cavities

ENEE 381

# Role of Cavities

Cavities are resonant structures: Support EM modes at specific frequencies.

Used in:

- Filters

- Oscillators

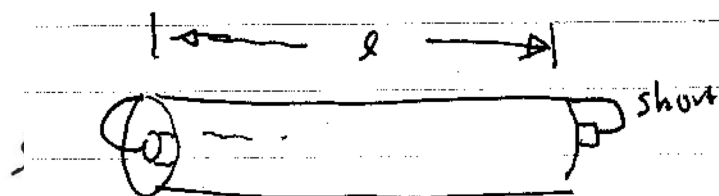
- Amplifiers

- Measurement of material properties

# Resonance

natural frequency of oscillation of fields

Example:



Transmission line

Terminated in short circuits

$$kl = n\pi$$

~~$$\omega = kv$$~~

$$k = \frac{\omega}{v}$$

$$\omega_n = \frac{n\pi v}{l}$$

$$n = 1, 2, \dots$$

$$\frac{d^2}{dx^2} V(x) + \frac{\omega^2}{v^2} V(x) = 0$$

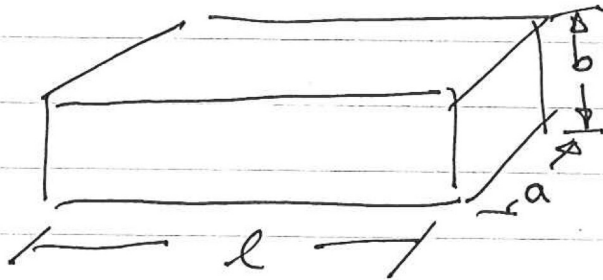
$$V(x) = V_+ e^{i\omega x/v} + V_- e^{-i\omega x/v}$$

$$V(x=0) = 0$$

$$V(x=l) = 0$$

# Enclosed Rectangular Prism

~~an~~ enclosed box

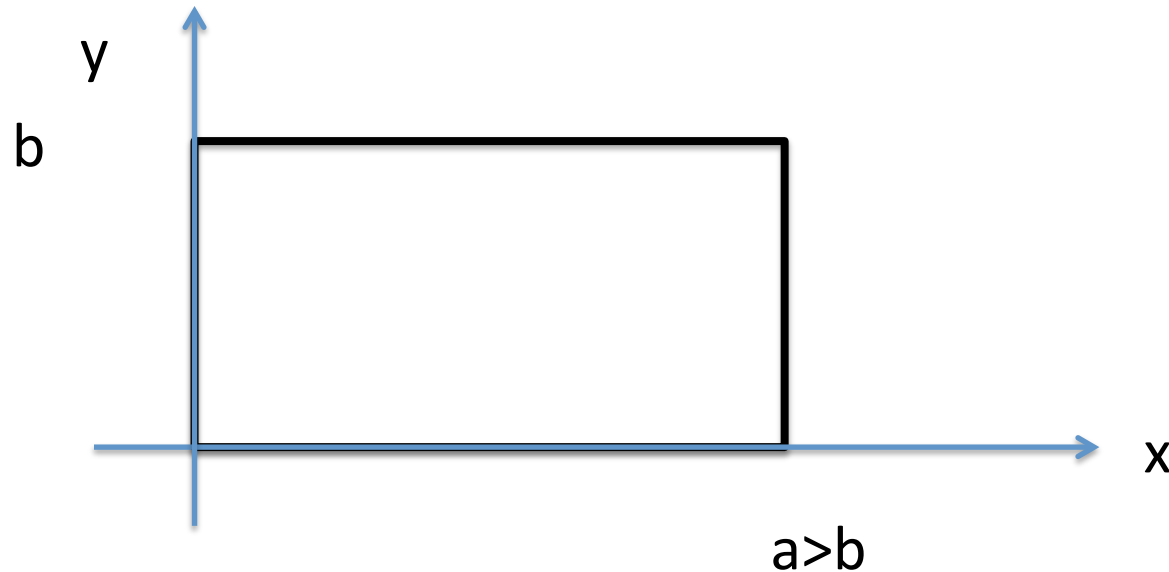


$TE_{nm}$  or  $TM_{nm}$

$$\omega_{nmp} = V \sqrt{\left(\frac{n\pi}{a}\right)^2 + \left(\frac{m\pi}{b}\right)^2 + \left(\frac{p\pi}{l}\right)^2}$$

$n, m, p$   
are  
integers

# Modes of a Rectangular WG



$$\text{TM}_{nm}: \hat{E}_z = E_0 \sin(k_x x) \sin(k_y y) \quad k_x = \frac{n\pi}{a}, \quad k_y = \frac{m\pi}{b}: \quad n, m = 1, 2, 3, \dots$$

$$\text{TE}_{nm}: \hat{H}_z = H_0 \cos(k_x x) \cos(k_y y) \quad k_x = \frac{n\pi}{a}, \quad k_y = \frac{m\pi}{b}: \quad n, m = 0^*, 1, 2, 3, \dots$$

\* one or the other, but not both

Cut-Off frequencies

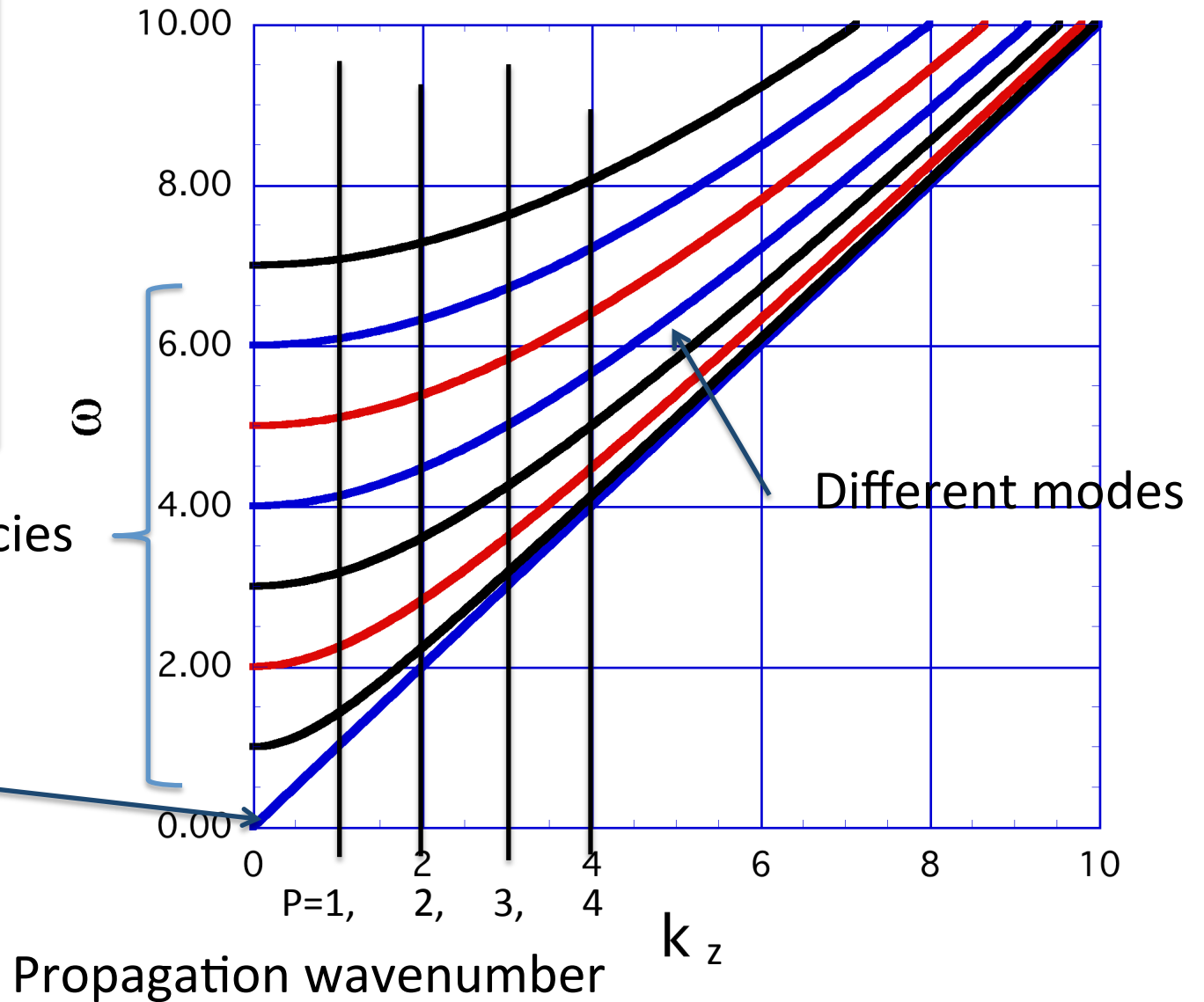
$$\omega_{c,n,m} = v \sqrt{\left(\frac{n\pi}{a}\right)^2 + \left(\frac{m\pi}{b}\right)^2}$$

# WG Dispersion Relations

$$\frac{\omega^2}{v^2} = k_{\perp}^2 + k_z^2$$
$$k_z = p \frac{\pi}{L}$$

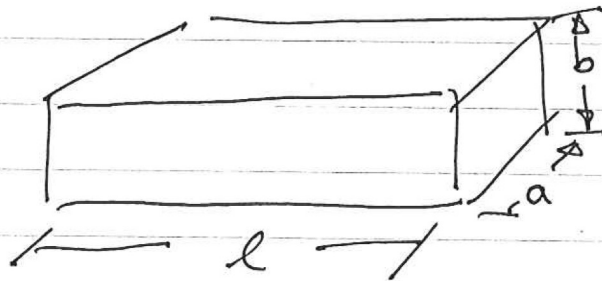
Cut off frequencies

Transmission  
Lines Only



# Enclosed Rectangular Prism

~~enclosed~~ enclosed box



TE<sub>nm</sub> or TM<sub>nm</sub>

$$\omega_{nmp} = v \sqrt{\left(\frac{n\pi}{a}\right)^2 + \left(\frac{m\pi}{b}\right)^2 + \left(\frac{p\pi}{l}\right)^2}$$

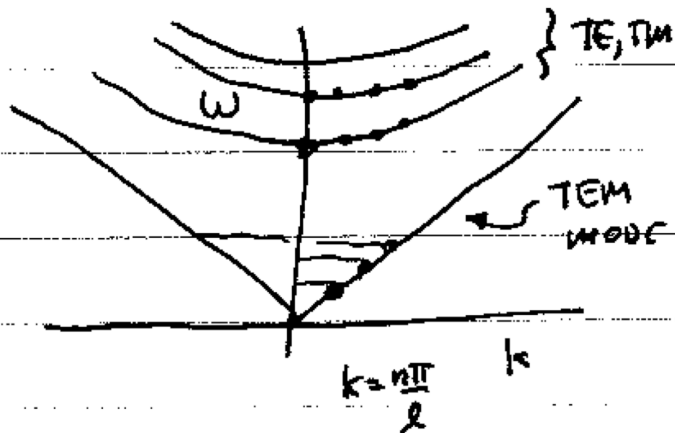
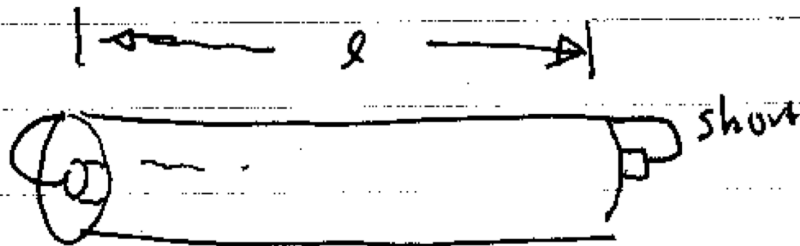
$n, m, p$   
are  
integers

$$\omega^2 = k_z^2 v^2 + \omega_{c,nm}^2$$

$$\omega_{c,nm} = v \sqrt{\left(\frac{n\pi}{a}\right)^2 + \left(\frac{m\pi}{b}\right)^2}$$

$$k_z = p \frac{\pi}{L}$$

# Transmission Line – TEM mode



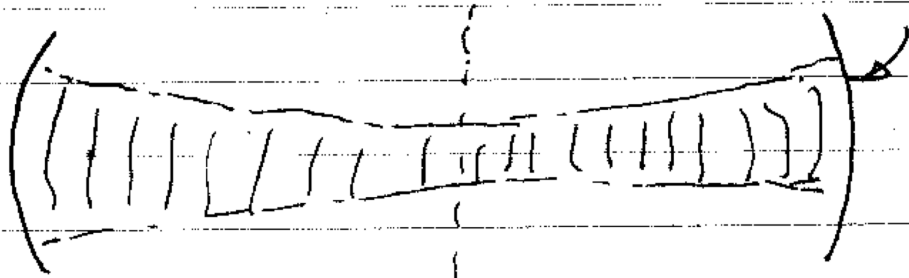
Operate at frequencies well below  
TE and TM cut-off



# Fabry-Perot Cavity

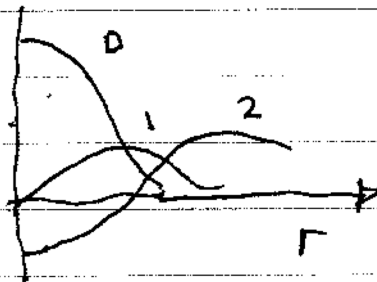
Fabry-Perot Cavity

Curved mirror



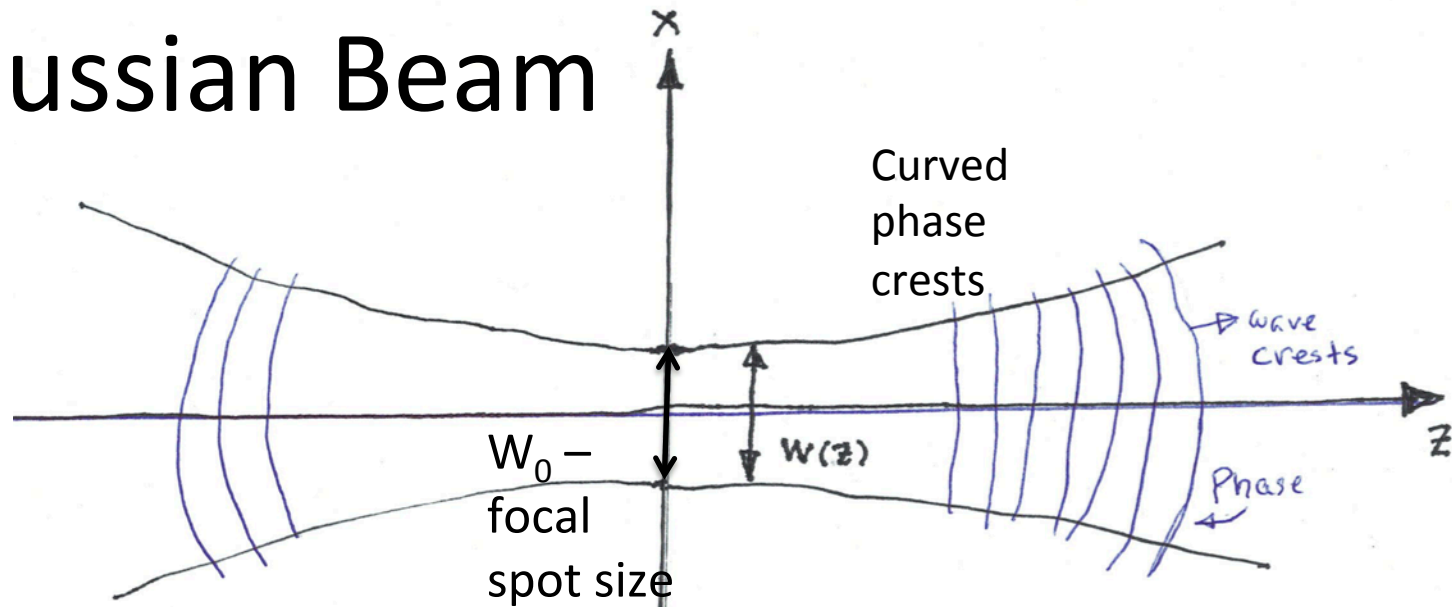
curved mirror

E



Gaussian-Laguerre  
MODES

# Gaussian Beam



$$E_{x,y}(x,y,z) = \frac{E_0}{1 + iz/Z_R} \exp \left[ -\frac{(x^2 + y^2)}{W_0^2 (1 + iz/Z_R)} + ikz \right]$$

$$W(z) = W_0 \sqrt{1 + z^2/Z_R^2}$$

$$Z_R = \frac{1}{2} kW_0^2$$

Rayleigh Length

$$E_{x,y}(0,0,z) = \frac{E_0}{1 + iz/Z_R} \exp[+ikz] = \frac{E_0}{\sqrt{1 + (z/Z_R)^2}} \exp[+ikz + i\phi(z)]$$

Guoy Phase  $\tan \phi = -z/Z_R$

# Gaussian Beam

$$E_{x,y}(x,y,z) = \frac{E_0}{1 + iz/Z_R} \exp \left[ -\frac{(x^2 + y^2)}{W_0^2 (1 + iz/Z_R)} + ikz \right]$$

$$= \underbrace{\frac{E_0}{\sqrt{1 + z^2/Z_R^2}} \exp \left[ -\frac{(x^2 + y^2)}{W_0^2 (1 + z^2/Z_R^2)} \right]}_{\text{Amplitude}} \exp \left\{ i \underbrace{\left[ kz + \frac{z(x^2 + y^2)}{Z_R W_0^2 (1 + z^2/Z_R^2)} + \phi_G \right]}_{\text{Phase}} \right\}$$

Amplitude

Phase

Pick parameters such that phase is constant on surface of mirror.

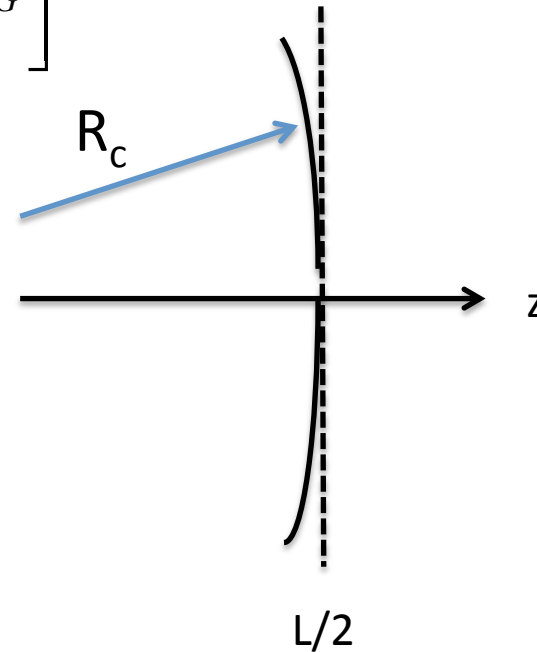
The pick k such that the phase changes by  $p\pi$  in going from one mirror to the next

Wavebeam phase  $\left[ kz + \frac{z(x^2 + y^2)}{Z_R W_0^2 (1 + z^2 / Z_R^2)} + \phi_G \right]$

Surface of mirror  $z \simeq \frac{L}{2} - \frac{(x^2 + y^2)}{2R_c}$

Will match if  $\frac{L}{2R_c} = \frac{(L/2)^2}{Z_R^2 + (L/2)^2} < 1$

Phase change mirror to mirror  $2 \left( k_p \frac{L}{2} + \phi_G \right) = p\pi$



For a given L and  $R_c$   $Z_R$  is determined above.  
Hence  $W_0$  focal spot determined.

$$Z_R = \frac{1}{2} k W_0^2$$

# Design a Fabrey-Perot resonator

Requirements:

Wavelength 1 micron =  $10^{-6}$  m.

Focal spot size = 100 microns =  $10^{-4}$  m.

Spot size on mirrors = 300 microns =  $3 \times 10^{-4}$  m

Find L and  $R_c$

Super bonus: How big must the mirrors be to keep “spill over” below 10%

# Quality Factor

Quality factor measures losses  $Q$

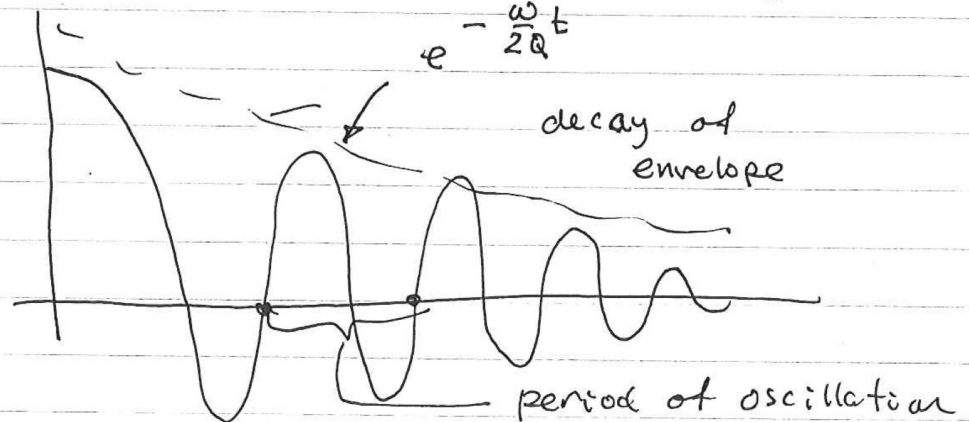
big  $Q$  low loss    low  $Q$  high loss

Time domain

$$P_{\text{in}} \left( \frac{\omega U}{P_{\text{a}}} \right) = Q$$

$V(t)$

$E(t)$



$$\frac{1}{Q} \equiv \frac{\text{Power Dissipated}}{\omega \text{ Energy Stored}}$$

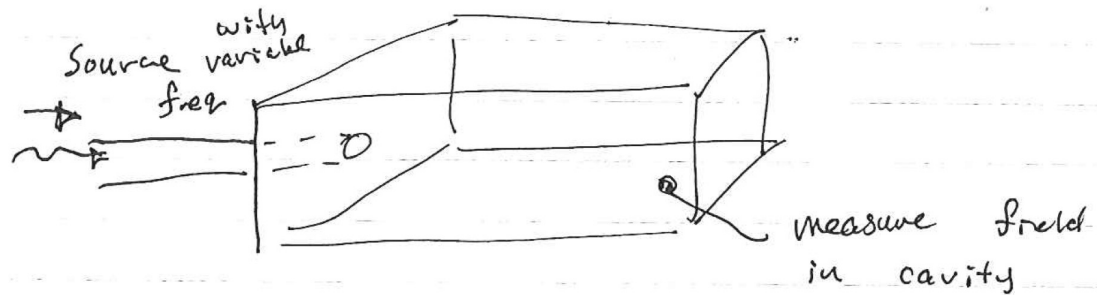
Field decay rate

$$E, H \sim \exp\left(-\frac{\omega}{2Q}t\right)$$

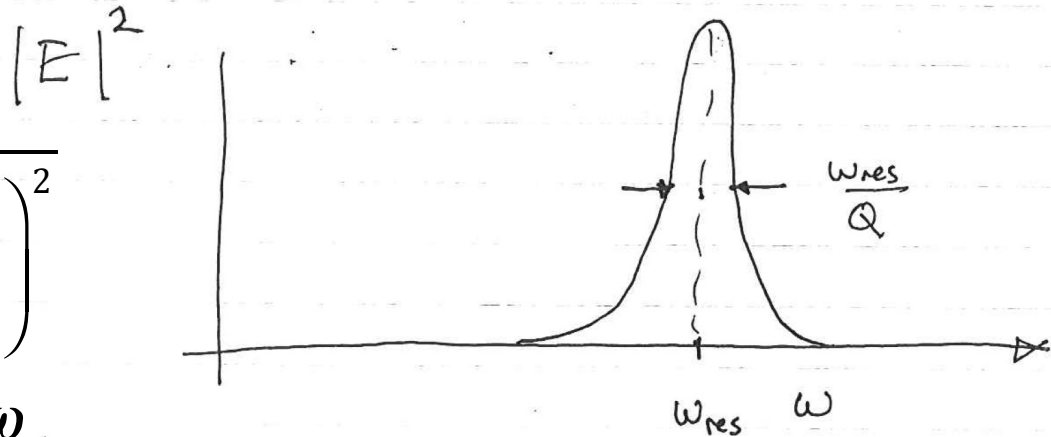
# Frequency Domain

Steady state field response

$$E, H \sim \frac{\text{Source}}{\omega - \omega_{res} + i \frac{\omega_{res}}{2Q}}$$



$$|E|^2, |H|^2 \sim \frac{1}{\left(\omega - \omega_{res}\right)^2 + \left(\frac{\omega_{res}}{2Q}\right)^2}$$



$$\text{Half maximum } \omega - \omega_{res} = \pm \frac{\omega_{res}}{2Q}$$

$$\text{Full Width at Half Maximum (FWHM)} = \frac{\omega_{res}}{Q}$$

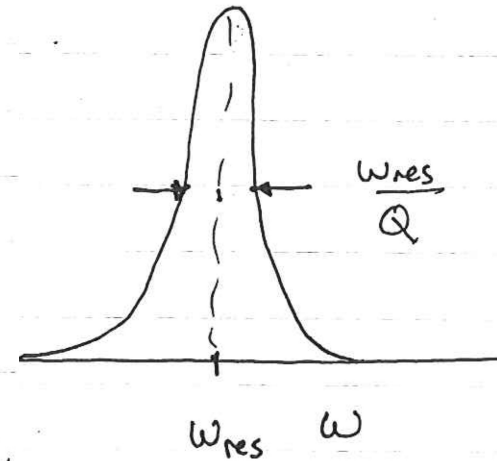
# Response Function

$$|E|^2 \propto$$

$$\frac{1}{(\omega - \omega_{res})^2 + \left(\frac{\omega_{res}}{2Q}\right)^2}$$

when  $\omega = \omega_{res} \pm \frac{\omega_{res}}{2Q}$

$|E|^2$  is  $\frac{1}{2}$  of peak value



Full width at half max of Power  
(FWHM)  $\rightarrow \frac{\omega_{res}}{Q}$



# Multiple Contributions to Loss

if. losses are small ( $Q \gg 1$ )

losses are additive

different  
loss mechan

$$\frac{1}{Q} = \frac{P_d}{\omega U} = \frac{P_{d1}}{\omega U} + \frac{P_{d2}}{\omega U} + \dots$$

$$= \frac{1}{Q_1} + \frac{1}{Q_2} + \dots$$

Reciprocals of  $Q$  add.

# Dielectric and Conductor Loss

Q due to lossy dielectric

$$\epsilon = \epsilon' - j\epsilon''$$

$$Q = \frac{\epsilon'}{\epsilon''}$$

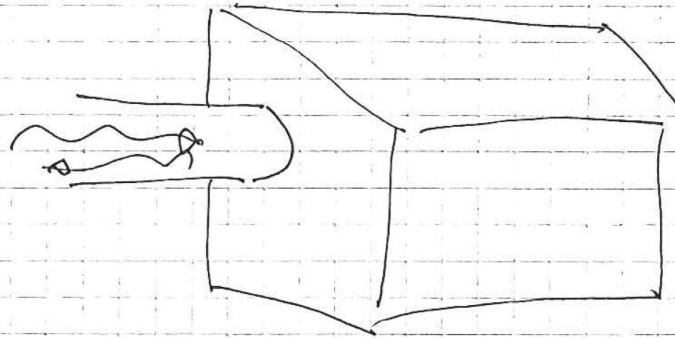
$\epsilon$   
must completely  
fill cavity

Losses due to conductors

$$Q = \frac{\omega}{R_s} \left( \frac{\mu}{R_s} \right) \frac{\int d^3x |\hat{H}|^2 \leftarrow \text{energy stored}}{\int da |\hat{H}_t|^2 \leftarrow \text{losses}}$$

# Coupling to Cavities

A closed box is useless

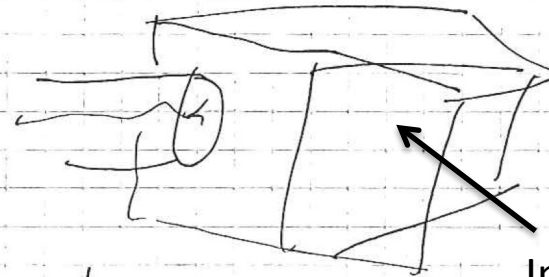


need to be able to get power in  
it

Adding a hole (or coupling port  
does two things) power can  
come in and power can go  
out.

# Coupling Also Characterized by Q

Modifies Q



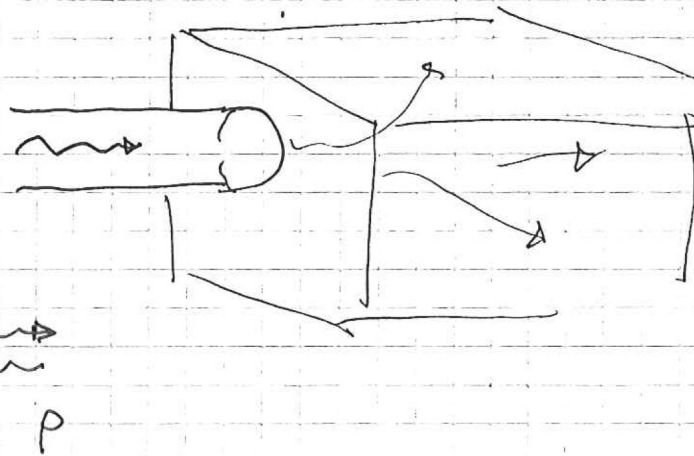
Internal losses

$$\frac{1}{Q_T} = \frac{1}{Q_{\text{internal}}} + \frac{1}{Q_{\text{coupling}}}$$

$Q_{\text{internal}}$

# Critically Coupled

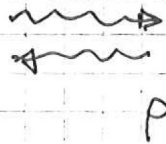
$$Q_{\text{coupling}} = Q_{\text{internal}}$$



At resonance  
all power  
from source  
is absorbed  
in cavity!

$V_{\text{incident}}$

$V_{\text{reflected}}$



Voltage reflection coefficient

$$\rho = - \frac{i \left( \frac{\omega}{\omega_{res}} - 1 \right) - \left( \frac{1}{2Q_i} - \frac{1}{2Q_e} \right)}{i \left( \frac{\omega}{\omega_{res}} - 1 \right) - \left( \frac{1}{2Q_i} + \frac{1}{2Q_e} \right)} \quad \rho = \frac{V_{\text{reflected}}}{V_{\text{incident}}}$$

# Universal Response

$$\rho = \frac{i\left(\frac{\omega}{\omega_{res}} - 1\right) - \left(\frac{1}{2Q_i} - \frac{1}{2Q_e}\right)}{i\left(\frac{\omega}{\omega_{res}} - 1\right) - \left(\frac{1}{2Q_i} + \frac{1}{2Q_e}\right)} \quad \rho = \frac{V_{reflected}}{V_{incident}}$$

Knowing  $Q_i$ ,  $Q_e$  and  $\omega_{res}$  determines

far from resonance  $\rho \approx -1$

Reflectivity at resonance

$$|\rho|_{res}^2 = \left( \frac{Q_i^{-1} - Q_e^{-1}}{Q_i^{-1} + Q_e^{-1}} \right)^2$$

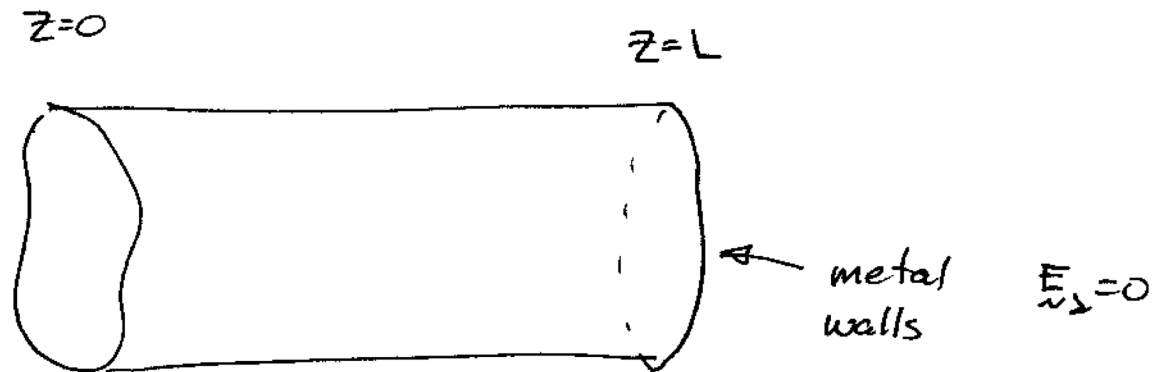
0 - if  $Q_i = Q_e$

damping rate for fields  $\rho \rightarrow \infty$

$$\frac{\omega}{\omega_{res}} = 1 - i \frac{1}{2Q_r} \quad \frac{1}{Q_r} = \frac{1}{Q_i} + \frac{1}{Q_e}$$

# Waveguide Cavities

CAVITIES constructed from cylindrical  
waveguides will have normal modes



# Waveguide fields

$$\mathbf{E} = \text{Re} \left\{ \hat{\mathbf{E}}(x, y) \exp \left[ i(k_z z - \omega t) \right] \right\}$$

$$\mathbf{H} = \text{Re} \left\{ \hat{\mathbf{H}}(x, y) \exp \left[ i(k_z z - \omega t) \right] \right\} \quad \frac{\partial^2 \hat{H}_z}{\partial x^2} + \frac{\partial^2 \hat{H}_z}{\partial y^2} = - \left( (\omega / v)^2 - k_z^2 \right) \hat{H}_z$$

$$\frac{\partial^2 \hat{E}_z}{\partial x^2} + \frac{\partial^2 \hat{E}_z}{\partial y^2} = - \left( (\omega / v)^2 - k_z^2 \right) \hat{E}_z$$

$$\hat{\mathbf{E}}_{\perp} = \frac{i}{(\omega / v)^2 - k_z^2} \left[ k_z \nabla_{\perp} \hat{E}_z - \omega \mu \hat{\mathbf{z}} \times \nabla_{\perp} \hat{H}_z \right]$$

$$\hat{E}_z \Big|_{\text{wall}} = 0, \quad \mathbf{n} \cdot \nabla_{\perp} \hat{H}_z \Big|_{\text{wall}} = 0$$

$$\hat{\mathbf{H}}_{\perp} = \frac{i}{(\omega / v)^2 - k_z^2} \left[ k_z \nabla_{\perp} \hat{H}_z + \omega \epsilon \hat{\mathbf{z}} \times \nabla_{\perp} \hat{E}_z \right]$$



# Forward and Backward Waves

TM modes

$$\text{BC: } \hat{\mathbf{E}}_{\perp}(z=0,L)=0$$

$$\hat{E}_z = \hat{E}_{z,nm}(x,y) \left( A_+ \exp(ik_z z) + A_- \exp(-ik_z z) \right) \quad A_+ = A_- \quad A_+ = A_- e^{-2ik_z L}$$

$$\hat{\mathbf{E}}_{\perp} = \frac{ik_z \nabla_{\perp} \hat{E}_{z,nm}}{(\omega/v)^2 - k_z^2} \left( A_+ \exp(ik_z z) - A_- \exp(-ik_z z) \right) \quad k_z L = p\pi, \quad p = 0, 1, 2, \dots$$

TE modes

$$\hat{H}_z = H_{z,nm}(x,y) \left( A_+ \exp(ik_z z) + A_- \exp(-ik_z z) \right) \quad A_+ = -A_- \quad A_+ = -A_- e^{-2ik_z L}$$

$$\hat{\mathbf{E}}_{\perp} = \frac{-i\omega\mu\hat{\mathbf{z}} \times \nabla_{\perp} \hat{H}_{z,nm}}{(\omega/v)^2 - k_z^2} \left( A_+ \exp(ik_z z) + A_- \exp(-ik_z z) \right) \quad k_z L = p\pi, \quad p = 1, 2, \dots$$

# Resonant Frequencies

$$k_{11} = \frac{\pi p}{L}$$

$p = 0$  not allowed

all fields zero

$$\frac{\omega^2}{c^2} \epsilon \mu = k_c^2 + \left( \frac{\pi p}{L} \right)^2$$

determined by cross section

$$\omega_{\text{Res}} = \frac{c}{\sqrt{\epsilon \mu}} \sqrt{k_c^2 + \left( \frac{\pi p}{L} \right)^2}$$

Rectangular cross section

$\omega_{\text{Res}}$   
cavity mode

$$\omega_{\text{Res}} = \omega_{\text{Res}} = \frac{c}{\sqrt{\epsilon \mu}} \sqrt{\left( \frac{m\pi}{a} \right)^2 + \left( \frac{n\pi}{b} \right)^2 + \left( \frac{p\pi}{L} \right)^2}$$

# TM Modes

$$E_z = \frac{1}{2} \left\{ \hat{E}_{||}(x_{\perp}) \left[ A_+ e^{i(k_z z - \omega t)} + A_- e^{-i k_z z - i \omega t} \right] + c.c. \right\}$$

$$\vec{E}_{\perp} = \frac{1}{2} \left\{ \frac{i k_{||} \nabla_{\perp} \hat{E}_{||}}{k_c^2} \left[ A_+ e^{i(k_z z - \omega t)} - A_- e^{-i k_z z - i \omega t} \right] \right\}$$

at  $z=0$   $A_+ - A_- = 0$   $A_+ = A_-$

at  $z=L$   $A_+ e^{i k_z L} = A_- e^{-i k_z L}$

again

$$2k_{||}L = 2\pi p \quad p=0 \quad \text{o.k.}$$

# Cavity Losses

$$Q = \frac{\omega U}{P_d}$$

$\omega U$  ← energy stored  
 $P_d$  ← power dissipated

Quality Factor

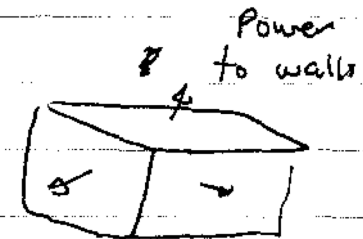
Poynting's Theorem

$$\frac{\partial}{\partial t} \int d^3x \frac{1}{4} (\epsilon |\hat{\mathbf{E}}|^2 + \mu |\hat{\mathbf{H}}|^2) + \int_S dA \hat{n} \cdot \frac{1}{2} \text{Re} \{ \hat{\mathbf{E}} \times \hat{\mathbf{H}} \} = 0$$

For cavity modes

$$\int d^3x \frac{\epsilon |\hat{\mathbf{E}}|^2}{4} = \int d^3x \frac{\mu |\hat{\mathbf{H}}|^2}{4}$$

average energy  
stored in  $\mathbf{E}$  &  $\mathbf{H}$   
equal



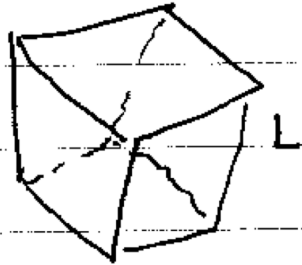
$$P_d = \int dA \frac{1}{2} R_s |\hat{\mathbf{H}}_t|^2$$

$$Q = \frac{\omega}{c} \sqrt{\frac{\mu}{\epsilon}} \frac{1}{R_s} \frac{\int d^3x |\hat{\mathbf{H}}|^2}{\int dA |\hat{\mathbf{H}}_t|^2}$$

# Weyl's Formula

How many modes in a cavity of volume  $V$   
have  $\omega_{res} < \omega$ ?

Consider a cubic cavity of side  $L$   $V = L^3$



Resonant Frequencies

$$\omega_{\underline{n}} = \frac{\pi c}{L} \sqrt{n_x^2 + n_y^2 + n_z^2}$$

$$\underline{n} = (n_x, n_y, n_z)$$

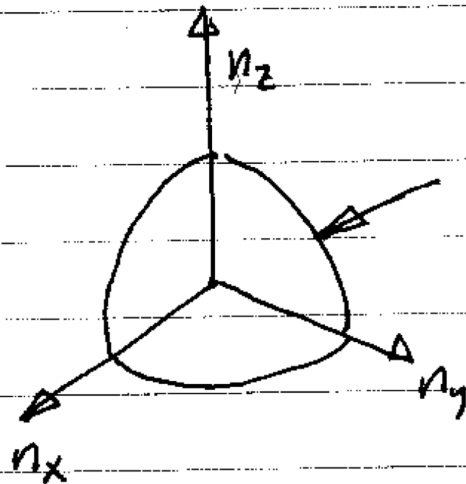
# Estimate of Number of Modes

How many combinations of integers  $(n_x, n_y, n_z)$  have

$$n_x^2 + n_y^2 + n_z^2 < \left(\frac{\omega L}{\pi c}\right)^2 = \left(\frac{kL}{\pi}\right)^2$$

$$\omega_n = \frac{\pi c}{L} \sqrt{n_x^2 + n_y^2 + n_z^2}$$

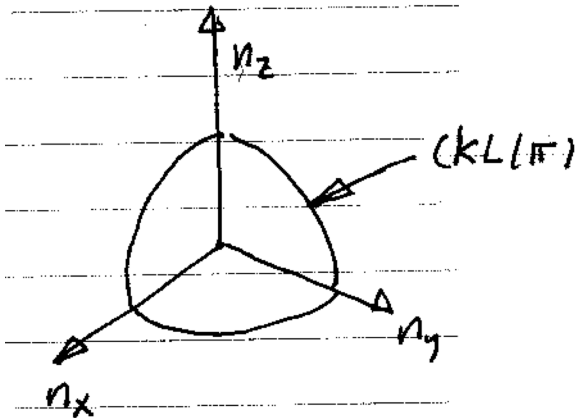
$$\mathbf{n} = (n_x, n_y, n_z)$$



Spherical surface  
of radius  $(kL/\pi)$

Each combination  
occupies a cube of  
unit volume

# Volume Inside Spherical Surface



$$N = \frac{1}{8} \frac{4}{3} \pi \left( \frac{kL}{\pi} \right)^3$$

fraction of sphere

$$N(k) = \frac{(kL)^3}{6\pi^2} = \frac{k^3 V}{6\pi^2}$$

But wait!

For each set of integers  
there are 2 polarizations

For EM modes

$$N(k) = \frac{1}{3} \frac{k^3 V}{\pi^2}$$

# Example

$$\text{Volume} = 1 \text{ m}^3$$

$$f = 1 \text{ GHz}$$

$$k = \frac{2\pi f}{c} = \frac{2\pi \times 10^9}{3 \times 10^8} = 21 \text{ m}^{-1}$$

$$N(k) \approx 310$$

What is the typical spacing

$$\delta k = \text{spacing in } k = \omega/c$$

$$N(k + \delta k) = N(k) + 1$$

$$N(k + \delta k) \approx N(k) + \delta k \frac{dN}{dk}$$

$$\frac{dN}{dk} = \frac{k^2 V}{\pi^2}$$

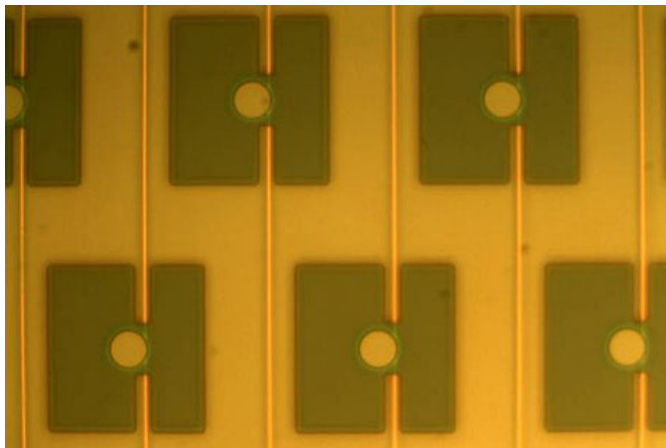
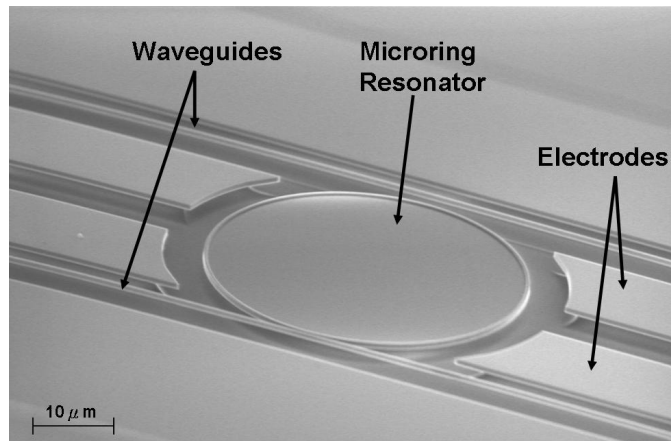
fractional  
spacing

$$\frac{\delta k}{k} = \frac{1}{k \frac{dN}{dk}} = \frac{\pi^2}{k^3 V} \approx 1.07 \times 10^{-3}$$



# Integrated photonics

(Courtesy Edo Waks)

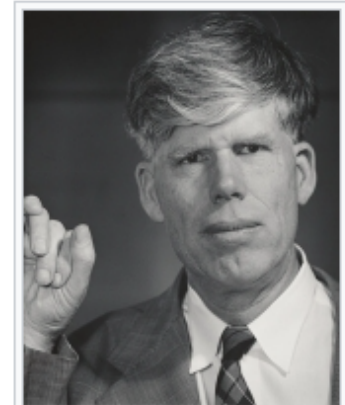


# Klystron – Beam Driven HPM source

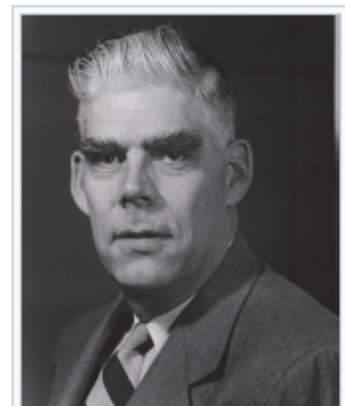
**Klystron:** invented in 1937 by the Varian brothers. One of the first Palo Alto High Tech. firms.

## High Power Source of Microwaves

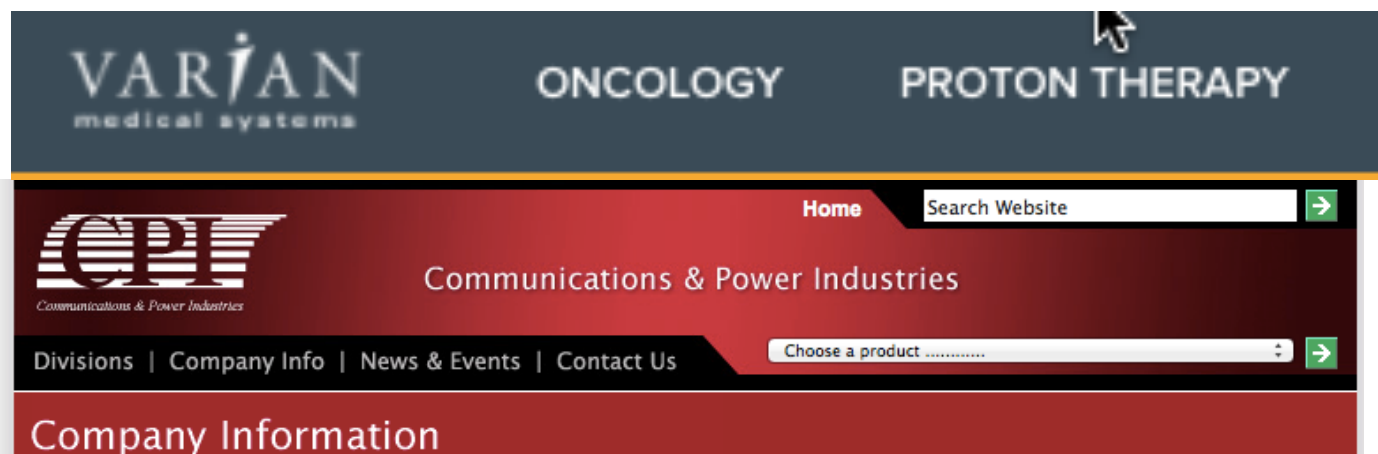
Radar, Particle Accelerators, (LHC 16 x 300 kW), etc



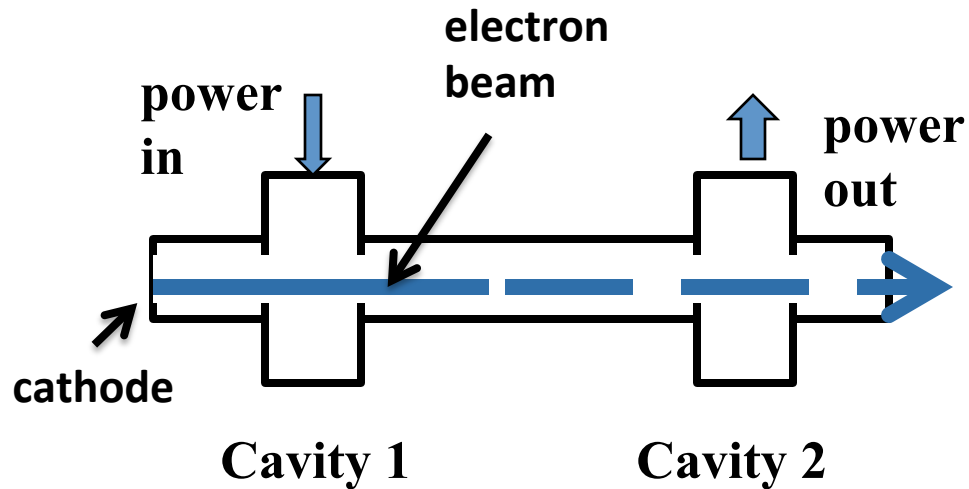
Russell Varian (1898–1959). Photograph by Ansel Adams.



Sigurd Varian (1901–1961) Photograph by Ansel Adams.

A screenshot of the Varian Medical Systems website. The top navigation bar is dark blue with the Varian logo and links for 'ONCOLOGY' and 'PROTON THERAPY'. Below this is a red banner for 'CPI Communications & Power Industries' with a search bar and a 'Home' link. The bottom section is dark red with a 'Choose a product' dropdown menu and a 'Company Information' link.

# Velocity Modulation      Ballistic Bunching

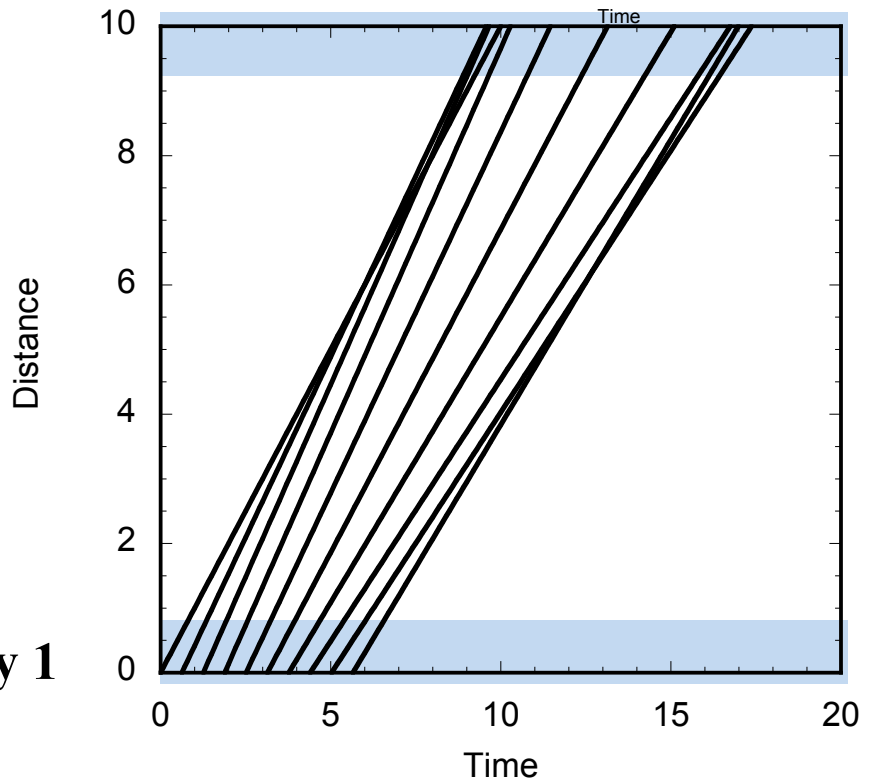
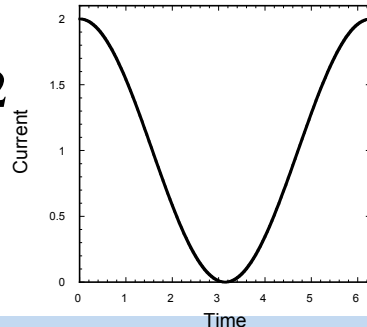


**Field in cavity 1 gives small time dependent velocity modulation**

**Fast electrons catch up to slow electrons giving large current modulation.**

**Cavity 1**

**Cavity 2  
 $I(t)$**



# Examples



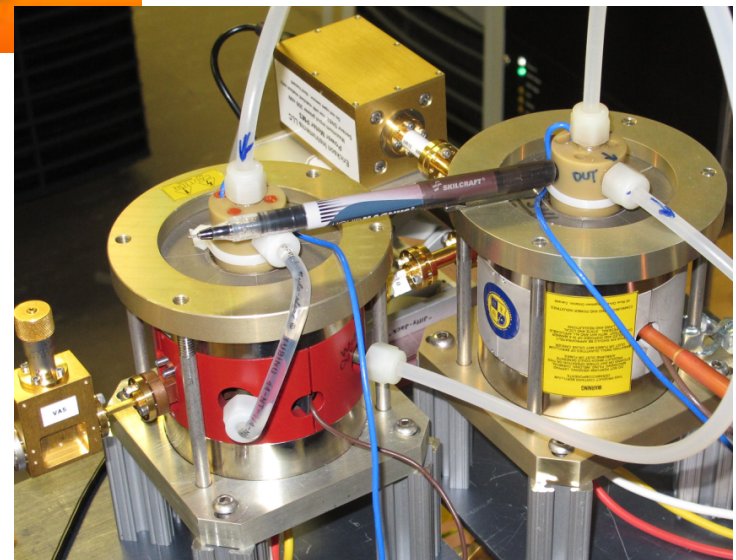
**Monica  
Blank**

**170 GHz CPI Gyrotron**  
IEEE IVEC

<http://ieeexplore.ieee.org>



**L3 Ka Band  
Power Module**  
[http://  
www.linkmicrotek.com](http://www.linkmicrotek.com)



Experimental high power set-up showing the CPI  
218.4 GHz EIK driving the compact NRL Serpentine  
Waveguide (SWG) TWT.