### ENEE 222 EXAM 2: SAMPLE PROBLEMS

Calculators will not be allowed.

The following formulas will be given on the exam sheet:

$$Time \ Domain \qquad Frequency \ Domain$$

$$x[n] = \frac{1}{N} \sum_{k=0}^{N-1} X[k] e^{j(2\pi/N)kn} \iff X[k] = \sum_{n=0}^{N-1} x[n] e^{-j(2\pi/N)kn}$$

$$\mathbf{x} = \frac{1}{N} \mathbf{V} \mathbf{X} \iff \mathbf{X} = \mathbf{W} \mathbf{x}$$

$$\alpha \mathbf{x} + \beta \mathbf{y} \iff \mathbf{X} = \mathbf{W} \mathbf{x}$$

$$\alpha \mathbf{x} + \beta \mathbf{y} \iff \mathbf{\alpha} \mathbf{X} + \beta \mathbf{Y}$$

$$\mathbf{x}^* \iff \mathbf{R} \mathbf{X}^*$$

$$\mathbf{R} \mathbf{x} \iff \mathbf{R} \mathbf{X}$$

$$\mathbf{P}^m \mathbf{x} \iff \mathbf{R} \mathbf{X}$$

$$\mathbf{P}^m \mathbf{x} \iff \mathbf{P}^m \mathbf{X}$$

$$\mathbf{X} \iff \mathbf{P}^m \mathbf{X}$$

$$\mathbf{X} \iff \mathbf{N} \mathbf{R} \mathbf{x}$$

$$\mathbf{x} \diamond \mathbf{y} \iff \frac{1}{N} (\mathbf{X} \circledast \mathbf{Y})[k] = \frac{1}{N} \mathbf{X}^T \mathbf{P}^k \mathbf{R} \mathbf{Y}$$

$$(\mathbf{x} \circledast \mathbf{y})[n] = \mathbf{x}^T \mathbf{P}^n \mathbf{R} \mathbf{y} \iff \mathbf{X} \diamond \mathbf{Y}$$

#### **PROBLEM 1**

The continuous-time real-valued signal s(t) is a sum of two real-valued sinusoids of frequency no higher than 25 Hz, plus a DC offset (i.e., a constant). It is sampled at times  $t = nT_s$ , where n = 0, ..., 19and  $T_s = 0.02$  seconds. The amplitude and phase spectra of the sampled signal are shown in the figure below.



Use the information given above to write an equation for s(t), valid for all  $t \in \mathbf{R}$ .

The time-domain signal

has DFT given by

$$\mathbf{S} = \begin{bmatrix} 4 & 4 - j(2 - \sqrt{2}) & 2 & 4 + j(2 + \sqrt{2}) & 8 & 4 - j(2 + \sqrt{2}) & 2 & 4 + j(2 - \sqrt{2}) \end{bmatrix}^T$$

- (i) Compute the numerical value of a b + c d.
- (ii) Compute the DFT X of the time-domain signal

$$\mathbf{x} = \begin{bmatrix} a & b & c & d & a & b & c & d \end{bmatrix}^T$$

(iii) Compute the DFT Y of the time-domain signal

$$\mathbf{y} = \begin{bmatrix} a & b & c & d & -a & -b & -c & -d \end{bmatrix}^T$$

#### **PROBLEM 3**

The time-domain signal

$$\mathbf{s} = \left[\begin{array}{cccc} a & b & c & d & e & f\end{array}\right]^T$$

has DFT given by

$$\mathbf{S} = \begin{bmatrix} S_0 & S_1 & S_2 & S_3 & S_4 & S_5 \end{bmatrix}^T$$

(i) Determine (in terms of  $a, \ldots, f$ ) the time-domain signal **x** whose DFT is given by

$$\mathbf{X} = \begin{bmatrix} S_0 - S_3 & S_1 - S_4 & S_2 - S_5 & S_3 - S_0 & S_4 - S_1 & S_5 - S_2 \end{bmatrix}^T$$

(ii) Determine (in terms of  $a, \ldots, f$ ) the time-domain signal **y** whose DFT is given by

$$Y[k] = S_k \cos(2\pi k/3) , \qquad k = 0, \dots, 5$$

#### **PROBLEM 4**

The time-domain signals

$$\mathbf{x} = \begin{bmatrix} 1 & 2 & 0 & 1 \end{bmatrix}^T$$
 and  $\mathbf{y} = \begin{bmatrix} 3 & -1 & 4 & -2 \end{bmatrix}^T$ 

have DFT's given by

$$\mathbf{X} = \begin{bmatrix} A & B & C & D \end{bmatrix}^T \quad \text{and} \quad \mathbf{Y} = \begin{bmatrix} E & F & G & H \end{bmatrix}^T$$

(i) Compute the time-domain signal **s** whose DFT is given by

$$\mathbf{S} = \left[ \begin{array}{ccc} AE & BH & CG & DF \end{array} \right]^T$$

(ii) Let **u** be the time-domain signal defined by the circular convolution

$$\mathbf{u} = \begin{bmatrix} A & B & C & D \end{bmatrix}^T \circledast \begin{bmatrix} 3 & -1 & 4 & -2 \end{bmatrix}^T$$

Write an expression for the DFT  $\mathbf{U}$  of  $\mathbf{u}$  in terms of E, F, G and H.

The time-domain signal  $\mathbf{x}$  shown below takes three distinct values (zero being one of them) and has DFT  $\mathbf{X}$ . For simplicity, let  $X[k] = X_k$ .



(i) For which values of k (if any) does  $X_k$  equal 0?

(ii) For which pairs (k, k') (if any) does  $X_k$  equal  $X_{k'}$ ?

(iii) Express the DFT **S** of the time-domain signal **s** shown below (left) in terms of  $X_k$ 's.

(iv) Express the DFT Y of the time-domain signal y shown below (right) in terms of  $X_k$ 's.



### PROBLEM 6

The DFT of the signal

$$\mathbf{x} = \begin{bmatrix} a & b & c & d & 0 & 0 & 0 \end{bmatrix}^T$$

is given by

$$\mathbf{X} = \begin{bmatrix} X_0 & X_1 & X_2 & X_3 & X_4 & X_5 & X_6 & X_7 \end{bmatrix}^T$$

(i) Determine the DFT S of the time-domain signal

$$\mathbf{s} = \left[ \begin{array}{ccc} c & d & a & b \end{array} \right]^T$$

(ii) Compute the value of  $\cos(n\pi/2)$  for  $n = 0, \ldots, 7$ .

(iii) Using your answer to part (ii) above, determine the time-domain signal y whose DFT is given by

$$\mathbf{Y} = \begin{bmatrix} X_0 & 0 & -X_2 & 0 & X_4 & 0 & -X_6 & 0 \end{bmatrix}^T$$

A continuous-time signal is given by

$$x(t) = A_1 \cos(2\pi f_1 t + \phi_1) + A_2 \cos(2\pi f_2 t + \phi_2) + z(t)$$

where z(t) is noise. The signal x(t) is sampled every 2.0 ms for a total of 50 samples. The figure shows the graph of the magnitude (or amplitude) spectrum of the 50-point signal as a function of the frequency index  $k = 0, \ldots, 49$ .



Based on the given graph, what are your estimates of  $f_1$  and  $f_2$ ?

(You may assume that no aliasing has occurred, i.e., the sampling rate of 500 samples/sec is no less than twice each of  $f_1$  and  $f_2$ .)

### PROBLEM 8

The continuous-time signal

$$x(t) = -4 + 5\cos(700\pi t + 0.2) + 2\cos(450\pi t - 0.8)$$

is sampled every  $T_s = 0.001$  seconds starting at t = 0. The first N = 40 samples are stored in the vector **x**.

(i) Verify that the frequencies of the sinusoidal components of x(t) become (after sampling) Fourier frequencies for the N-point vector  $\mathbf{x}$ .

(ii) Compute the nonzero entries in the DFT X of x. Give your answers in either Cartesian or polar form, clearly showing the corresponding indices k.

(iii) Sketch the amplitude and phase spectra |X[k]| and  $\angle X[k]$  as functions of the frequency index  $k = 0, \ldots, 39$ .

## PROBLEM 9

Let

$$\mathbf{A} = \begin{bmatrix} -3 & 1 & 3 & 9\\ 9 & -3 & 1 & 3\\ 3 & 9 & -3 & 1\\ 1 & 3 & 9 & -3 \end{bmatrix}$$

(i) Carefully compute  $\mathbf{A}^T \mathbf{A}$ . What special property do the columns of  $\mathbf{A}$  have?

(ii) The vector  $\mathbf{b} = \begin{bmatrix} 2 & -1 & 8 & -9 \end{bmatrix}^T$  can be written as a linear combination of three columns of  $\mathbf{A}$ . Determine that linear combination (i.e., its coefficients) without solving a  $4 \times 4$  system.

Consider the matrix of Fourier sinusoids

$$\mathbf{V} = \begin{bmatrix} \mathbf{v}^{(0)} & \mathbf{v}^{(1)} & \mathbf{v}^{(2)} & \mathbf{v}^{(3)} \end{bmatrix} = \begin{bmatrix} 1 & 1 & 1 & 1 \\ 1 & j & -1 & -j \\ 1 & -1 & 1 & -1 \\ 1 & -j & -1 & j \end{bmatrix}$$

and the vector  $\mathbf{s} = \begin{bmatrix} 6 & 2 & -1 & 4 \end{bmatrix}^T$ .

(i) Determine the coefficient vector  $\mathbf{c}$  such that  $\mathbf{s} = \mathbf{V}\mathbf{c}$ .

(ii) Display the vector  $\hat{\mathbf{s}}$ , which is the least-squares approximation of  $\mathbf{s}$  in terms of  $\mathbf{v}^{(1)}$  and  $\mathbf{v}^{(3)}$  only.

(iii) Consider the vector

$$\mathbf{x} = \mathbf{s} + \begin{bmatrix} z+u & z-u & z+u & z-u \end{bmatrix}^T,$$

where z and u are arbitrary complex constants. Let  $\hat{\mathbf{x}}$  be its least-squares approximation in terms of  $\mathbf{v}^{(1)}$  and  $\mathbf{v}^{(3)}$ . Explain why  $\hat{\mathbf{x}} = \hat{\mathbf{s}}$ , where  $\hat{\mathbf{s}}$  was defined in part (ii).